

SN65LVCP114

SLLSEA8 -JANUARY 2012

14.2-Gbps Quad 1:2-2:1 MUX, Linear-Redriver With Signal Conditioning

Check for Samples: SN65LVCP114

FEATURES

- Quad 2:1 Mux / 1:2 Demux
- Multi-Rate Operation up to 14.2 Gbps Serial Data Rate
- Linear Receiver Equalization Which Increases Margin at System Level of Decision Feedback Equalizer
- Bandwidth: 18 GHz, Typical
- Per-Lane P/N Pair Inversion
- Port or Single Lane Switching
- Low Power: 150 mW/Channel, Typical
- Loopback Mode on All Three Ports
- I²C Control in Addition to GPIO
- DIAG Mode That Outputs Data of Line Side Port to Both Fabric Side Ports
- 2.5-V/3.3-V Single Power Supply
- PBGA Package 12-mm × 12-mm × 1-mm, 0.8-mm Terminal Pitch
- Excellent Impedance Matching to 100- Ω PCB Transmission Lines

DESCRIPTION

- Small Package Size Provides Board Real Estate Saving
- Adjustable Output Swing Provides Flexible
 EMI and Crosstalk Control
- Low Power
- Supports 10GBASE-KR Applications With Ability to Transparency for Link Training

APPLICATIONS

- High-Speed Redundancy Switch in Telecom and Data Communication
- Backplane Interconnect for 10G-KR, 16GFC

The SN65LVCP114 is an asynchronous, protocol-agnostic, low-latency QUAD mux, linear-redriver optimized for use in systems operating at up to 14.2 Gbps. The device linearly compensates for channel loss in backplane and active-cable applications. The architecture of SN65LVCP114 linear-redriver is designed to work effectively with ASIC or FPGA products implementing digital equalization using decision feedback equalizer (DFE) technology. The SN65LVCP114 mux, linear-redriver preserves the integrity (composition) of the received signal, ensuring optimum DFE and system performance. The SN65LVCP114 provides a low-power mux-demux, linear-redriver solution while at the same time extending the effectiveness of DFE.

SN65LVCP114 is configurable via GPIO or an I²C interface.

A single 2.5-V or 3.3-V power supply supports the operation of the SN65LVCP114.

The SN65LVCP114 is packaged in a 12-mm × 12-mm × 1-mm PBGA package with 0.8-mm pitch.

The SN65LVCP114 has three ports; each port is a quad lane. The switch logic of SN65LVCP114 can be implemented to support a 2:1 MUX per lane, 1:2 DEMUX per lane, and independent lane switching. The receive equalization can be independently programmed for each of the ports. The SN65LVCP114 supports loopback on all three ports.



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

TYPICAL IMPLEMENTATION

SN65LVCP114 can be implemented on the transmit side or the receive side of a backplane channel as shown in Figure 1.



Figure 1. SN65LVCP114 Typical Implementation



BLOCK DIAGRAM

A simplified block diagram of the SN65LVCP114 is shown in Figure 2 for input quad channels AIN and BIN through the 2:1 MUX and output quad channel COUT, together with the input quad channel CIN through the 1:2 DEMUX through output quad channels AOUT and BOUT. The MUX and DEMUX channels contain a linear receive equalizer and an output linear driver.

The SN65LVCP114 provides both GPIO and I^2C interfaces to control the configuration of the device. A detailed description of the SN65LVCP114 pin functions is provided in .



Figure 2. Simplified Block Diagram of SN65LVCP114

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PACKAGE

The package for the SN65LVCP114 is a 12-mm × 12-mm × 1-mm, 167 pin PBGA with 0.8-mm pitch. The top view with the pin names is shown in Figure 3.

	TOP VIEW														
	Α	в	с	D	Е	F	G	н	J	к	L	м	N	Ρ	
14	ADD0/ EQA1	ADD1/ EQB1	EQCO	I2C_SEL	PWD#	AINPO	AINNO	GND	AINP1	AINN1	LN_2_EN	LPC	LPB	LPA	14
13	CINNO	vec	CS	FST_SW	VCC	GND	VCC	VCC	GND	LN_1_EN	LN_3_EN	AOUTP2	VCC	REXT	13
12	CINPO	GND	COUTNO	GND	AOUTPO	AOUTNO	GND	AOUTP1	AOUTNI	LN_O_EN	DIAG	AOUTN2	GND	AINP2	12
11	GND	VCC	СОЛТРО									GND	VCC	AINN2	11
10	CINNI	vcc	GND		GND	GND	GND	GND	GND	GND		AOUTP3	vec	GND	10
9	CINP1	GND	COUTNI		GND	GND	GND	GND	GND	GND		AOUTN3	GND	AINP3	э
8	GND	VCC	COUTP1		GND	GND	GND	GND	GND	GND		GND	GND	AINNS	8
7	CINN2	vcc	GND		GND	GND	GND	GND	GND	GND		BOUTPO	VCC	GND	7
6	CINP2	GND	COUTN2		GND	GND	GND	GND	GND	GND		BOUTNO	GND	BINPO	6
5	GND	vec	COUTP2			GND	GND	GND	GND	GND		GND	VCC	BINNO	5
4	CINN3	vee	GND									BOUTP1	VCC	GND	4
3	CINP3	GND	COUTN3	EQAO	GAIN_C	BOUTN3	BOUTP3	GND	BOUTN2	BOUTP2	GND	BOUTNI	GND	BINP1	3
2	SDA	SCL	COUTP3	GAIN_B	GAIN_A	GND	vcc	VCC	GND	VCC	VOD_C	VOD_B	VOD_A	BINN1	2
1	ADD2/ EQC1	DIS_AGC_A	DIS_AGC_B	DIS_AGC_C	BINN3	BINP3	GND	BINN2	BINP2	EQ_B0	SEL3	SEL2	SEL1	SELO	1
	A	в	с	D	E	F	G	н	J	к	L	м	N	Р	

