

150 mA, Tiny CMOS LDO With Shutdown

Features

- AEC-Q100 Automotive Qualified, See [Product Identification System](#)
- Space-saving 5-Pin SC-70 and SOT-23 Packages
- Extremely Low Operating Current for Longer Battery Life: 53 μ A (typ.)
- Very Low Dropout Voltage
- Rated 150 mA Output Current
- Requires Only 1 μ F Ceramic Output Capacitance
- High Output Voltage Accuracy: $\pm 0.5\%$ (typical)
- 10 μ s (typ.) Wake-Up Time from $\overline{\text{SHDN}}$
- Power-Saving Shutdown Mode: 0.05 μ A (typ.)
- Overcurrent and Overtemperature Protection
- Pin-Compatible Upgrade for Bipolar Regulators

Applications

- Cellular/GSM/PHS Phones
- Battery-Operated Systems
- Portable Computers
- Medical Instruments
- Electronic Games
- Pagers

General Description

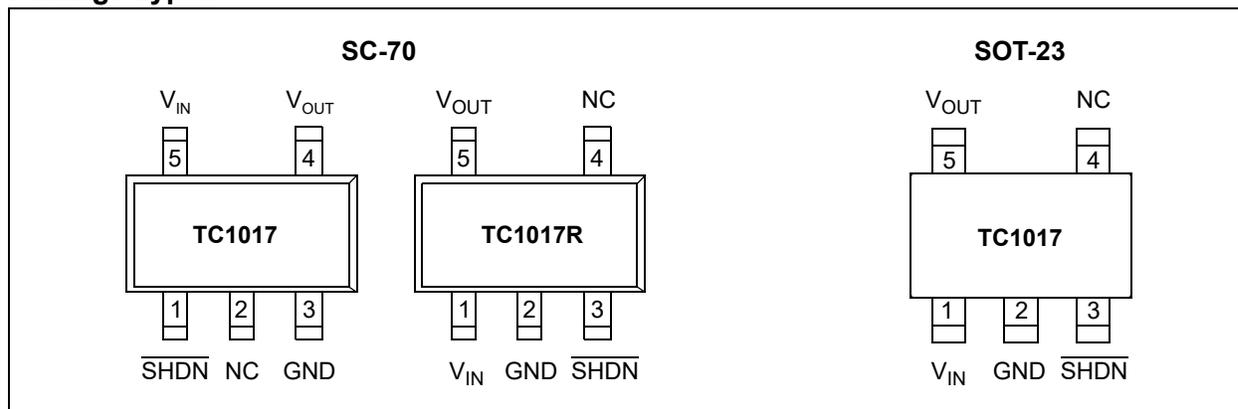
The TC1017 is a high-accuracy (typically $\pm 0.5\%$) CMOS upgrade for bipolar Low Dropout regulators (LDOs). It is offered in a SC-70 or SOT-23 package. The SC-70 package represents a 50% footprint reduction versus the popular SOT-23 package and is offered in two pinouts to make board layout easier.

Developed specifically for battery-powered systems, the TC1017's CMOS construction consumes only 53 μ A typical supply current over the entire 150 mA operating load range. This can be as much as 60 times less than the quiescent operating current consumed by bipolar LDOs.

The TC1017 is designed to be stable, over the entire input voltage and output current range, with low-value (1 μ F) ceramic or tantalum capacitors. This helps to reduce board space and save cost. Additional integrated features, such as shutdown, overcurrent and overtemperature protection, further reduce the board space and cost of the entire voltage-regulating application.

Key performance parameters for the TC1017 include low dropout voltage (285 mV typical at 150 mA output current), low supply current while shutdown (0.05 μ A typical) and fast stable response to sudden input voltage and load changes.

Package Types



TC1017

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Input Voltage	6.5V
Power Dissipation	Internally Limited (Note 7)
Maximum Voltage On Any Pin	$V_{IN} + 0.3V$ to $-0.3V$
ESD Protection on All Pins (Note 1)	
HBM	±2 kV
MM	±200V
CDM	±1.5 kV

† **Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Note 1: Testing was performed per AEC-Q100 standard. For additional information contact your local Microchip sales office.

ELECTRICAL CHARACTERISTICS

Electrical Specifications: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$ Boldface type specifications apply for junction temperatures of $-40^\circ C$ to $+125^\circ C$.						
Parameter	Sym.	Min.	Typ.	Max.	Units	Test Conditions
Input Operating Voltage	V_{IN}	2.7	—	6.0	V	Note 1
Maximum Output Current	I_{OUTMAX}	100	—	—	mA	Note 1 $V_{IN} \geq 3V$ and $V_{IN} \geq (V_R + 2.5\%) + V_{DROPOUTMAX}$
		150	—	—		
Output Voltage	V_{OUT}	$V_R - 2.5\%$	$V_R \pm 0.5\%$	$V_R + 2.5\%$	V	Note 2
V_{OUT} Temperature Coefficient	TCV_{OUT}	—	40	—	ppm/ $^\circ C$	Note 3
Line Regulation	$[(\Delta V_{OUT} / \Delta V_{IN})] / V_R$	—	0.04	0.2	%/V	$(V_R + 1V) < V_{IN} < 6V$
Load Regulation (Note 4)	$ \Delta V_{OUT} / V_R$	—	0.38	1.5	%	$I_L = 0.1 \text{ mA to } I_{OUTMAX}$
Dropout Voltage (Note 5)	$V_{IN} - V_{OUT}$	—	2	—	mV	$I_L = 100 \mu A$ $I_L = 50 \text{ mA}$ $I_L = 100 \text{ mA}$ $I_L = 150 \text{ mA}$
		—	90	200		
		—	180	350		
		—	285	500		
Supply Current	I_{IN}	—	53	90	μA	$\overline{SHDN} = V_{IH}$, $I_L = 0$
Shutdown Supply Current	I_{INSD}	—	0.05	2	μA	$\overline{SHDN} = 0V$

Note 1: The minimum V_{IN} has to meet two conditions: $V_{IN} \geq 2.7V$ and $V_{IN} \geq (V_R + 2.5\%) + V_{DROPOUT}$.

2: V_R is the regulator voltage setting. For example: $V_R = 1.8V, 2.7V, 2.8V, 3.0V$.

3:

$$TCV_{OUT} = \frac{(V_{OUTMAX} - V_{OUTMIN}) \times 10^6}{V_{OUT} \times \Delta T}$$

4: Regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 0.1 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

5: Dropout voltage is defined as the input-to-output differential at which the output voltage drops 2% below its nominal value at a 1V differential.

6: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to I_{LMAX} at $V_{IN} = 6V$ for $t = 10 \text{ msec}$.

7: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e., T_A , T_J , θ_{JA}). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see [Section 5.1 “Thermal Shutdown”](#), for more details.

8: Output current is limited to 120 mA (typ) when V_{OUT} is less than 0.5V due to a load fault or short-circuit condition.

ELECTRICAL CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $SHDN > V_{IH}$, $T_A = +25^\circ C$ Boldface type specifications apply for junction temperatures of $-40^\circ C$ to $+125^\circ C$.						
Parameter	Sym.	Min.	Typ.	Max.	Units	Test Conditions
Power Supply Rejection Ratio	PSRR	—	58	—	dB	$f = 1 \text{ kHz}$, $I_L = 50 \text{ mA}$
Wake-Up Time (from Shutdown mode)	t_{WK}	—	10	—	μs	$V_{IN} = 5V$, $I_L = 60 \text{ mA}$, $C_{IN} = C_{OUT} = 1 \mu F$, $f = 100 \text{ Hz}$
Settling Time (from Shutdown mode)	t_S	—	32	—	μs	$V_{IN} = 5V$, $I_L = 60 \text{ mA}$, $C_{IN} = 1 \mu F$, $C_{OUT} = 1 \mu F$, $f = 100 \text{ Hz}$
Output Short-Circuit Current	I_{OUTSC}	—	120	—	mA	$V_{OUT} = 0V$, Average Current (Note 8)
Thermal Regulation	V_{OUT}/P_D	—	0.04	—	V/W	Notes 6, 7
Thermal Shutdown Die Temperature	T_{SD}	—	160	—	$^\circ C$	
Thermal Shutdown Hysteresis	ΔT_{SD}	—	10	—	$^\circ C$	
Output Noise	eN	—	800	—	nV/ \sqrt{Hz}	$f = 10 \text{ kHz}$
SHDN Input High Threshold	V_{IH}	45	—	—	% V_{IN}	$V_{IN} = 2.7V$ to $6.0V$
SHDN Input Low Threshold	V_{IL}	—	—	15	% V_{IN}	$V_{IN} = 2.7V$ to $6.0V$

Note 1: The minimum V_{IN} has to meet two conditions: $V_{IN} \geq 2.7V$ and $V_{IN} \geq (V_R + 2.5\%) + V_{DROPOUT}$.

2: V_R is the regulator voltage setting. For example: $V_R = 1.8V, 2.7V, 2.8V, 3.0V$.

3:

$$TCV_{OUT} = \frac{(V_{OUTMAX} - V_{OUTMIN}) \times 10^6}{V_{OUT} \times \Delta T}$$

4: Regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 0.1 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

5: Dropout voltage is defined as the input-to-output differential at which the output voltage drops 2% below its nominal value at a 1V differential.

6: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to I_{LMAX} at $V_{IN} = 6V$ for $t = 10 \text{ msec}$.

7: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e., T_A , T_J , θ_{JA}). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see **Section 5.1 "Thermal Shutdown"**, for more details.

8: Output current is limited to 120 mA (typ) when V_{OUT} is less than 0.5V due to a load fault or short-circuit condition.

TEMPERATURE CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $V_{DD} = +2.7V$ to $+6.0V$ and $V_{SS} = GND$.						
Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Temperature Ranges						
Specified Temperature Range	T_A	-40	—	+125	$^\circ C$	Extended Temperature parts
Operating Temperature Range	T_A	-40	—	+125	$^\circ C$	
Storage Temperature Range	T_A	-65	—	+150	$^\circ C$	
Thermal Package Resistances3						
Thermal Resistance, 5L-SOT23	θ_{JA}	—	255	—	$^\circ C/W$	
Thermal Resistance, 5L-SC-70	θ_{JA}	—	450	—	$^\circ C/W$	

2.0 TYPICAL PERFORMANCE CHARACTERISTICS

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $SHDN > V_{IH}$, $T_A = +25^\circ C$.