Figure 3. Package Pinout

PIN DESCRIPTIONS

	PIN		DESCRIPTION
SIGNAL	BALLS	DIRECTION TIPE SUPPLY	DESCRIPTION
LINE-SIDE HIGH	SPEED I/O		
CINP0 CINN0	A12 A13	Input (with 50- Ω termination to input common mode)	Differential input, lane 0 line side.
CINP1 CINN1	A9 A10	Input (with 50- Ω termination to input common mode)	Differential input, lane 1 line side
CINP2 CINN2	A6 A7	Input (with 50- Ω termination to input common mode)	Differential input, lane 2 line side
CINP3 CINN3	A3 A4	Input (with 50- Ω termination to input common mode)	Differential input, lane 3 line side
COUTP0 COUTN0	C11 C12	Output	Differential output, lane 0 line side
COUTP1 COUTN1	C8 C9	Output	Differential output, lane 1 line side
COUTP2 COUTN2	C5 C6	Output	Differential output, lane 2 line side
COUTP3 COUTN3	C2 C3	Output	Differential output, lane 3 line side

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PIN DESCRIPTIONS (continued)

PIN		DIRECTION TYPE SUPPLY	DESCRIPTION					
SIGNAL	BALLS							
SWITCH-SIDE HI	GH-SPEED I/O							
AINP0 AINN0	F14G14	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 0, fabric switch	n_A_side		
AINP1 AINN1	J14 K14	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 1, fabric switch	n_A_side		
AINP2 AINN2	P12 P11	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 2, fabric switch	n_A_side		
AINP3 AINN3	P9 P8	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 3, fabric switch	n_A_side		
BINP0 BINN0	P6 P5	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 0, fabric switch	n_B_side		
BINP1 BINN1	P3 P2	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 1, fabric switch	n_B_side		
BINP2 BINN2	J1 H1	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 2, fabric switch	n_B_side		
BINP3 BINN3	F1 E1	Input (with 50- Ω termination to input common mode)	Different	tial input, lane	e 3, fabric switch	n_B_side		
AOUTP0 AOUTN0	E12 F12	Output	Different	Differential output, lane 0, fabric switch_A_side				
AOUTP1 AOUTN1	H12 J12	Output	Different	Differential output, lane 1, fabric switch_A_side				
AOUTP2 AOUTN2	M13 M12	Output	Differential output, lane 2, fabric switch_A_side					
AOUTP3 AOUTN3	M10 M9	Output	Differential output, lane 3, fabric switch_A_side					
BOUTP0 BOUTN0	M7 M6	Output	Differential output, lane 0, fabric switch_B_side					
BOUTP1 BOUTN1	M4 M3	Output	Different	tial output, lan	ne 1, fabric swite	ch_B_side		
BOUTP2 BOUTN2	К3 J3	Output	Different	tial output, lan	ne 2, fabric swite	ch_B_side		
BOUTP3 BOUTN3	G3 F3	Output	Different	tial output, lan	ne 3, fabric swite	ch_B_side		
CONTROL SIGN	ALS							
ADD0/EQA1 ADD1/EQB1 ADD2/EQC1	A14 B14 A1	Input, 2.5-V/3.3-V CMOS - 3-state	GPIO m EQ cont EQ gain EQ cont EQ gain EQ cont EQ gain	rol pins. EQA of port A. rol pins. EQB of port B. rol pins. EQC of port C.	1 and EQA0 pir 1 and EQB0 pir 1 and EQC0 pir	ns are 3-state and cor ns are 3-state and cor ns are 3-state and cor	ntrol the ntrol the ntrol the	I²C mode ADD0 along with pins ADD1 and ADD2 comprise the three bits of the I ² C slave address.
				EQ[x]0	EQ[x]0	Peaking in dB		
				0	0	1.3		
				0	HiZ	2		
				0	1	3.6		
				HiZ	0	5		
				HiZ	HiZ	6.5		
				HiZ	1	8.3		
				1	0	10		
				1	HiZ	11.9		
				1	1	13.9		

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PIN DESCRIPTIONS (continued)

	PIN	DIRECTION TYPE SUPPLY	UPPLY DESCRIPTION				
SIGNAL	BALLS						
EQA0 EQB0 EQC0	D3 K1 C14	Input, 2.5-V/3.3-V CMOS - 3-state	GPIO mo EQ contro theEQ ga EQ contro theEQ ga EQ contro theEQ ga	ode ol pins. EQA1 ain of port A. ol pins. EQB1 ain of port B. ol pins. EQC ² ain of port C.	and EQA0 pir and EQB0 pir and EQC0 pi	ns are 3-state and contro ns are 3-state and contro ns are 3-state and contro	I ² C mode No action needed
			ſ	EQ[x10	EQ[x]0	Peaking in dB	
			-	0	0	1.3	
				0	HiZ	2	
				0	1	3.6	
				HiZ	0	5	
				HiZ	HiZ	6.5	
				HiZ	1	8.3	
				1	0	10	
			-	1	HiZ	11.9	
				1	1	13.9	
LPA LPB LPC	P14 N14 M14	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	GPIO mo LPx enab HIGH: Lo LOW: Loo See Table	ode bles loopback bopback enab opback disab e 2 and Figur	for port x led led e 13 for device	a logic	I ² C mode No action needed
SEL0	P1	Input (with 48-kΩ pulldown)	GPIO mo	ode Riowitch cont	ral far lana y		I ² C mode
SEL2	M1	2.5-1/3.3-1 CIVIOS	HIGH: po	ort B is selecte	ed		No action needed
SEL3	L1		LOW: por	rt A is selecte			
REYT	P13		See Table	e 2 and Figur	1 200 O to gro		
CS	C13	Input (with 48-kΩ pulldown)	GPIO mo	ode	1,200 12 10 910		J ² C mode
		2.5-V/3.3-V CMOS	No action needed				HIGH: acts as chip select LOW: disables I ² C interface
PWD	E14	Input (with 48-kΩ pullup) 2.5-V/3.3-V CMOS	LOW: Po HIGH: No	wers down th ormal operatio	ie device, inpu on	ts off and outputs disable	d, resets I ² C
DIAG	L12	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	GPIO mo HIGH: En output on LOW: No See Table	ode hables the same both fabric s formal operations e 2 and Figur	me data on the ide ports (Port n e 12 for device	e line side (Port C) to be A and Port B). e logic.	I ² C mode No action needed
LN_0_EN	K12	Input (with 48-kΩ pullup)	GPIO mo	ode			I ² C mode
LN_1_EN LN_2_EN LN_3_EN	K13 L14 L13	2.5-V/3.3-V CMOS	LN_X_EN LN_X_EN	l = High, enal l = Low, disal	bles lane x of p bles lane x of p	ports A, B, and C ports A, B, and C	No action needed
DIS_AGC_A DIS_AGC_B DIS_AGC_C	B1 C1 D1	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	GPIO mo Disables DIS_AGO DIS_AGO	ode the AGC loop C = High, disa C = Low, enat	o internal to the bles the AGC bles the AGC b	e SN65LVCP114 loop pop	I ² C mode No action needed
VOD_A VOD_B VOD_C	N2 M2 L2	Input, 2.5-V/3.3-V CMOS - 3-state	GPIO mo HIGH: se 2.2 LOW: sel 1.1 If left floa	ode Iects VOD ou Iects VOD ou ting, it defaul	tput range: 1.2 tput range: 600 ts to 1.2 V max	2 V maximum and a gain) mV maximum and a gai kimum and a gain of 2.2.	of No action needed
Gain_A Gain_B Gain_C	E2 D2 E3	Input, 2.5-V/3.3-V CMOS - 3-state	GPIO mo HIGH: Re LOW: Re If left floa	ode eceiver gain = eceiver gain = ting, it defaul	= 1 0.5 ts to 0.5		I ² C mode No action needed
SDA	A2	Input / output, open-drain output	GPIO mo No action	ode n needed			I ² C mode I ² C data. Connect a 10-kΩ pullup resistor externally
SCL	B2	Input, open-drain input	GPIO mo No action	ode n needed			I ² C mode I ² C clock. Connect a 10-kΩ pullup resistor externally



PIN DESCRIPTIONS (continued)

PIN			DESCRIPTION				
SIGNAL	BALLS	DIRECTION TYPE SUPPLY	DESCRIPTION				
FST_SW	D13	Input (with 48-kΩ pullup) 2.5-V/3.3-V CMOS input	GPIO mode HIGH: Fast switching; the idle outputs are squelched (see tSM specification). LOW: Slow switching; the idle outputs are powered off (see tSM1 specification).	I ² C mode No action needed			
I2C_SEL	D14	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS input	Configures the device in I ² C or GPIO mode of operation HIGH: Enables I ² C mode LOW: Enables GPIO mode				
POWER SUPPLY	(•					
VCC	B4, B5, B7, B8, B10, B11, B13, E13, G2, G13, H2, H13, K2,N4, N5, N7, N10, N11, N13	Power, 2.5 V ±5% / 3.3 V ±5%	Power supply pins				
GROUND							
GND	A5, A8, A11, B3, B6, B9, B12, C4, C7, C10,D12, F2, F13,G1, G12, G1, G12, H3, H14,J2, J13, L3, M5,M8, M11, N3, N6, N8, N9, N12, P4, P7, P10	Ground	Ground pins				
GND CenterPad	E6, E7, E8, E9, E10, F5, F6, F7, F8, F9, F10, G5, G6, G7,G8, G9, G10, H5, H6, H7, H8, H9, H10, J5, J6, J7, J8, J9, J10, K5, K6, K7,K8, K9, K10	Ground	These pins must be connected to the GND plane.				

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

				VALUE / UNIT		
V _{CC}	Supply volta	age range ⁽²⁾		–0.3 V to 4 V		
V _{IN,DIFF}	Differential	voltage between xINx_P and xINx	_N	±2.5 V		
V _{IN+, IN}	+, IN- Voltage at xINx_P and xINx_N			-0.5 V to VCC + 0.5 V		
V _{IO}	Voltage on control I/O pins			–0.3 V to VCC + 0.5 V		
I _{IN+} I _{IN-}	Continuous	current at high-speed differential	data inputs (differential)	-25 mA to 25 mA		
I _{OUT+} I _{OUT-}	Continuous	current at high-speed differential	-25 mA to 25 mA			
		Human-body model ⁽³⁾	All pins	2 kV		
ESD		Charged-device model ⁽⁴⁾	All pins	500 V		

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

(3) Tested in accordance with JEDEC Standard 22, Test Method A114-A.

(4) Tested in accordance with JEDEC Standard 22, Test Method C101.

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STRUMENTS

EXAS

THERMAL INFORMATION

		SN65LVCP114	
		ZJA (167 PINS)	UNITS
θ_{JA}	Junction-to-ambient thermal resistance ⁽²⁾	38.8	
θ _{JCtop}	Junction-to-case (top) thermal resistance ⁽³⁾	7.55	
θ_{JB}	Junction-to-board thermal resistance ⁽⁴⁾	17.8	°C111
Ψ _{JT}	Junction-to-top characterization parameter ⁽⁵⁾	0.2	C/W
Ψ_{JB}	Junction-to-board characterization parameter ⁽⁶⁾	17.5	
θ_{JCbot}	Junction-to-case (bottom) thermal resistance ⁽⁷⁾	N/A	

For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, SPRA953.
 The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as

specified in JESD51-7, in an environment described in JESD51-2a.