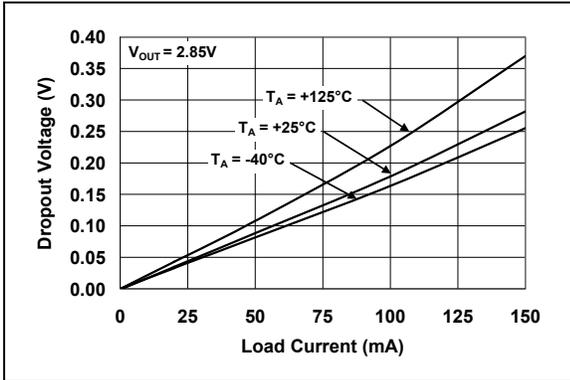


FIGURE 2-1: Dropout Voltage vs. Output Current.

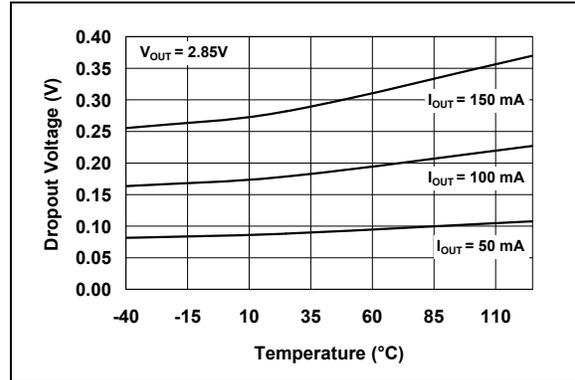


FIGURE 2-4: Dropout Voltage vs. Temperature.

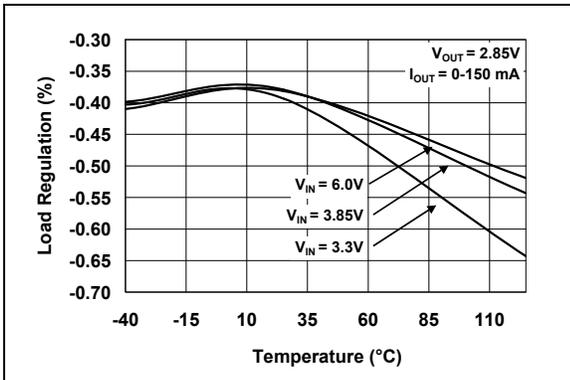


FIGURE 2-2: Load Regulation vs. Temperature.

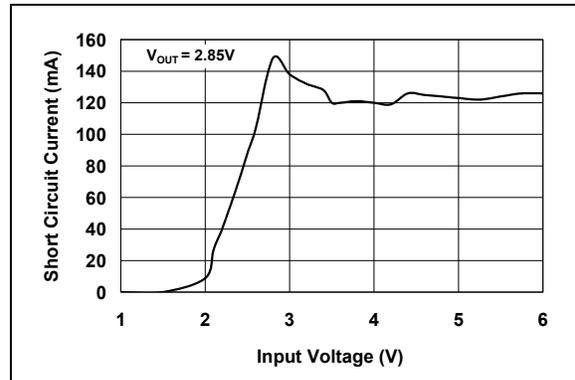


FIGURE 2-5: Short-Circuit Current vs. Input Voltage.

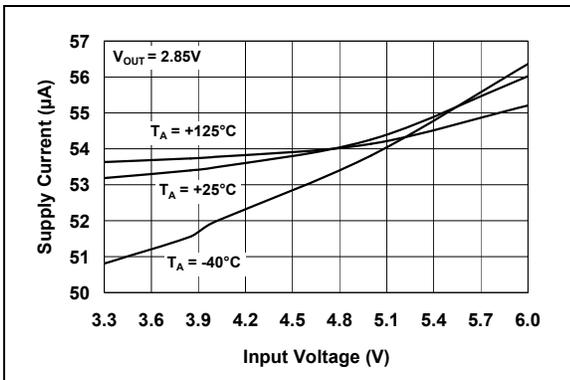


FIGURE 2-3: Supply Current vs. Input Voltage.

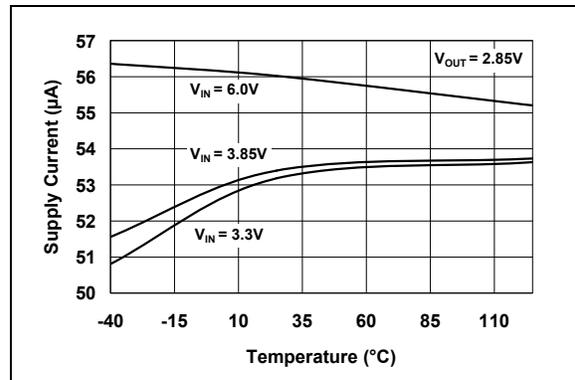


FIGURE 2-6: Supply Current vs. Temperature.

Note: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$.

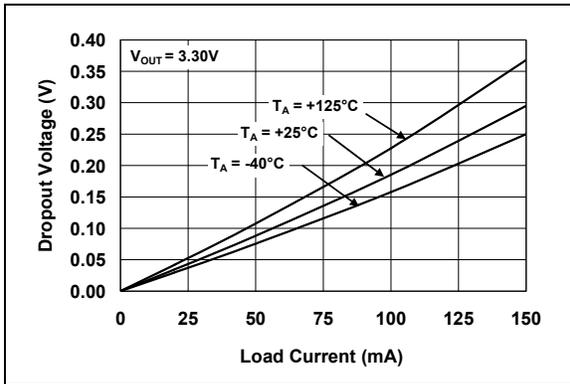


FIGURE 2-7: Dropout Voltage vs. Output Current.

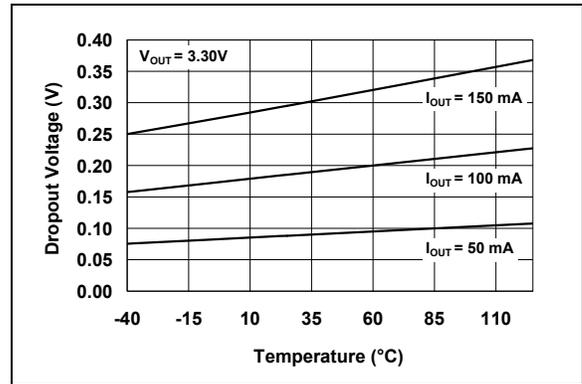


FIGURE 2-10: Dropout Voltage vs. Temperature.

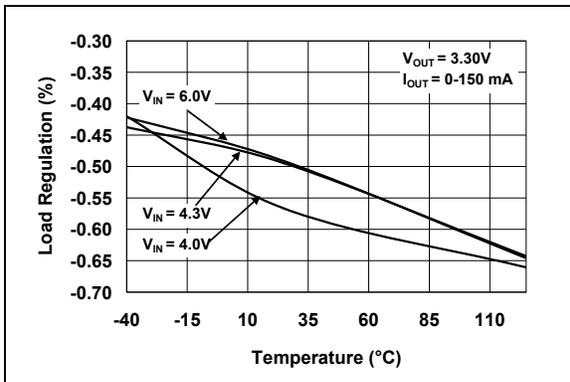


FIGURE 2-8: Load Regulation vs. Temperature.

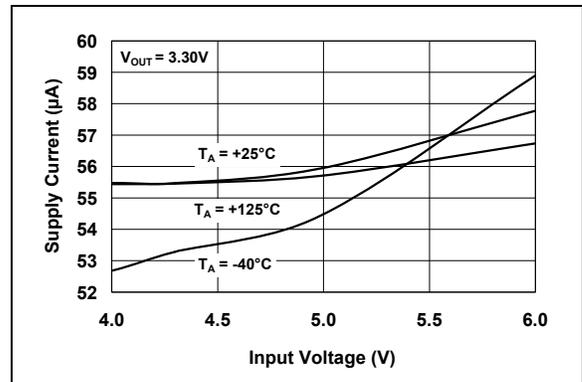


FIGURE 2-11: Supply Current vs. Input Voltage.

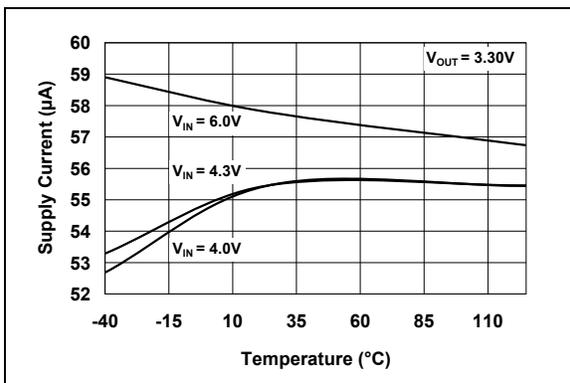


FIGURE 2-9: Supply Current vs. Temperature.

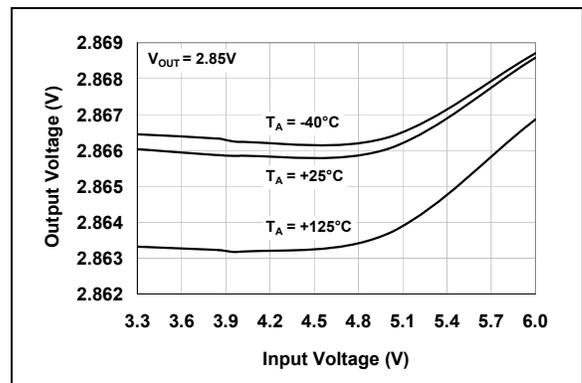


FIGURE 2-12: Output Voltage vs. Supply Voltage.

TC1017

Note: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$.

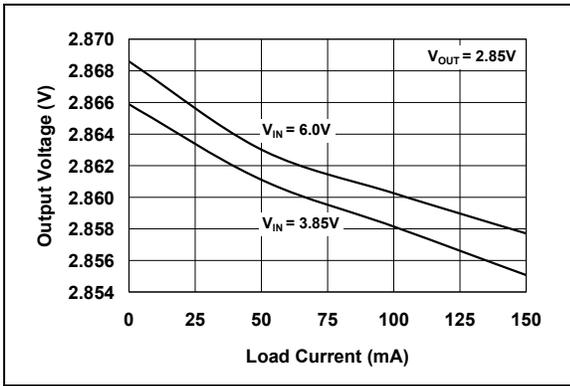


FIGURE 2-13: Output Voltage vs. Output Current.

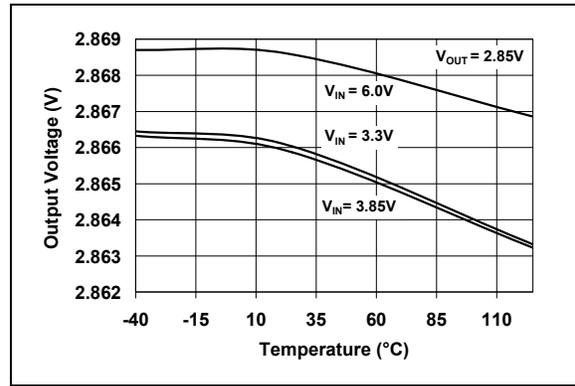


FIGURE 2-16: Output Voltage vs. Temperature.

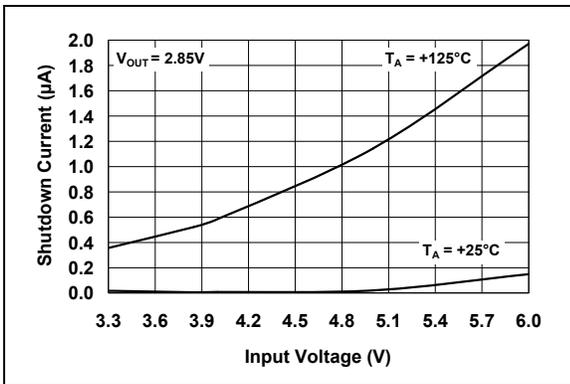


FIGURE 2-14: Shutdown Current vs. Input Voltage.

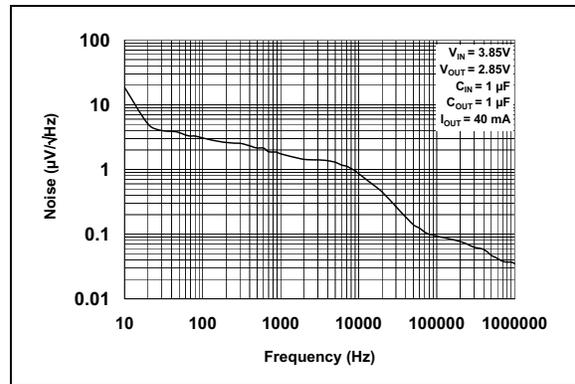


FIGURE 2-17: Output Noise vs. Frequency.

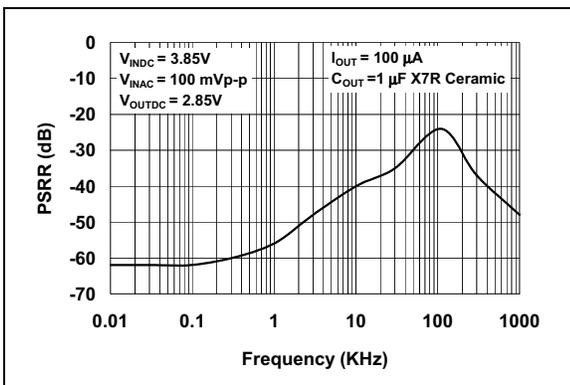


FIGURE 2-15: Power Supply Rejection Ratio vs. Frequency.

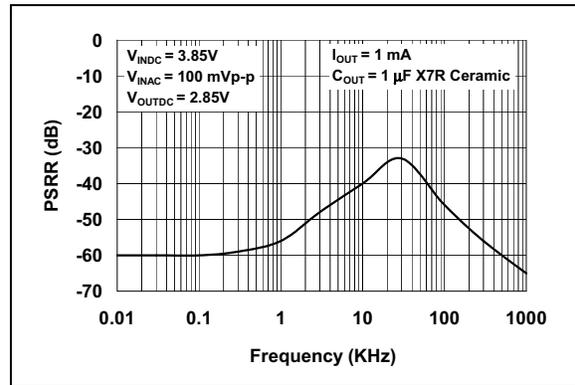


FIGURE 2-18: Power Supply Rejection Ratio vs. Frequency.

Note: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$.

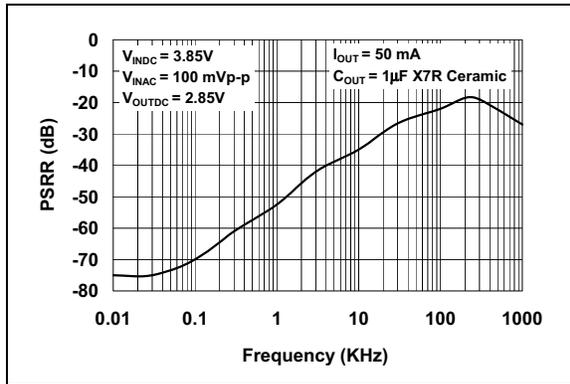


FIGURE 2-19: Power Supply Rejection Ratio vs. Frequency.

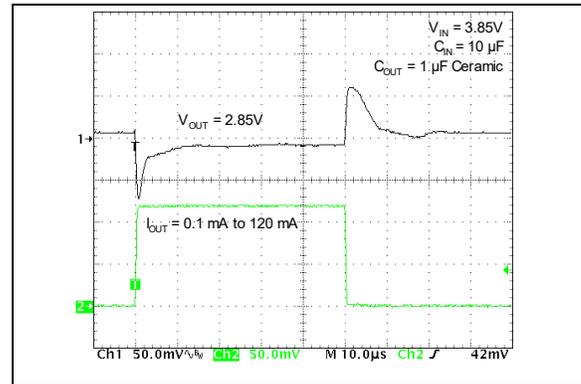


FIGURE 2-22: Load Transient Response.

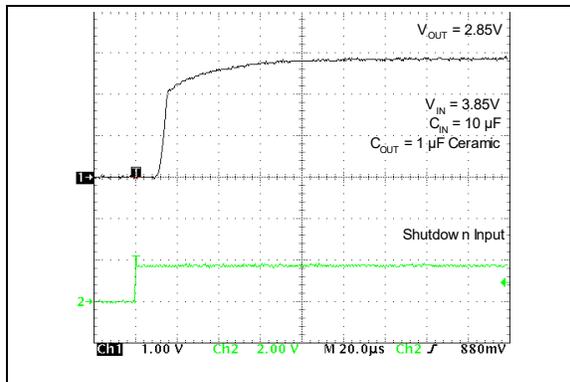


FIGURE 2-20: Wake-Up Response.

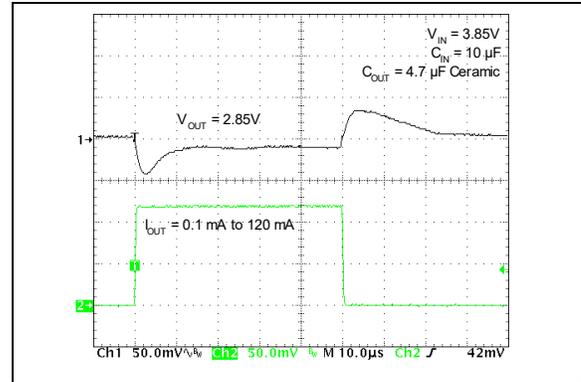


FIGURE 2-23: Load Transient Response.

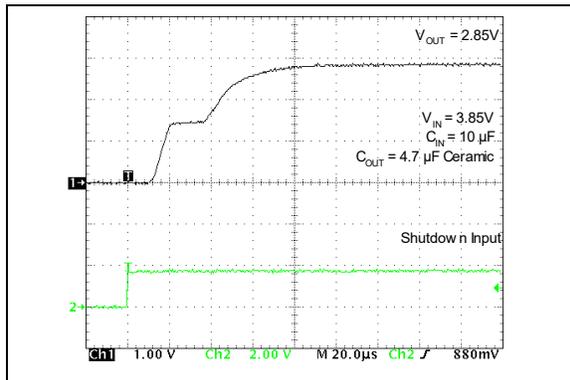


FIGURE 2-21: Wake-Up Response.

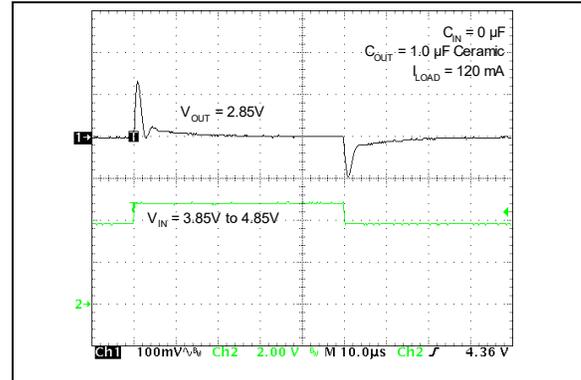


FIGURE 2-24: Line Transient Response.

TC1017

Note: Unless otherwise noted, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_L = 1.0 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$.

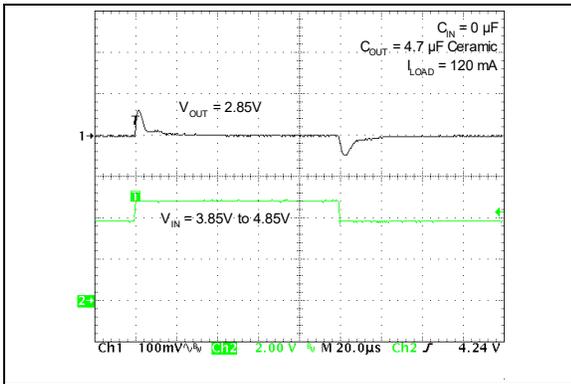


FIGURE 2-25: Line Transient Response.

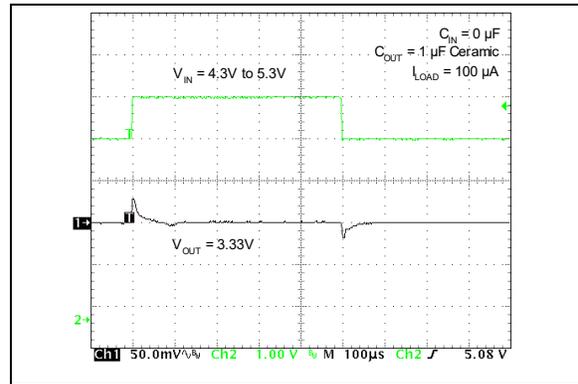


FIGURE 2-26: Line Transient Response.

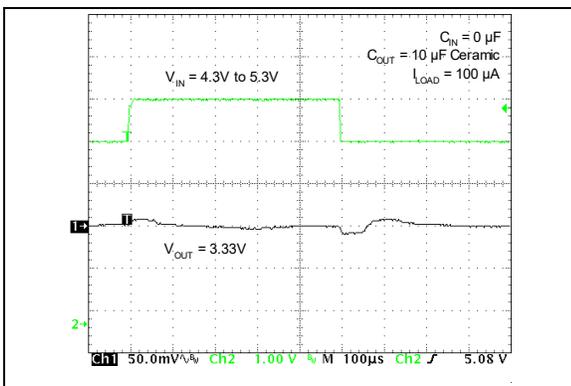


FIGURE 2-27: Line Transient Response.

3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in [Table 3-1](#).