(3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter, ψ_{JT} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter, ψ_{JB} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

RECOMMENDED OPERATING CONDITIONS

SPEC	CIFICATION	MIN	NOM	MAX	UNIT
	Operating data rate, dR			14.2	Gbps
	Supply voltage, V _{CC} , 2.5-V nominal supply	2.375	2.5	2.625	V
	Supply voltage, V _{CC} , 3.3-V nominal supply	3.135	3.3	3.465	V
	PSNR BG, bandgap circuitry PSNR, 10 Hz–10 GHz	20			dB
CON	TROL INPUTS			·	
VIH	High-level input voltage	0.8 × V _{CC}			
V_{IM}	Mid-level input voltage	V _{CC} /2 – 0.3		$V_{CC}/2 + 0.3$	V
V_{IL}	Low- level input voltage			$0.2 \times V_{CC}$	V
T _C	Junction temperature ⁽¹⁾	-10		125	°C
	Maximum Board Temperature ⁽¹⁾			See Table 1	°C

(1) Use of θJB and φJB are recommended for thermal calculations. For more information about traditional and new thermal metrics, see IC Package Thermal Metrics application report, SPRA953.



ELECTRICAL CHARACTERISTICS (V_{cc} 2.5 V ±5%)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER	CONSUMPTION					
PDL	Device power dissipation, loopback mode	Ports A, B, and C in loopback mode with all 12 channels active. VOD = LOW		1800	2300	mW
PD _N	Device power dissipation, normal mode	Device configured in mux-demux mode with 8 channels active. VOD = LOW		1400	1800	mW
PD _{OFF}	Device power dissipation, lanes disabled	All 4 lanes disabled. See the I ² C section for device configuration.		50		mW
PD _{STB}	Device power dissipation, standby	All 12 channels active, VOD = LOW, FAST_SW = HIGH. See the I^2C section for device configuration.		1800	2300	mW

ELECTRICAL CHARACTERISTICS (V_{cc} 3.3 V ±5%)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT			
POWER CONSUMPTION									
PD_L	Device power dissipation, loopback mode	Ports A, B, and C in loopback mode with all 12 channels active. VOD = LOW		2500	3150	mW			
PD _N	Device power dissipation, normal mode	Device configured in mux-demux mode with 8 channels active. VOD = LOW		1800	2500	mW			
PD _{OFF}	Device power dissipation, lanes disabled	All 4 lanes disabled. See the I ² C section for device configuration.		50		mW			
PD _{STB}	Device power dissipation, standby	All 12 channels active, VOD = LOW, FAST_SW = HIGH. See I^2C section for device configuration.		2500	3150	mW			

ELECTRICAL CHARACTERISTICS (V_{cc} 3.3 V \pm 5%, 2.5 V \pm 5%)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT				
CMOS DC SP	ECIFICATIONS									
I _{IH}	High-level input current	$V_{IN} = 0.9 \times V_{CC}$			80	μA				
I _{IL}	Low-level input current	$V_{IN} = 0.1 \times V_{CC}$	-80			μA				
CML INPUTS	(AINP[3:0], AINN[3:0], BINP[3:0], BINN[3:0], CINF	P[3:0], CINN[3:0])								
r _{IN}	Differential input resistance	INx_P to INx_N		100		Ω				
VINPP	Input linear dynamic range	Gain = 0.5		1200		mVpp				
VICM	Common-mode input voltage	Internally biased		$V_{CC} - 0.3$		V				
SCD11	Input differential to common-mode conversion	100 MHz to 7.1GHz		-25		dB				
SDD11	Differential input return loss	100 MHz to 7.1GHz		-10		dB				
CML OUTPUT	CML OUTPUTS (AOUTP[3:0], AOUTN[3:0], BOUTP[3:0], BOUTN[3:0], COUTP[3:0], COUTN[3:0])									
V	Output linear dynamia range	$R_L = 100 \Omega$, $V_{OD} = High$		1200		m\/				
VOD	Output linear dynamic range	$R_L = 100 \Omega$, $V_{OD} = Low$		600		Пурр				
V _{OS}	Output offset voltage	$R_L = 100 \Omega$, 0 V applied at inputs			20	mV_{PP}				
V _{CM,RIP}	Common-mode output ripple	K28.5 pattern at 14.2 Gbps, no interconnect loss, V _{op} = HIGH		10	20	mV _{RMS}				
V _{OD,RIP}	Differential path output ripple	K28.5 pattern at 14.2Gbps, no interconnect loss, $V_{\rm IN}$ = 1200 mVpp. Outputs squelched.			20	mV _{PP}				
V _{OCM}	Output common mode voltage	See Figure 4		V _{CC} - 0.35		V				
V _{OC(SS)}	Change in steady-state common-mode output voltage between logic states			±10		mV				
t _{PLH}	Low-to-high propagation delay	Dee Figure 5		200		ps				
t _{PHL}	High-to-low propagation delay	See Figure 5		200		ps				
t _{SK(O)}	Inter-pair output skew (2)	All outputs terminated with 100 Ω . See Figure 7		50		ps				

(1) All typical values are at 25°C and with 2.5-V and 3.3-V supply, unless otherwise noted.

(2) t_{SK(O)} is the magnitude of the time difference between the channels within a Port. For more information, see SN65LVCP114 Guidelines for Skew Compensation, SLLA323.