TABLE 3-1: PIN FUNCTION TABLE

Pin No. 5-Pin SC-70	Pin No. 5-Pin SOT-23 5-Pin SC-70R	Symbol	Description
1	3	$\overline{\text{SHDN}}$	Shutdown Control Input
2	4	NC	No Connect
3	2	GND	Ground Terminal
4	5	V_{OUT}	Regulated Voltage Output
5	1	V_{IN}	Unregulated Supply Input

3.1 Shutdown Control Input ($\overline{\text{SHDN}}$)

The regulator is fully enabled when a logic-high is applied to $\overline{\text{SHDN}}$. The regulator enters shutdown when a logic-low is applied to this input. During shutdown, the output voltage falls to zero and the supply current is reduced to 0.05 μA (typ.)

3.2 Ground Terminal

For best performance, it is recommended that the ground pin be tied to a ground plane.

3.3 Regulated Voltage Output (V_{OUT})

Bypass the regulated voltage output to GND with a minimum capacitance of 1 μF . A ceramic bypass capacitor is recommended for best performance.

3.4 Unregulated Supply Input (V_{IN})

The minimum V_{IN} has to meet two conditions in order to ensure that the output maintains regulation: $V_{\text{IN}} \geq 2.7\text{V}$ and $V_{\text{IN}} \geq [(V_{\text{R}} + 2.5\%) + V_{\text{DROPOUT}}]$. The maximum V_{IN} should be less than or equal to 6V. Power dissipation may limit V_{IN} to a lower potential in order to maintain a junction temperature below 125°C. Refer to [Section 5.0 “Thermal Considerations”](#), for determining junction temperature.

It is recommended that V_{IN} be bypassed to GND with a ceramic capacitor.

TC1017

4.0 DETAILED DESCRIPTION

The TC1017 is a precision, fixed-output, linear voltage regulator. The internal linear pass element is a P-channel MOSFET. As with all P-channel CMOS LDOs, there is a body drain diode with the cathode connected to V_{IN} and the anode connected to V_{OUT} (Figure 4-1).

As is shown in Figure 4-1, the output voltage of the LDO is sensed and divided down internally to reduce external component count. The internal error amplifier has a fixed bandgap reference on the inverting input and the sensed output voltage on the noninverting input. The error amplifier output will pull the gate voltage down until the inputs of the error amplifier are equal to regulate the output voltage.

Output overload protection is implemented by sensing the current in the P-channel MOSFET. During a shorted or faulted load condition in which the output voltage falls to less than 0.5V, the output current is limited to a typical value of 120 mA. The current-limit protection helps prevent excessive current from damaging the Printed Circuit Board (PCB).

An internal thermal sensing device is used to monitor the junction temperature of the LDO. When the sensed temperature is over the set threshold of 160°C (typical), the P-channel MOSFET is turned off. When the P-channel is off, the power dissipation internal to the device is almost zero. The device cools until the junction

temperature is approximately 150°C and the P-channel is turned on. If the internal power dissipation is still high enough for the junction to rise to 160°C, it will again shut off and cool. The maximum operating junction temperature of the device is 125°C. Steady-state operation at or near the 160°C overtemperature point can lead to permanent damage of the device.

The output voltage V_{OUT} remains stable over the entire input operating voltage range (2.7V to 6.0V) and the entire load range (0 mA to 150 mA). The output voltage is sensed through an internal resistor divider and compared with a precision internal voltage reference. Several fixed-output voltages are available by changing the value of the internal resistor divider.

Figure 4-2 shows a typical application circuit. The regulator is enabled any time the shutdown input pin is at or above V_{IH} . It is shut down (disabled) any time the shutdown input pin is below V_{IL} . For applications where the SHDN feature is not used, tie the SHDN pin directly to the input supply voltage source. While in shutdown, the supply current decreases to 0.05 μ A (typical) and the P-channel MOSFET is turned off.

As shown in Figure 4-2, batteries have internal source impedance. An input capacitor is used to lower the input impedance of the LDO. In some applications, high input impedance can cause the LDO to become unstable. Adding more input capacitance can compensate for this.

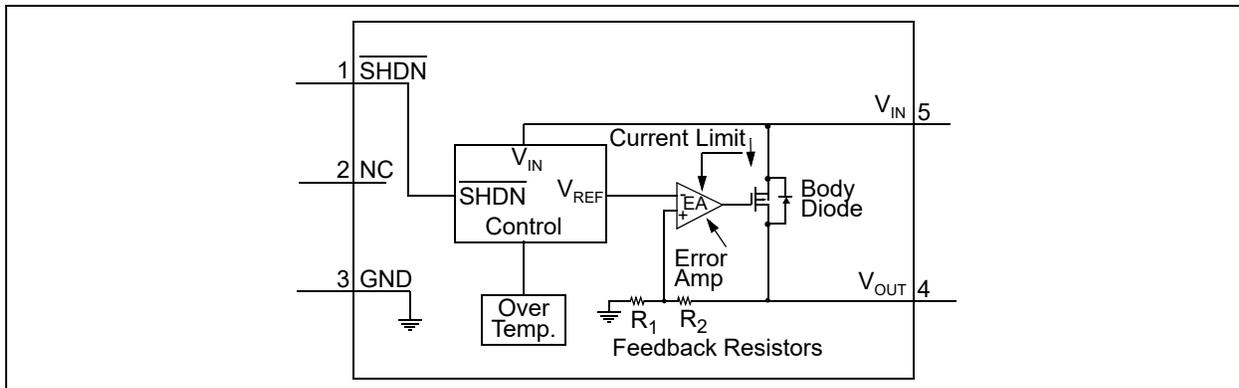


FIGURE 4-1: TC1017 Block Diagram (5-Pin SC-70 Pinout).

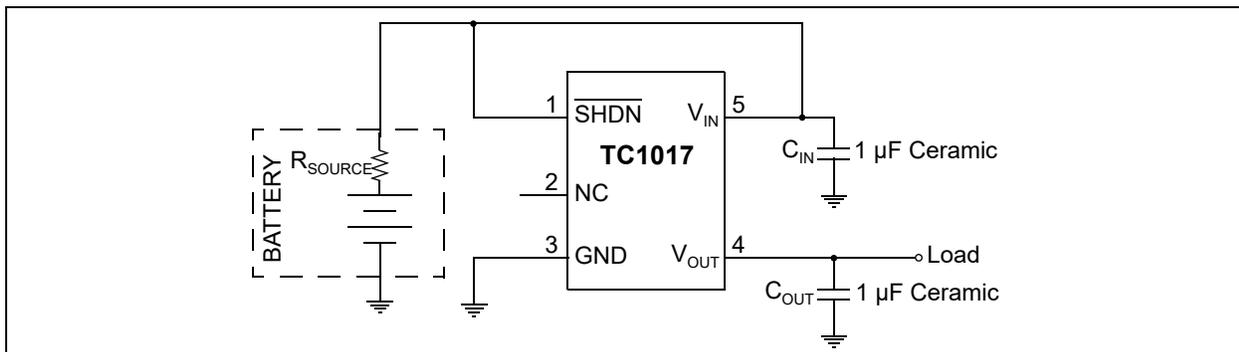


FIGURE 4-2: Typical Application Circuit (5-Pin SC-70 Pinout).

4.1 Input Capacitor

Low input source impedance is necessary for the LDO to operate properly. When operating from batteries, or in applications with long lead length ($> 10''$) between the input source and the LDO, some input capacitance is required. A minimum of $0.1 \mu\text{F}$ is recommended for most applications and the capacitor should be placed as close to the input of the LDO as is practical. Larger input capacitors will help reduce the input impedance and further reduce any high-frequency noise on the input and output of the LDO.

4.2 Output Capacitor

A minimum output capacitance of $1 \mu\text{F}$ for the TC1017 is required for stability. The Equivalent Series Resistance (ESR) requirements on the output capacitor are between 0 and 2 ohms. The output capacitor should be located as close to the LDO output as is practical. Ceramic materials X7R and X5R have low temperature coefficients and are well within the acceptable ESR range required. A typical $1 \mu\text{F}$ X5R 0805 capacitor has an ESR of 50 milli-ohms. Larger output capacitors can be used with the TC1017 to improve dynamic behavior and input ripple-rejection performance.

Ceramic, aluminum electrolytic or tantalum capacitor types can be used. Since many aluminum electrolytic capacitors freeze at approximately -30°C , ceramic or solid tantalums are recommended for applications operating below -25°C . When operating from sources other than batteries, supply-noise rejection and transient response can be improved by increasing the value of the input and output capacitors and employing passive filtering techniques.

4.3 Turn-On Response

The turn-on response is defined as two separate response categories, wake-up time (t_{WK}) and settling time (t_{S}).

The TC1017 has a fast wake-up time ($10 \mu\text{sec}$, typical) when released from shutdown. See Figure 4-3 for the wake-up time designated as t_{WK} . The wake-up time is defined as the time it takes for the output to rise to 2% of the V_{OUT} value after being released from shutdown.

The total turn-on response is defined as the settling time (t_{S}) (see Figure 4-3). Settling time (inclusive with t_{WK}) is defined as the condition when the output is within 98% of its fully-enabled value ($32 \mu\text{sec}$, typical) when released from shutdown. The settling time of the output voltage is dependent on load conditions and output capacitance on V_{OUT} (RC response).

The table below demonstrates the typical turn-on response timing for different input voltage power-up frequencies: $V_{\text{OUT}} = 2.85\text{V}$, $V_{\text{IN}} = 5.0\text{V}$, $I_{\text{OUT}} = 60 \text{mA}$ and $C_{\text{OUT}} = 1 \mu\text{F}$.

Frequency	Typical (t_{WK})	Typical (t_{S})
1000 Hz	5.3 μsec	14 μsec
500 Hz	5.9 μsec	16 μsec
100 Hz	9.8 μsec	32 μsec
50 Hz	14.5 μsec	52 μsec
10 Hz	17.2 μsec	77 μsec

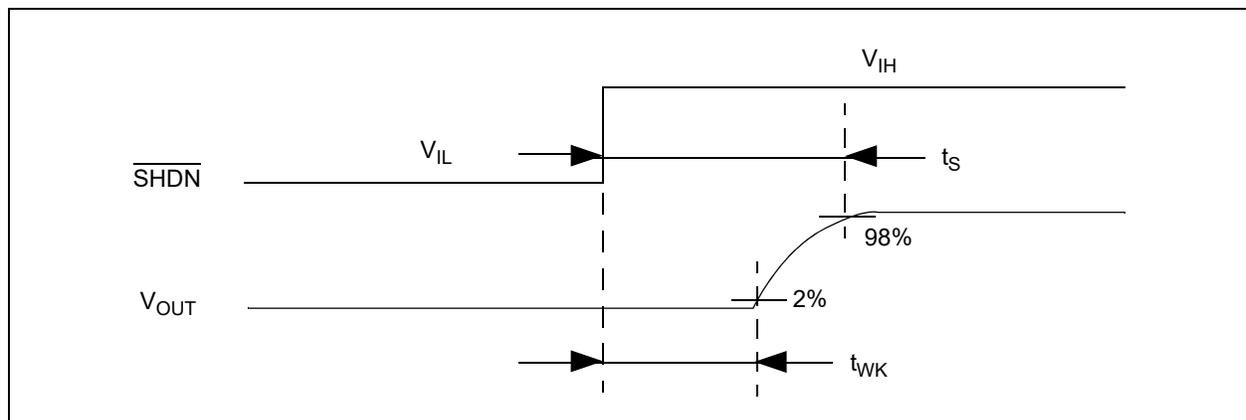


FIGURE 4-3: Wake-Up Time from Shutdown.