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ELECTRICAL CHARACTERISTICS (V_{cc} 3.3 V \pm 5%, 2.5 V \pm 5%) (continued)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	8 8 1			,		
	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
t _{SK(PP)}	Part-to-part skew ⁽³⁾				100	ps
t _R	Rise time	Input signal with 30ps rise time, 20% to 80%. See Figure 6		31		ps
t _F	Fall time	Input signal with 30ps fall time, 20% to 80%. See Figure 6		31		ps
SDD22	Differential output return loss	100 MHz to 7.1GHz		-10		dB
SCC22	Common-mode output return loss	100 MHz to 7.1GHz		-5		dB
t _{SM}		Mux to valid output (idle outputs are squelched)		100		ns
t _{SM1}		Mux to valid output (idle outputs are turned off)		10		μs
Ch	Channel to channel isolation ⁽⁴⁾	Frequency at 5.1625 GHz		52.2		ЧĿ
Chiso	Channel-to-channel isolation**	Frequency at 7.1GHz			ав	
OUT	Output of and a size	10 MHz to 7.1 GHz. No other noise source present. V_{OD} = LOW			1500	
OUTNOISE	Output referred hoise	10 MHz to 7.1 GHz. No other noise source present. V_{OD} = HIGH			3000	P ^v RMS
Vpre	Output pre-cursor pre-emphasis	Input signal with 3.75 pre-cursor and measured on the output signal. See Figure 8. Vpre = 20 log(V3/V2)		5		dB
Vpst	Output post-cursor pre-emphasis	Input signal with 12 dB post-cursor and measure it on the output signal. See Figure 8. Vpst = 20 log(V1/V2)		14		dB
r _{OT}	Single-ended output resistance	Single-ended on-chip terminations to VCC, outputs are ac-coupled		50		Ω
r _{om}	Output termination mismatch at 1 MHz	$\Delta rom = 2 \times \frac{rp - rn}{rp + rn} \times 100$			5	%
EQUALIZATIO	N					
EQ _{Gain}	At 7.1 GHz input signal	Equalization gain, EQ = MAX	10	15		dB
DJ1	TX residual deterministic jitter at 10.3125 Gbps	Tx launch amplitude = 0.6 Vpp, EQ= 1.3 dB, VOD and GAIN are High. Test Channel = 0 ". See Figure 10.		0.08		Ulp-p
DJ2	TX residual deterministic jitter at 14.2 Gbps	Tx launch Amplitude = 0.6 Vpp, EQ=1.3dB, VOD and GAIN are High. Test Channel = 0". See Figure 10.		0.06		Ulp-p
DJ3	RX residual deterministic jitter at 10.3125 Gbps	Tx Launch Amplitude = 0.6 Vpp, test channel = 12" (9dB loss at 5GHz), EQ=13.9dB, VOD and GAIN are High. See Figure 9.		0.04		Ulp-p
DJ4	RX residual deterministic Jitter at 14.2 Gbps	Tx Launch Amplitude = 0.6 Vpp, test channel = 8" (9dB loss at 7GHz), EQ=13.9dB, VOD=LOW and GAIN=HIGH. See		0.08		Ulp-p

(3) t_{SK(PP)} is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

(4) All noise sources added.

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					Maximun Bo	oard Temperatu	re ⁽¹⁾	
LOOP_A	LOOP_B	LOOP_C	DIAG	V _{cc} =	= 2.5V	V _{CC} = 3.3V		
				V _{OD} = LOW	V _{OD} = HIGH	V _{OD} = LOW	V _{OD} = HIGH	
LOW	LOW	LOW	LOW	85°C	85°C	85°C	75°C	
LOW	LOW	LOW	HIGH	85°C	85°C	75°C	System Specific (2)	
LOW	LOW	HIGH	LOW	85°C	85°C	85°C	75°C	
LOW	LOW	HIGH	HIGH	85°C	85°C	85°C	75°C	
LOW	HIGH	LOW	LOW	85°C	85°C	75°C	System Specific ⁽²⁾	
LOW	HIGH	LOW	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	
LOW	HIGH	HIGH	LOW	85°C	85°C	75°C	System Specific ⁽²⁾	
LOW	HIGH	HIGH	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	LOW	LOW	LOW	85°C	85°C	85°C	75°C	
HIGH	LOW	LOW	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	LOW	HIGH	LOW	85°C	85°C	85°C	75°C	
HIGH	LOW	HIGH	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	HIGH	LOW	LOW	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	HIGH	LOW	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	HIGH	HIGH	LOW	85°C	85°C	75°C	System Specific ⁽²⁾	
HIGH	HIGH	HIGH	HIGH	85°C	85°C	75°C	System Specific ⁽²⁾	

Table 1. RECOMMENDED MAXIMUM BOARD TEMPERATURE

(1)

Maximum board temperature is allowed as long as the device maximum junction temperature is not exceeded. Texas Instruments recommends a system thermal and device use case power analysis to decide possible use of a heat sink. (2)

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PARAMETER MEASUREMENT INFORMATION



Figure 4. Common-Mode Output-Voltage Test Circuit



Figure 5. Propagation Delay, Input to Output



Figure 6. Output Rise and Fall Times



Figure 7. Output Inter-Pair Skew







Figure 8. Vpre and Vpost [The Test Pattern is 111111100000000 (Eight 1s, Eight 0s)]



Figure 9. Receive-Side Performance Test Circuit



Figure 10. Transmit-Side Performance Test Circuit

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vcc

48 kΩ

48 kΩ

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PARAMETER MEASUREMENT INFORMATION (continued) EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS





Figure 12. 3-Level Input Biasing Network

vcc

ESD

ESD

FUNCTIONAL DEFINITIONS

Loop_A	Loop_B	Loop_C	DIAG	SEL[3:0]	Output Port A	Output Port B	Output Port C				
0	0	0	0	0	In_Port_C[3:0]	idle	In_Port_A[3:0]				
0	0	0	0	1	idle	In_Port_C[3:0]	In_Port_B[3:0]				
0	0	0	1	0	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_A[3:0]				
0	0	0	1	1	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_B[3:0]				
0	0	1	0	0	In_Port_C[3:0]	Idle	In_Port_C[3:0]				
0	0	1	0	1	Idle	In_Port_C[3:0]	In_Port_C[3:0]				
0	0	1	1	0	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_C[3:0]				
0	0	1	1	1	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_C[3:0]				
0	1	0	0	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_A[3:0]				
0	1	0	0	1	Idle	In_Port_B[3:0]	In_Port_B[3:0]				
0	1	0	1	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_A[3:0]				
0	1	0	1	1	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_B[3:0]				
0	1	1	0	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]				
0	1	1	0	1	Idle	In_Port_B[3:0]	In_Port_C[3:0]				
0	1	1	1	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]				
0	1	1	1	1	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]				
1	0	0	0	0	In_Port_A[3:0]	Idle	In_Port_A[3:0]				
1	0	0	0	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_B[3:0]				
1	0	0	1	0	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_A[3:0]				
1	0	0	1	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_B[3:0]				
1	0	1	0	0	In_Port_A[3:0]	Idle	In_Port_C[3:0]				
1	0	1	0	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]				
1	0	1	1	0	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]				
1	0	1	1	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]				