5.0 THERMAL CONSIDERATIONS

5.1 Thermal Shutdown

Integrated thermal protection circuitry shuts the regulator off when the die temperature exceeds approximately 160°C. The regulator remains off until the die temperature drops to approximately 150°C.

5.2 Power Dissipation: SC-70

The TC1017 is available in the SC-70 package. The thermal resistance for the SC-70 package is approximately 450°C/W when the copper area used in the PCB layout is similar to the JEDEC J51-7 high thermal conductivity standard or semi-G42-88 standard. For applications with a larger or thicker copper area, the thermal resistance can be lowered. See AN792, "A Method to Determine How Much Power a SOT-23 Can Dissipate in an Application" (DS00792), for a method to determine the thermal resistance for a particular application.

The TC1017 power dissipation capability is dependant upon several variables: input voltage, output voltage, load current, ambient temperature and maximum junction temperature. The absolute maximum steady-state junction temperature is rated at +125°C. The power dissipation within the device is equal to:

EQUATION 5-1:

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD} + V_{IN} \times I_{GND}$$

The $V_{IN} \times I_{GND}$ term is typically very small when compared to the $(V_{IN} - V_{OUT}) \times I_{LOAD}$ term, simplifying the power dissipation within the LDO to be:

EQUATION 5-2:

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD}$$

To determine the maximum power dissipation capability, the following equation is used:

EQUATION 5-3:

$$P_{D_{MAX}} = \frac{(T_{J_{MAX}} - T_{A_{MAX}})}{R\theta_{JA}}$$

Where:

$T_{J_{MAX}}$ = the maximum junction temperature allowed

$T_{A_{MAX}}$ = the maximum ambient temperature

$R\theta_{JA}$ = the thermal resistance from junction to air

Given the following example:

$$V_{IN} = 3.0V \text{ to } 4.1V$$

$$V_{OUT} = 2.85V \pm 2.5\%$$

$$I_{LOAD} = 120 \text{ mA (output current)}$$

$$T_A = 55^\circ\text{C (max. desired ambient)}$$

Find:

1. Internal power dissipation:

$$\begin{aligned} P_{D_{MAX}} &= (V_{IN_{MAX}} - V_{OUT_{MIN}}) \times I_{LOAD} \\ &= (4.1V - 2.85 \times (0.975)) \times 120mA \\ &= 158.5mW \end{aligned}$$

2. Maximum allowable ambient temperature:

$$\begin{aligned} T_{A_{MAX}} &= T_{J_{MAX}} - P_{D_{MAX}} \times R\theta_{JA} \\ &= (125^\circ\text{C} - 158.5mW \times 450^\circ\text{C/W}) \\ &= (125^\circ\text{C} - 71^\circ\text{C}) \\ &= 54^\circ\text{C} \end{aligned}$$

3. Maximum allowable power dissipation at desired ambient:

$$\begin{aligned} P_D &= \frac{T_{J_{MAX}} - T_A}{R\theta_{JA}} \\ &= \frac{125^\circ\text{C} - 55^\circ\text{C}}{450^\circ\text{C/W}} \\ &= 155mW \end{aligned}$$

In this example, the TC1017 dissipates approximately 158.5 mW and the junction temperature is raised 71°C over the ambient. The absolute maximum power dissipation is 155 mW when given a maximum ambient is 55°C.

Input voltage, output voltage or load current limits can also be determined by substituting known values in the power dissipation equations.

Figure 5-1 and Figure 5-2 depict typical maximum power dissipation versus ambient temperature, as well as typical maximum current versus ambient temperature, with a 1V input voltage to output voltage differential, respectively.

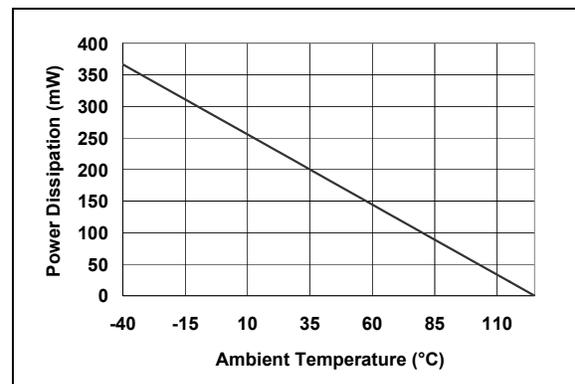


FIGURE 5-1: Power Dissipation vs. Ambient Temperature (SC-70 package).

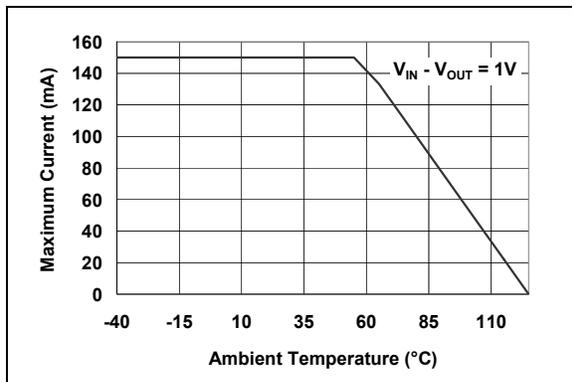


FIGURE 5-2: Maximum Current vs. Ambient Temperature (SC-70 package).

5.3 Power Dissipation: SOT-23

The TC1017 is also available in a SOT-23 package for improved thermal performance. The thermal resistance for the SOT-23 package is approximately 255°C/W when the copper area used in the printed circuit board layout is similar to the JEDEC J51-7 low thermal conductivity standard or semi-G42-88 standard. For applications with a larger or thicker copper area, the thermal resistance can be lowered. See AN792, “A Method to Determine How Much Power a SOT-23 Can Dissipate in an Application” (DS00792), for a method to determine the thermal resistance for a particular application.

The TC1017 power dissipation capability is dependant upon several variables: input voltage, output voltage, load current, ambient temperature and maximum junction temperature. The absolute maximum steady-state junction temperature is rated at +125°C. The power dissipation within the device is equal to:

EQUATION 5-4:

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD} + V_{IN} \times I_{GND}$$

The $V_{IN} \times I_{GND}$ term is typically very small when compared to the $(V_{IN} - V_{OUT}) \times I_{LOAD}$ term, simplifying the power dissipation within the LDO to be:

EQUATION 5-5:

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD}$$

To determine the maximum power dissipation capability, the following equation is used:

EQUATION 5-6:

$$P_{D_{MAX}} = \frac{(T_{J_{MAX}} - T_{A_{MAX}})}{R\theta_{JA}}$$

Where:

$T_{J_{MAX}}$ = the maximum junction temperature allowed

$T_{A_{MAX}}$ = the maximum ambient temperature

$R\theta_{JA}$ = the thermal resistance from junction to air

Given the following example:

$$V_{IN} = 3.0V \text{ to } 4.1V$$

$$V_{OUT} = 2.85V \pm 2.5\%$$

$$I_{LOAD} = 120 \text{ mA (output current)}$$

$$T_A = +85^\circ\text{C (max. desired ambient)}$$

Find:

1. Internal power dissipation:

$$\begin{aligned} P_{D_{MAX}} &= (V_{IN_{MAX}} - V_{OUT_{MIN}}) \times I_{LOAD} \\ &= (4.1V - 2.85 \times (0.975)) \times 120mA \\ &= 158.5mW \end{aligned}$$

2. Maximum allowable ambient temperature:

$$\begin{aligned} T_{A_{MAX}} &= T_{J_{MAX}} - P_{D_{MAX}} \times R\theta_{JA} \\ &= (125^\circ\text{C} - 158.5mW \times 255^\circ\text{C/W}) \\ &= (125^\circ\text{C} - 40.5^\circ\text{C}) \\ &= 84.5^\circ\text{C} \end{aligned}$$

3. Maximum allowable power dissipation at desired ambient:

$$\begin{aligned} P_D &= \frac{T_{J_{MAX}} - T_A}{R\theta_{JA}} \\ &= \frac{125^\circ\text{C} - 85^\circ\text{C}}{255^\circ\text{C/W}} \\ &= 157mW \end{aligned}$$

In this example, the TC1017 dissipates approximately 158.5 mW and the junction temperature is raised 40.5°C over the ambient. The absolute maximum power dissipation is 157 mW when given a maximum ambient temperature of +85°C.

Input voltage, output voltage or load current limits can also be determined by substituting known values in the power dissipation equations.

Figure 5-3 and Figure 5-4 depict typical maximum power dissipation versus ambient temperature, as well as typical maximum current versus ambient temperature with a 1V input voltage to output voltage differential, respectively.

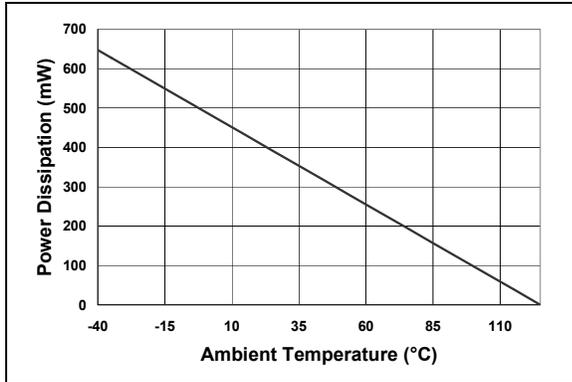


FIGURE 5-3: Power Dissipation vs. Ambient Temperature (SOT-23 Package).

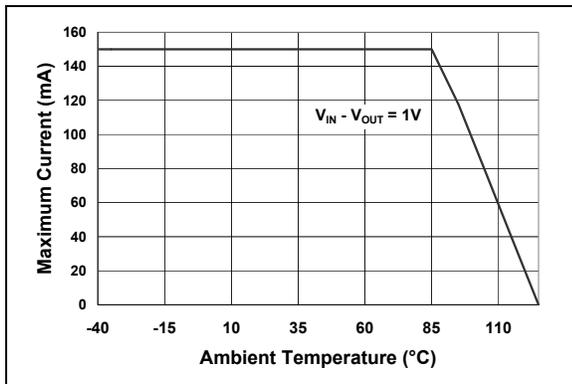


FIGURE 5-4: Maximum Current vs. Ambient Temperature (SOT-23 Package).