Table 2. Loopback, DIAG, and SEL Controls

IN



PARAMETER MEASUREMENT INFORMATION (continued) Table 2. Loopback, DIAG, and SEL Controls (continued)

Loop_A	Loop_B	Loop_C	DIAG	SEL[3:0]	Output Port A	Output Port B	Output Port C
1	1	0	0	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_A[3:0]
1	1	0	0	1	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_B[3:0]
1	1	0	1	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_A[3:0]
1	1	0	1	1	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_B[3:0]
1	1	1	0	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_C[3:0]



Figure 13. Loopback Mode

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SELx = 0, DIAG = 1 AOUTx[3:0] = CINx[3:0] AOUTx[3:0] 1X2 Demux CINx[3:0] BOUTx[3:0] BOUTx[3:0] = CINx[3:0] COUTx[3:0] = AINx[3:0] ------2X1 MUX SELx=1, DIAG = 1 AOUTx[3:0] = CINx[3:0] 1X2 Demux CINx[3:0] BOUTx[3:0] = CINx[3:0]

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Figure 14. Diagnostic Mode



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Table 3. Overall Gain Settings

Gain_x	Input Gain [dB]	VOD_x	VOD Gain [dB]	EQ[x]0	EQ[x]1	Total DC gain [dB]	Total EQ gain (7 GHz) [dB]
LOW	-6	LOW	1	LOW	LOW	-5	1.3
LOW	-6	LOW	1	LOW	HiZ	-5	2
LOW	6	LOW	1	LOW	HIGH	-5	3.6
LOW	6	LOW	1	HiZ	LOW	8	5
LOW	6	LOW	1	HiZ	HiZ	8	6.5
LOW	6	LOW	1	HiZ	HIGH	8	8.3
LOW	6	LOW	1	HIGH	LOW	-11	10
LOW	6	LOW	1	HIGH	HiZ	-11	11.9
LOW	-6	LOW	1	HIGH	HIGH	-11	13.9
LOW	-6	HIGH	6.8	LOW	LOW	0.8	1.3
LOW	-6	HIGH	6.8	LOW	HiZ	0.8	2
LOW	6	HIGH	6.8	LOW	HIGH	0.8	3.6
LOW	6	HIGH	6.8	HiZ	LOW	-2.2	5
LOW	6	HIGH	6.8	HiZ	HiZ	-2.2	6.5
LOW	6	HIGH	6.8	HiZ	HIGH	-2.2	8.3
LOW	6	HIGH	6.8	HIGH	LOW	-5.2	10
LOW	-6	HIGH	6.8	HIGH	HiZ	-5.2	11.9
LOW	-6	HIGH	6.8	HIGH	HIGH	-5.2	13.9
HIGH	0	LOW	1	LOW	LOW	1	1.3
HIGH	0	LOW	1	LOW	HiZ	1	2
HIGH	0	LOW	1	LOW	HIGH	1	3.6
HIGH	0	LOW	1	HiZ	LOW	-2	5
HIGH	0	LOW	1	HiZ	HiZ	-2	6.5
HIGH	0	LOW	1	HiZ	HIGH	-2	8.3
HIGH	0	LOW	1	HIGH	LOW	-5	10
HIGH	0	LOW	1	HIGH	HiZ	-5	11.9
HIGH	0	LOW	1	HIGH	HIGH	-5	13.9
HIGH	0	HIGH	6.8	LOW	LOW	6.8	1.3
HIGH	0	HIGH	6.8	LOW	HiZ	6.8	2
HIGH	0	HIGH	6.8	LOW	HIGH	6.8	3.6
HIGH	0	HIGH	6.8	HiZ	LOW	3.8	5
HIGH	0	HIGH	6.8	HiZ	HiZ	3.8	6.5
HIGH	0	HIGH	6.8	HiZ	HIGH	3.8	8.3
HIGH	0	HIGH	6.8	HIGH	LOW	0.8	10
HIGH	0	HIGH	6.8	HIGH	HiZ	0.8	11.9
HIGH	0	HIGH	6.8	HIGH	HIGH	0.8	13.9

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TYPICAL CHARACTERISTICS

Typical operating condition is at $V_{CC} = 2.5$ V and $T_A = 25$ °C, no interconnect line at the output, and with default device settings (unless otherwise noted).

EQ = 1.3 dB,

EQ = 13.9 dB, Gain = 1, VOD = 1.2



Figure 16. Differential Input Return Loss

Figure 17. Differential to Common-Mode Conversion

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TYPICAL CHARACTERISTICS (continued)

Typical operating condition is at $V_{CC} = 2.5$ V and $T_A = 25$ °C, no interconnect line at the output, and with default device settings (unless otherwise noted).



Figure 20. Transmit-Side Typical Application

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TYPICAL CHARACTERISTICS (continued)

Typical operating condition is at $V_{CC} = 2.5$ V and $T_A = 25^{\circ}$ C, no interconnect line at the output, and with default device settings (unless otherwise noted).



Figure 21. Receive-Side Typical Application



APPLICATION INFORMATION

TWO-WIRE SERIAL INTERFACE AND CONTROL LOGIC

The SN65LVCP114 uses a 2-wire serial interface for digital control. The two circuit inputs, SDA and SCL, are driven, respectively, by the serial data and serial clock from a microprocessor, for example. The SDA and SCL pins require external 10-k Ω pullups to V_{CC}.

The 2-wire interface allows write access to the internal memory map to modify control registers and read access to read out the control signals. The SN65LVCP114 is a slave device only, which means that it cannot initiate a transmission itself; it always relies on the availability of the SCL signal for the duration of the transmission. The master device provides the clock signal as well as the START and STOP commands. The protocol for a data transmission is as follows:

- 1. START command
- 7-bit slave address (0000ADD[2:0]) followed by an 8th bit which is the data direction bit (R/W). A zero indicates a WRITE and a 1 indicates a READ. The ADD[2:0] address bits change with the status of the ADD2, ADD1, and ADD0 device pins, respectively. If the pins are left floating or pulled down, the 7-bit slave address is 0000000.
- 3. 8-bit register address
- 4. 8-bit register data word
- 5. STOP command

Regarding timing, the SN65LVCP114 is I²C compatible. The typical timing is shown in Figure 22, and a complete data transfer is shown in Figure 23. Parameters for Figure 22 are defined in Table 4.

Bus Idle: Both SDA and SCL lines remain HIGH

Start data transfer: A change in the state of the SDA line, from HIGH to LOW, while the SCL line is HIGH, defines a START condition (S). Each data transfer is initiated with a START condition.

Stop Data Transfer: A change in the state of the SDA line from LOW to HIGH while the SCL line is HIGH defines a STOP condition (P). Each data transfer is terminated with a STOP condition; however, if the master still must communicate on the bus, it can generate a repeated START condition and address another slave without first generating a STOP condition.

Data Transfer: The number of data bytes transferred between a START and a STOP condition is not limited and is determined by the master device. The receiver acknowledges the transfer of data.

Acknowledge: Each receiving device, when addressed, is obliged to generate an acknowledge bit. The transmitter releases the SDA line, and a device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge clock pulse. Setup and hold times must be taken into account. When a slave-receiver does not acknowledge the slave address, the data line must be left HIGH by the slave. The master can then generate a STOP condition to abort the transfer. If the slave-receiver does acknowledge the slave address but some time later in the transfer cannot receive any more data bytes, the master must abort the transfer. This is indicated by the slave generating the not acknowledge on the first byte to follow. The slave leaves the data line HIGH and the master generates the STOP condition.



Figure 22. Two-Wire Serial Interface Timing Diagram

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SYMBOL	PARAMETER	MIN	MAX	UNIT
f _{SCL}	SCL clock frequency		400	kHz
t _{BUF}	Bus free time between START and STOP conditions	1.3		μs
t _{HDSTA}	Hold time after repeated START condition. After this period, the first clock pulse is generated.	0.6		μs
t _{LOW}	Low period of the SCL clock	1.3		μs
t _{HIGH}	High period of the SCL clock	0.6		μs
t _{SUSTA}	Setup time for a repeated START condition	0.6		μs
t _{HDDAT}	Data hold time	0		μs
t _{SUDAT}	Data setup time	100		ns
t _R	Rise time of both SDA and SCL signals		300	ns
t _F	Fall time of both SDA and SCL signals		300	ns
t _{susto}	Setup time for STOP condition	0.6		μs





Figure 23. Two-wire Serial Interface Data Transfer

SN65LVCP114 Register Mapping Information

Register 0x00 (General Device Settings) R/W											
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0				
SW_GPIO	PWRDOWN						RSVD				

Register 0x01 (Device Control Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
DIAG	LOOP[C]	LOOP[B]	LOOP[A]	SEL[3]	SEL[2]	SEL[1]	SEL[0]			
					*					

Register 0x02 (Port A Control Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
OUT_DIS_0	OUT_DIS_1	OUT_DIS_2	OUT_DIS_3	FAST_SW	RSVD		DIS_AGC			

Register 0x03 (Port A Input Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
INOFF	RSVD	RSVD	RSVD	EQ3	EQ2	EQ1	EQ0			

Register 0x04 (Port A Output Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
		VOD1	VOD0			GAIN1	GAIN0			

Register 0x06 (Port B Control Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
OUT_DIS_0	OUT_DIS_1	OUT_DIS_2	OUT_DIS_3	FAST_SW	RSVD		DIS_AGC			



Register 0x07 (Port B Input Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 INOFF RSVD RSVD RSVD EQ0 EQ3 EQ2 EQ1 Register 0x08 (Port B Output Settings) R/W Bit 4 Bit 0 Bit 7 Bit 6 Bit 5 Bit 3 Bit 2 Bit 1 VOD1 VOD0 GAIN1 GAIN0 Register 0x0A (Port C Control Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 OUT_DIS_0 OUT_DIS_1 OUT_DIS_2 OUT_DIS_3 FAST_SW DIS_AGC RSVD Register 0x0B (Port C Input Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 RSVD INOFF RSVD RSVD EQ3 EQ1 EQ0 EQ2 _ ~ ~ 0v0C /D 2

Register 0x0C (Port C Output Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
		VOD1	VOD0			GAIN1	GAIN0			

Register 0x0D (Reserved Settings) R/W									
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		
RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD		

Register 0x0F (Reserved Settings) Read Only									
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		
RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD		

Register 0x10 (Polarity Control Settings for Port A and B) R/W								
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
POL_B[3]	POL_B[2]	POL_B[1]	POL_B[0]	POL_A[3]	POL_A[2]	POL_A[1]	POL_A[0]	

Register 0x11 (Register 0x11 (Polarity Control Settings for Port C) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0				
				POL_C[3]	POL_C[2]	POL_C[1]	POL_C[0]				

Register 0x12 (Lane Enable) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
				LN_EN[3]	LN_EN[2]	LN_EN[1]	LN_EN[0]			