5.4 Layout Considerations

The primary path for heat conduction out of the SC-70/SOT-23 package is through the package leads. Using heavy, wide traces at the pads of the device will facilitate the removal of the heat within the package, thus lowering the thermal resistance $R_{\theta JA}$. By lowering the thermal resistance, the maximum internal power dissipation capability of the package is increased.

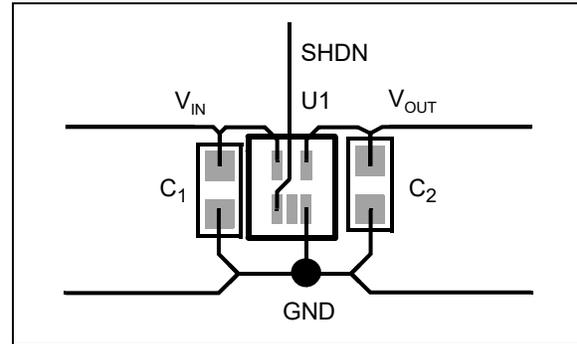
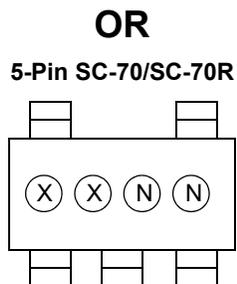
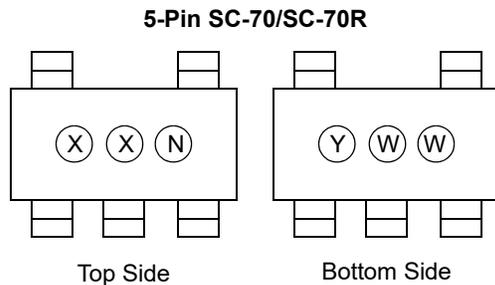


FIGURE 5-5: SC-70 Package Suggested Layout.

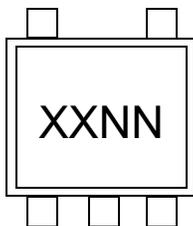
6.0 PACKAGE INFORMATION

6.1 Package Marking Information



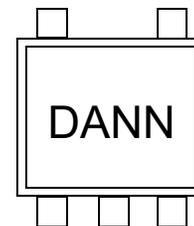
Part Number	TC1017 Pinout Code	TC1017R Pinout Code
TC1017 – 1.8VLT	CE	CU
TC1017 – 1.85VLT	CQ	DF
TC1017 – 1.9VLT	CB	
TC1017 – 2.5VLT	CR	CV
TC1017 – 2.6VLT	CF	CW
TC1017 – 2.7VLT	CG	CX
TC1017 – 2.8VLT	CH	CY
TC1017 – 2.85VLT	CJ	CZ
TC1017 – 2.9VLT	CK	DA
TC1017 – 3.0VLT	CL	DB
TC1017 – 3.2VLT	CC	DC
TC1017 – 3.3VLT	CM	DD
TC1017 – 4.0VLT	CP	DE

5-Lead SOT-23



Part Number	Code
TC1017 – 1.8VCT	DA
TC1017 – 1.85VCT	DK
TC1017 – 2.6VCT	DB
TC1017 – 2.7VCT	DC
TC1017 – 2.8VCT	DD
TC1017 – 2.85VCT	DE
TC1017 – 2.9VCT	DF
TC1017 – 3.0VCT	DG
TC1017 – 3.3VCT	DH
TC1017 – 4.0VCT	DJ

Example:



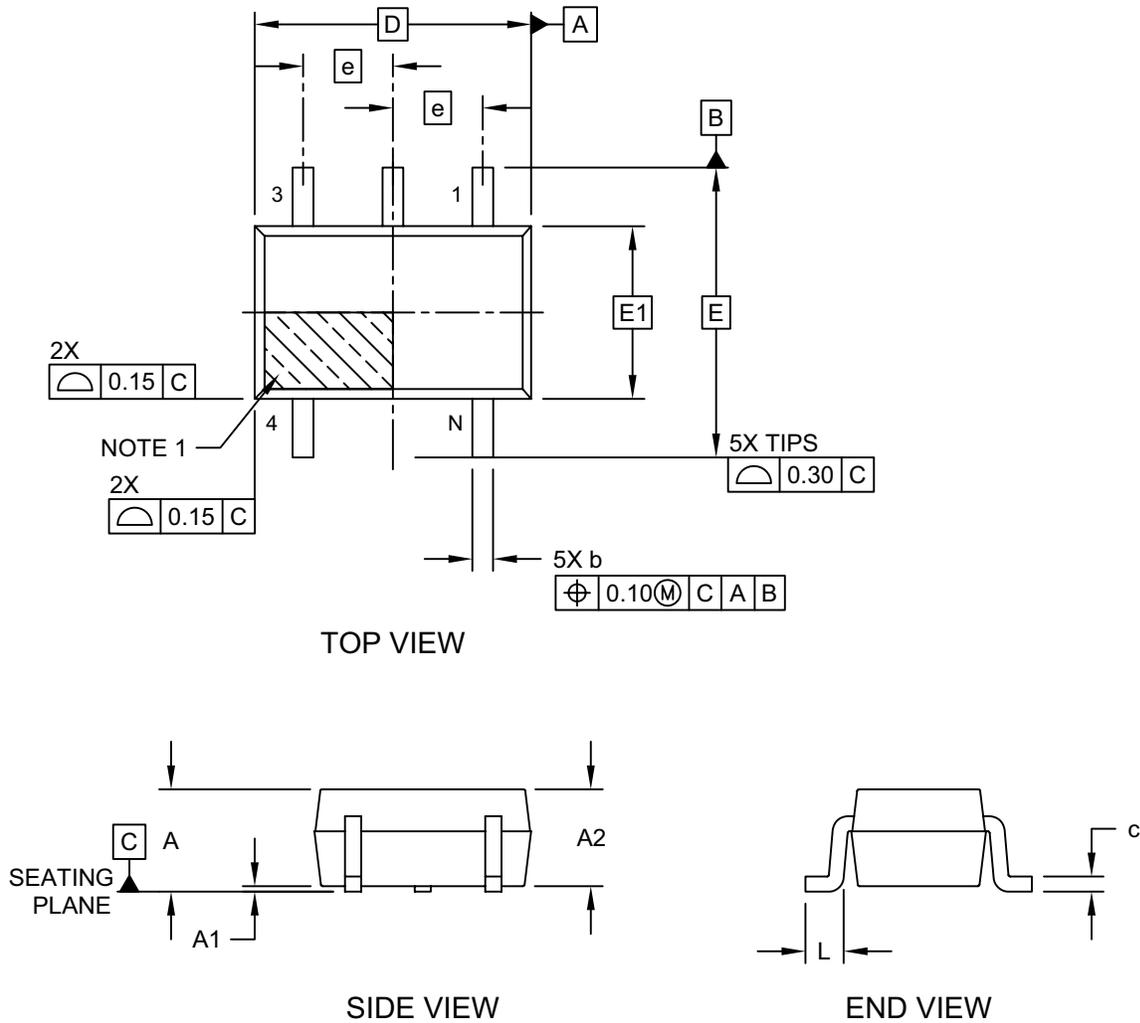
Legend:	XX...X	Customer-specific information*
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

TC1017

5-Lead Plastic Small Outline Transistor (LT) [SC70]

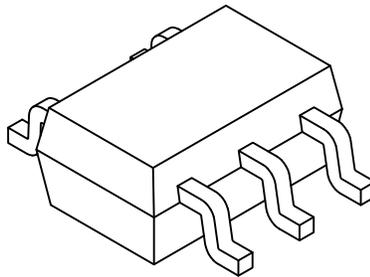
Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Microchip Technology Drawing C04-061-LT Rev E Sheet 1 of 2

5-Lead Plastic Small Outline Transistor (LT) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Number of Pins	N	5		
Pitch	e	0.65 BSC		
Overall Height	A	0.80	-	1.10
Standoff	A1	0.00	-	0.10
Molded Package Thickness	A2	0.80	-	1.00
Overall Length	D	2.00 BSC		
Overall Width	E	2.10 BSC		
Molded Package Width	E1	1.25 BSC		
Terminal Width	b	0.15	-	0.40
Terminal Length	L	0.10	0.20	0.46
Lead Thickness	c	0.08	-	0.26

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
3. Dimensioning and tolerancing per ASME Y14.5M

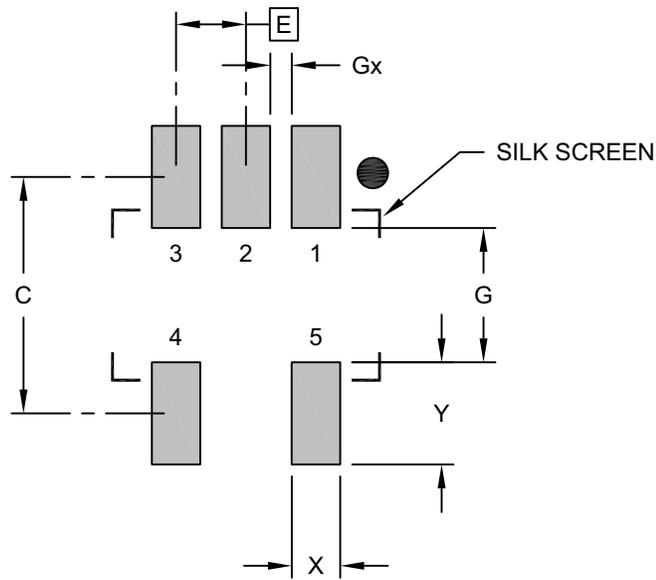
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-061-LT Rev E Sheet 2 of 2

5-Lead Plastic Small Outline Transistor (LT) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Contact Pitch	E	0.65 BSC		
Contact Pad Spacing	C		2.20	
Contact Pad Width	X			0.45
Contact Pad Length	Y			0.95
Distance Between Pads	G	1.25		
Distance Between Pads	Gx	0.20		

Notes:

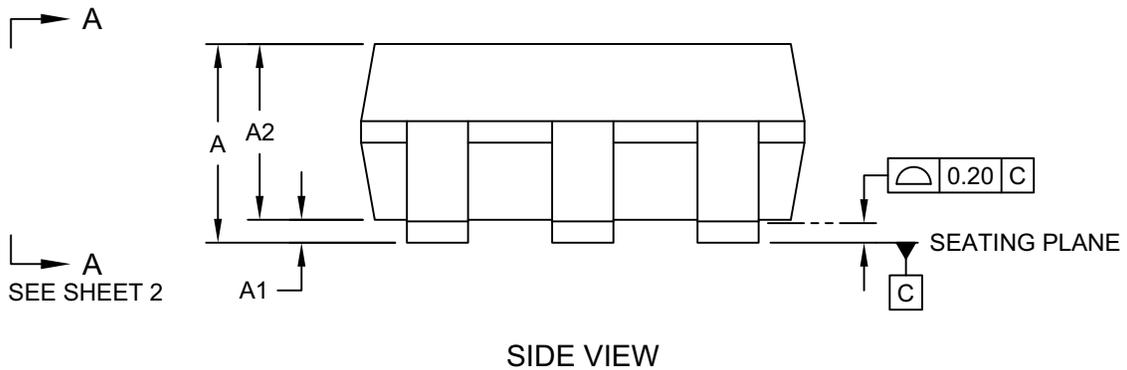
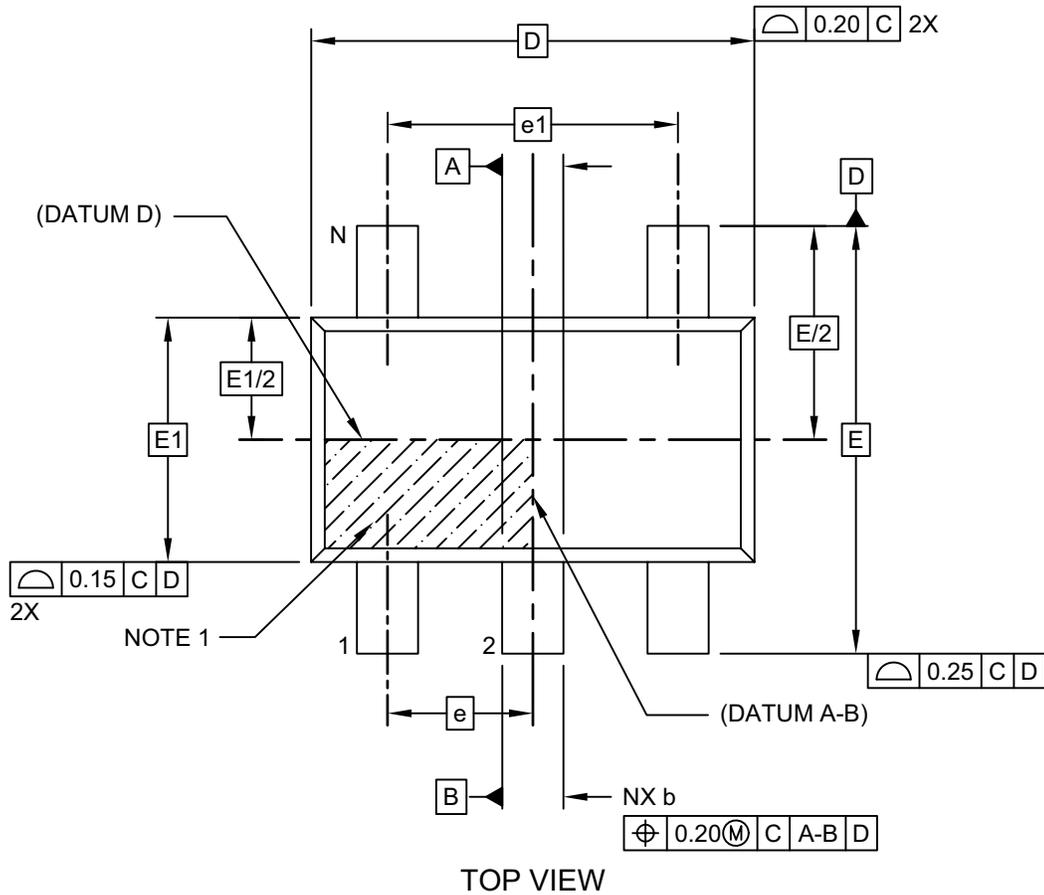
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2061-LT Rev E

5-Lead Plastic Small Outline Transistor (C7X) [SOT-23]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>

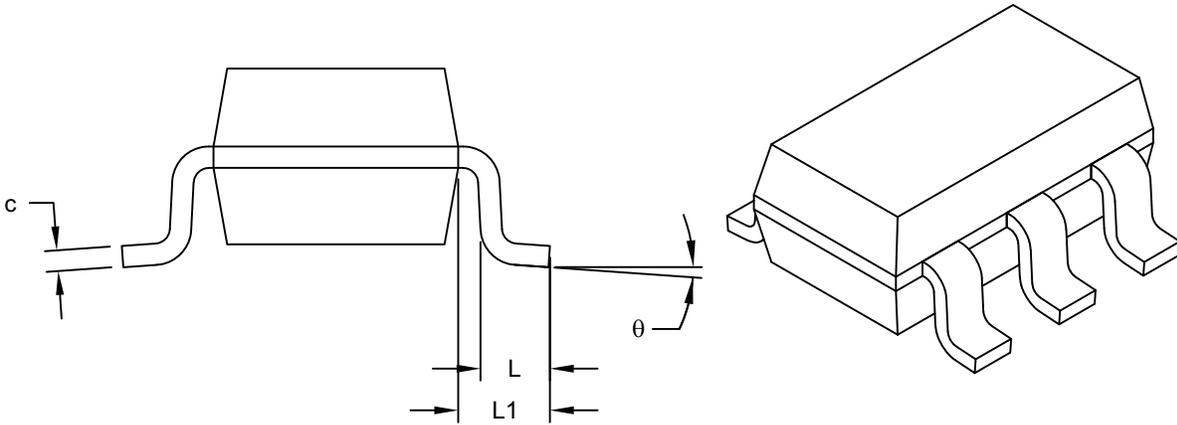


Microchip Technology Drawing C04-091-C7X Rev H Sheet 1 of 2

TC1017

5-Lead Plastic Small Outline Transistor (C7X) [SOT-23]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



VIEW A-A
SHEET 1

Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Number of Pins	N	5		
Pitch	e	0.95 BSC		
Outside lead pitch	e1	1.90 BSC		
Overall Height	A	0.90	-	1.45
Molded Package Thickness	A2	0.89	-	1.30
Standoff	A1	-	-	0.15
Overall Width	E	2.80 BSC		
Molded Package Width	E1	1.60 BSC		
Overall Length	D	2.90 BSC		
Foot Length	L	0.30	-	0.60
Footprint	L1	0.60 REF		
Foot Angle	θ	0°	-	10°
Lead Thickness	c	0.08	-	0.26
Lead Width	b	0.20	-	0.51

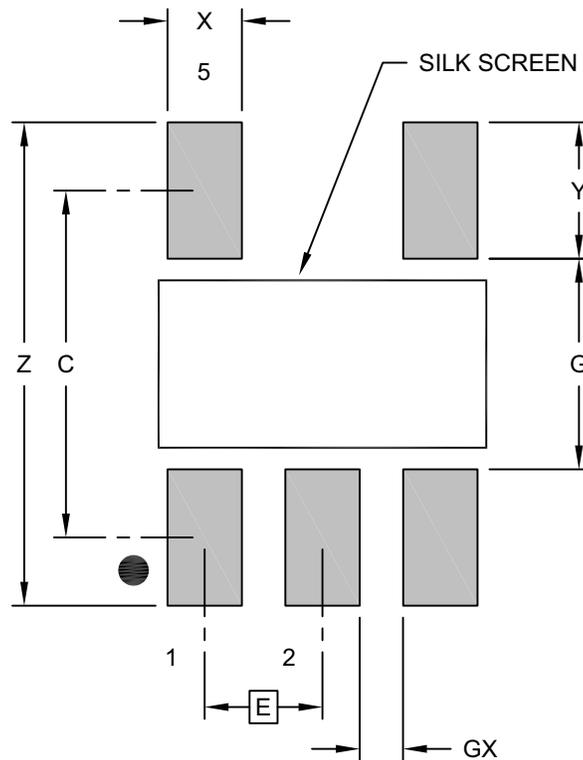
Notes:

- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25mm per side.
- Dimensioning and tolerancing per ASME Y14.5M
BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-091-C7X Rev H Sheet 2 of 2

5-Lead Plastic Small Outline Transistor (C7X) [SOT-23]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Contact Pitch	E	0.95 BSC		
Contact Pad Spacing	C		2.80	
Contact Pad Width (X5)	X			0.60
Contact Pad Length (X5)	Y			1.10
Distance Between Pads	G	1.70		
Distance Between Pads	GX	0.35		
Overall Width	Z			3.90

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2091-C7X Rev H

TC1017

NOTES:

APPENDIX A: REVISION HISTORY

Revision G (August 2023)

- Added information about the Automotive Qualification status of the device in [Section , Features](#).
- Added ESD values to [Section 1.0, Electrical Characteristics](#).
- Updated package drawings in [Section 6.0, Package Information](#).
- Updated [Section , Product Identification System](#), with Automotive Qualified devices.

Revision F (April 2013)

The following is the list of modifications:

- Updated the information for the Maximum Output Current parameter in the [Electrical Characteristics](#) table.

Revision E (January 2013)

- Added a note to each package outline drawing.

Revision D (February 2005)

- Undocumented changes.

TC1017

NOTES:

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<u>PART NO.</u>	<u>X.XX</u>	<u>X</u>	<u>XXXX</u>	<u>-XXX</u>	Examples:
Device	Voltage Options	Temperature Range	Package	Qualification	
Device:	TC1017: 150 mA Tiny CMOS LDO with Shutdown TC1017R: 150 mA Tiny CMOS LDO with Shutdown (SC-70 only)				a) TC1017-1.8VLTTR: 150 mA, Tiny CMOS LDO with Shutdown, SC-70 package
Voltage Options:* (Standard)	1.8V 1.85V 2.5V SC-70 only 2.6V 2.7V 2.8V 2.85V 2.9V 3.0V 3.2V SC-70 only 3.3V 4.0V				b) TC1017R-1.8VLTTR: 150mA, Tiny CMOS LDO with Shutdown, SC-70 package
	* Other voltage options available. Please contact your local Microchip sales office for details.				c) TC1017-2.6VCTTR: 150 mA, Tiny CMOS LDO with Shutdown, SOT-23 package
Temperature Range:	V = -40°C to +125°C				d) TC1017-2.7VLTTR: 150 mA, Tiny CMOS LDO with Shutdown, SC-70 package
Package:	LTTR = 5-pin SC-70 (Tape and Reel) CTTR = 5-pin SOT-23 (Tape and Reel)				e) TC1017-2.8VCTTR: 150 mA, Tiny CMOS LDO with Shutdown, SOT-23 package
Qualification**:	<Blank> = Standard Part VAO = AEC-Q100 Automotive Qualified ** Contact your local Microchip sales office to request automotive qualified part variants.				f) TC1017-2.85VLTTR: 150 mA, Tiny CMOS LDO with Shutdown, SC-70 package
					g) TC1017-2.9VCTTR: 150 mA, Tiny CMOS LDO with Shutdown, SOT-23 package
					h) TC1017-3.0VLTTR: 150 mA, Tiny CMOS LDO with Shutdown, SC-70 package
					i) TC1017-3.3VCTTR: 150 mA, Tiny CMOS LDO with Shutdown, SOT-23 package
					j) TC1017-3.3VCTTR-VAO**: 150 mA, Tiny CMO LDO with Shutdown, SOT-23 package, AEC-Q100Automotive Qualified
					k) TC1017-4.0VLTTR: 150 mA, Tiny CMOS LDO with Shutdown, SC-70 package
					Note: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.
					** Example of automotive part variant that can be set up.