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT							
	7	Switching logic is controlled by GPIO or I^2C : 1 = GPIO control 0 = I^2C control									
	6										
0x00	5			0000000							
	4										
	3										
	2										
	1										
	1 0 RSVD For TI use only										
	7	DIAG	Enables Diag Mode: 0 = Disable 1 = Enable								
	6	6 LOOP[C] Enables port C loopback: 0 = Disable 1 = Enable									
	5	LOOP[B]	Enables port B loopback: 0 = Disable 1 = Enable								
0.04	4	LOOP[A]	Enables port A loopback: 0 = Disable 1 = Enable	0000000							
0x01	3	SEL[3]	Lane 3, port A/port B switch control: 0 = Port A selected 1 = Port B selected	0000000							
	2	2 SEL[2] Lane 2, port A/port B switch control: 0 = Port A selected 1 = Port B selected									
	1	SEL[1]	EL[1] Lane 1, port A/port B switch control: 0 = Port A selected 1 = Port B selected								
	0	SEL[0]	Lane 0, port A/port B switch control: 0 = Port A selected 1 = Port B selected								



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	Table 5. SN65L	VCP114 Registe	r Descriptions	(continued)
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REGISTER	BIT	SYMBOL		FUNCTION										
	7	OUT_DIS0	Disables 0 = Enal 1 = Disa	s output lane (ble ble):									
	6	OUT_DIS1	Disables 0 = Enal 1 = Disa	Disables output lane 1: = Enable = Disable										
	5	OUT_DIS2	Disables 0 = Enal 1 = Disa	isables output lane 2: = Enable = Disable										
0x02 0x06 0x0A	4	OUT_DIS3	Disables 0 = Enal 1 = Disa	isables output lane 3: = Enable = Disable										
	3	FAST_SW	Fast swi 0 = Idle 1 = Idle	st switch: Idle outputs are disabled (save power) Idle outputs are squelched (fast switch time)										
	2	RSVD	For TI u	For TI use only										
	1													
	0	DIS_AGC	AGC loc 0 = Enal 1 = Disa	GC loop: = Enable = Disable										
	7	IN_OFF	Power d 0 = Norr 1 = Pow	ower down input stages: = Normal = Power down										
	6	RSVD	For TI u	se only										
	5	RSVD	For TI u	se only										
	4	RSVD	For TI u	se only										
	3	EQ3												
0x03	2	EQ2		0	х	x	x	1.3						
0x07	1	EQ1	-	1	0	0	0	2		0000000				
0x0B			1	1	0	0	1	3.6						
			-	1	0	1	0	5						
			-	1	0	1	1	6.5						
			-	1	1	0	0	83						
			-	1	1	0	1	10						
			-	1	1	1	0	11.0						
	0	FOO	-	1	1	1	1	13.0						
	7		<u> </u>	1	1	1	I	13.9						
	6													
	<u>б</u>													
	5	VOD1	00 = 120	ntroi [VOD1:V 00 mV maximi	UDUJ: um									
0x04	4	VOD0	00 = 120 01 = 600 10 = 120 11 = 120	0 mV maximu 00 mV maximu 00 mV maximu	m um um					0000000				
0x0C	3									0000000				
	2													
	1	GAIN1	GAIN co	ontrol [GAIN1:	GAIN0]:									
	0	GAINO	00 = 0.5 01 = 1 10 = 0.5 11 = 1	i										

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT						
	7									
	6			-						
	5									
0x05	4			0000000						
0x09 0x0E	3									
	2									
	1									
	0									
	7	RSVD	For TI use only	_						
	6	RSVD	For TI use only	_						
	5	RSVD	For TI use only	_						
ΟχΟΓ	4	RSVD	For TI use only	0000000						
0,00	3	RSVD	For TI use only	00000000						
	2	RSVD	For TI use only							
	1	RSVD	For TI use only	_						
	0	0 RSVD For TI use only								
0x0F	7	RSVD	For TI use only	_						
	6	RSVD For TI use only								
	5	RSVD	For TI use only	_						
	4	RSVD	For TI use only	00010001						
	3	RSVD	For TI use only	00010001						
	2	RSVD	For TI use only	_						
	1	RSVD	D For TI use only							
	0	RSVD	For TI use only							
	7	POL_B[3]	Polarity switch of output lane 3 of port B: 0 = Normal 1 = Switched							
	6	POL_B[2]	Polarity switch of output lane 2 of port B: 0 = Normal 1 = Switched							
	5	POL_B[1]	Polarity switch of output lane 1 of port B: 0 = Normal 1 = Switched							
0×10	4	POL_B[0]	Polarity switch of output lane 0 of port B OL_B[0] 0 = Normal 1 = Switched							
0.00	3	POL_A[3] Polarity switch of output lane 3 of port A: 0 = Normal 1 = Switched								
	2									
	1	POL_A[1]	Polarity switch of output lane 1 of port A: POL_A[1] 0 = Normal 1 = Switched							
	0	POL_A[0]	Polarity switch of output lane 0 of port A: 0 = Normal 1 = Switched							

Table 5. SN65LVCP114 Register Descriptions (continued)

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT				
	7							
	6							
	5							
	4							
0x11	3	POL_C[3]	Polarity switch of output lane 3 of port C: 0 = Normal 1 = Switched					
	2	POL_C[2] Polarity switch of output lane 2 of port C: 0 = Normal 1 = Switched						
	1	POL_C[1]	Polarity switch of output lane 1 of port C: 0 = Normal 1 = Switched					
	0	POL_C[0]	Polarity switch of output lane 0 of port C: 0 = Normal 1 = Switched					
	7							
	6							
	5							
	4							
	3	LN_EN_3	Lane 3 of ports A, B, and C: 0 = Disable 1 = Enable					
0x12	2	LN_EN_2	Lane 2 of ports A, B, and C: 0 = Disable 1 = Enable	00001111				
	1	LN_EN_1	Lane 1 of ports A, B, and C: 0 = Disable 1 = Enable					
	0	LN_EN_0	Lane 0 of ports A, B, and C: 0 = Disable 1 = Enable					

Table 5. SN65LVCP114 Register Descriptions (continued)



PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
SN65LVCP114ZJA	ACTIVE	NFBGA	ZJA	167	189	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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ZJA (S-PBGA-N167)

PLASTIC BALL GRID ARRAY



A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 B. This drawing is subject to change without notice.

C. This is a Pb-free solder ball design.



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