TC1017

NOTES:

Note the following details of the code protection feature on Microchip products:

- Microchip products meet the specifications contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is secure when used in the intended manner, within operating specifications, and under normal conditions.
- Microchip values and aggressively protects its intellectual property rights. Attempts to breach the code protection features of Microchip product is strictly prohibited and may violate the Digital Millennium Copyright Act.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of its code. Code protection does not mean that we are guaranteeing the product is "unbreakable" Code protection is constantly evolving. Microchip is committed to continuously improving the code protection features of our products.

This publication and the information herein may be used only with Microchip products, including to design, test, and integrate Microchip products with your application. Use of this information in any other manner violates these terms. Information regarding device applications is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. Contact your local Microchip sales office for additional support or, obtain additional support at <https://www.microchip.com/en-us/support/design-help/client-support-services>.

THIS INFORMATION IS PROVIDED BY MICROCHIP "AS IS". MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION INCLUDING BUT NOT LIMITED TO ANY IMPLIED WARRANTIES OF NON-INFRINGEMENT, MERCHANTABILITY, AND FITNESS FOR A PARTICULAR PURPOSE, OR WARRANTIES RELATED TO ITS CONDITION, QUALITY, OR PERFORMANCE.

IN NO EVENT WILL MICROCHIP BE LIABLE FOR ANY INDIRECT, SPECIAL, PUNITIVE, INCIDENTAL, OR CONSEQUENTIAL LOSS, DAMAGE, COST, OR EXPENSE OF ANY KIND WHATSOEVER RELATED TO THE INFORMATION OR ITS USE, HOWEVER CAUSED, EVEN IF MICROCHIP HAS BEEN ADVISED OF THE POSSIBILITY OR THE DAMAGES ARE FORESEEABLE. TO THE FULLEST EXTENT ALLOWED BY LAW, MICROCHIP'S TOTAL LIABILITY ON ALL CLAIMS IN ANY WAY RELATED TO THE INFORMATION OR ITS USE WILL NOT EXCEED THE AMOUNT OF FEES, IF ANY, THAT YOU HAVE PAID DIRECTLY TO MICROCHIP FOR THE INFORMATION.

Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

For information regarding Microchip's Quality Management Systems, please visit www.microchip.com/quality.

Trademarks

The Microchip name and logo, the Microchip logo, Adaptec, AVR, AVR logo, AVR Freaks, BesTime, BitCloud, CryptoMemory, CryptoRF, dsPIC, flexPWR, HELDO, IGLOO, JukeBlox, KeeLoq, Klear, LANCheck, LinkMD, maXStylus, maXTouch, MediaLB, megaAVR, Microsemi, Microsemi logo, MOST, MOST logo, MPLAB, OptoLyzer, PIC, picoPower, PICSTART, PIC32 logo, PolarFire, Prochip Designer, QTouch, SAM-BA, SenGenuity, SpyNIC, SST, SST Logo, SuperFlash, Symmetricom, SyncServer, Tachyon, TimeSource, tinyAVR, UNI/O, Vectron, and XMEGA are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AgileSwitch, APT, ClockWorks, The Embedded Control Solutions Company, EtherSynch, Flashtec, Hyper Speed Control, HyperLight Load, Libero, motorBench, mTouch, Powermite 3, Precision Edge, ProASIC, ProASIC Plus, ProASIC Plus logo, Quiet-Wire, SmartFusion, SyncWorld, Temux, TimeCesium, TimeHub, TimePictra, TimeProvider, TrueTime, and ZL are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Adjacent Key Suppression, AKS, Analog-for-the-Digital Age, Any Capacitor, AnyIn, AnyOut, Augmented Switching, BlueSky, BodyCom, Clockstudio, CodeGuard, CryptoAuthentication, CryptoAutomotive, CryptoCompanion, CryptoController, dsPICDEM, dsPICDEM.net, Dynamic Average Matching, DAM, ECAN, Espresso T1S, EtherGREEN, GridTime, IdealBridge, In-Circuit Serial Programming, ICSP, INICnet, Intelligent Paralleling, IntelliMOS, Inter-Chip Connectivity, JitterBlocker, Knob-on-Display, KoD, maxCrypto, maxView, memBrain, Mindi, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, PowerSmart, PureSilicon, QMatrix, REAL ICE, Ripple Blocker, RTAX, RTG4, SAM-ICE, Serial Quad I/O, simpleMAP, SimpliPHY, SmartBuffer, SmartHLS, SMART-I.S., storClad, SQI, SuperSwitcher, SuperSwitcher II, Switchtec, SynchroPHY, Total Endurance, Trusted Time, TSHARC, USBCheck, VariSense, VectorBlox, VeriPHY, ViewSpan, WiperLock, XpressConnect, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

The Adaptec logo, Frequency on Demand, Silicon Storage Technology, and Symmcom are registered trademarks of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2005-2023, Microchip Technology Incorporated and its subsidiaries.

All Rights Reserved.

ISBN: 978-1-6683-2958-0



MICROCHIP

Worldwide Sales and Service

AMERICAS

Corporate Office

2355 West Chandler Blvd.

Chandler, AZ 85224-6199

Tel: 480-792-7200

Fax: 480-792-7277

Technical Support:

<http://www.microchip.com/support>

Web Address:

www.microchip.com

Atlanta

Duluth, GA

Tel: 678-957-9614

Fax: 678-957-1455

Austin, TX

Tel: 512-257-3370

Boston

Westborough, MA

Tel: 774-760-0087

Fax: 774-760-0088

Chicago

Itasca, IL

Tel: 630-285-0071

Fax: 630-285-0075

Dallas

Addison, TX

Tel: 972-818-7423

Fax: 972-818-2924

Detroit

Novi, MI

Tel: 248-848-4000

Houston, TX

Tel: 281-894-5983

Indianapolis

Noblesville, IN

Tel: 317-773-8323

Fax: 317-773-5453

Tel: 317-536-2380

Los Angeles

Mission Viejo, CA

Tel: 949-462-9523

Fax: 949-462-9608

Tel: 951-273-7800

Raleigh, NC

Tel: 919-844-7510

New York, NY

Tel: 631-435-6000

San Jose, CA

Tel: 408-735-9110

Tel: 408-436-4270

Canada - Toronto

Tel: 905-695-1980

Fax: 905-695-2078

ASIA/PACIFIC

Australia - Sydney

Tel: 61-2-9868-6733

China - Beijing

Tel: 86-10-8569-7000

China - Chengdu

Tel: 86-28-8665-5511

China - Chongqing

Tel: 86-23-8980-9588

China - Dongguan

Tel: 86-769-8702-9880

China - Guangzhou

Tel: 86-20-8755-8029

China - Hangzhou

Tel: 86-571-8792-8115

China - Hong Kong SAR

Tel: 852-2943-5100

China - Nanjing

Tel: 86-25-8473-2460

China - Qingdao

Tel: 86-532-8502-7355

China - Shanghai

Tel: 86-21-3326-8000

China - Shenyang

Tel: 86-24-2334-2829

China - Shenzhen

Tel: 86-755-8864-2200

China - Suzhou

Tel: 86-186-6233-1526

China - Wuhan

Tel: 86-27-5980-5300

China - Xian

Tel: 86-29-8833-7252

China - Xiamen

Tel: 86-592-2388138

China - Zhuhai

Tel: 86-756-3210040

ASIA/PACIFIC

India - Bangalore

Tel: 91-80-3090-4444

India - New Delhi

Tel: 91-11-4160-8631

India - Pune

Tel: 91-20-4121-0141

Japan - Osaka

Tel: 81-6-6152-7160

Japan - Tokyo

Tel: 81-3-6880-3770

Korea - Daegu

Tel: 82-53-744-4301

Korea - Seoul

Tel: 82-2-554-7200

Malaysia - Kuala Lumpur

Tel: 60-3-7651-7906

Malaysia - Penang

Tel: 60-4-227-8870

Philippines - Manila

Tel: 63-2-634-9065

Singapore

Tel: 65-6334-8870

Taiwan - Hsin Chu

Tel: 886-3-577-8366

Taiwan - Kaohsiung

Tel: 886-7-213-7830

Taiwan - Taipei

Tel: 886-2-2508-8600

Thailand - Bangkok

Tel: 66-2-694-1351

Vietnam - Ho Chi Minh

Tel: 84-28-5448-2100

EUROPE

Austria - Wels

Tel: 43-7242-2244-39

Fax: 43-7242-2244-393

Denmark - Copenhagen

Tel: 45-4485-5910

Fax: 45-4485-2829

Finland - Espoo

Tel: 358-9-4520-820

France - Paris

Tel: 33-1-69-53-63-20

Fax: 33-1-69-30-90-79

Germany - Garching

Tel: 49-8931-9700

Germany - Haan

Tel: 49-2129-3766400

Germany - Heilbronn

Tel: 49-7131-72400

Germany - Karlsruhe

Tel: 49-721-625370

Germany - Munich

Tel: 49-89-627-144-0

Fax: 49-89-627-144-44

Germany - Rosenheim

Tel: 49-8031-354-560

Israel - Ra'anana

Tel: 972-9-744-7705

Italy - Milan

Tel: 39-0331-742611

Fax: 39-0331-466781

Italy - Padova

Tel: 39-049-7625286

Netherlands - Drunen

Tel: 31-416-690399

Fax: 31-416-690340

Norway - Trondheim

Tel: 47-7288-4388

Poland - Warsaw

Tel: 48-22-3325737

Romania - Bucharest

Tel: 40-21-407-87-50

Spain - Madrid

Tel: 34-91-708-08-90

Fax: 34-91-708-08-91

Sweden - Gothenberg

Tel: 46-31-704-60-40

Sweden - Stockholm

Tel: 46-8-5090-4654

UK - Wokingham

Tel: 44-118-921-5800

Fax: 44-118-921-5820