

ON Semiconductor®

Advance Information

# **Inverter IPM** for 3-phase Motor Drive

#### Overview

This "Inverter IPM" includes the output stage of a 3-phase inverter, pre-drive circuits, and bootstrap circuits as well as protection circuits in one package.

#### **Function**

- SIP(Single in-line package) using a solid transfer mold.
- Protective circuits including over current and pre-drive low voltage protection are built in.
- Direct input of CMOS level control signals without an insulating circuit is possible.
- A single power supply drive is enabled through the use of bootstrap circuits for upper IGBT gate drives.
- Overcurrent protection current is programmable with an external resistance.

#### Certification

• UL1557 (File Number : E339285).

#### **Specifications**

#### **Absolute Maximum Ratings** at Tc = 25°C

Parameter	Symbol	Conditions		Ratings	Unit
Supply voltage	V <sub>CC</sub>	+ to -, surge<500V	*1	450	V
Collector-emitter voltage	V <sub>CE</sub>	+ to U,V,W or U,V,W to -		600	V
Outroit surrout	1-	+, -, U,V,W terminal current		±10	Α
Output current	lo	+, -, U,V,W terminal current at Tc=100°C		±5	Α
Output peak current	lop	+, -, U,V,W terminal current for a Pulse width of 1ms		±20	Α
Pre-driver voltage	VD1,2,3,4	VB1 to U, VB2 to V, VB3 to W, VDD to VSS	*2	20	V
Input signal voltage	VIN	HIN1, 2, 3, LIN1, 2, 3		$-0.3$ to $V_{\hbox{\scriptsize DD}}$	V
FAULT terminal voltage	VFAULT	FAULT terminal		$-0.3$ to $V_{\mbox{\scriptsize DD}}$	V
Maximum power dissipation	Pd	IGBT per channel		22	W
Junction temperature	Tj	IGBT,FRD		150	°C
Storage temperature	Tstg			-40 to +125	°C
Operating substrate temperature	Tc	IPM case temperature		-40 to +100	°C
Tightening torque		Case mounting screws	*3	0.9	Nm
Withstand voltage	Vis	50Hz sine wave AC 1 minute	*4	2000	VR MS

Reference voltage is " $V_{SS}$ " terminal voltage unless otherwise specified.

- \*1 : Surge voltage developed by the switching operation due to the wiring inductance between + and terminal.
- \*2: VD1=VB1 to U, VD2=VB2 to V, VD3=VB3 to W, VD4=V<sub>DD</sub> to V<sub>SS</sub> terminal voltage.
- \*3: Flatness of the heat-sink should be less than 0.15mm.
- \*4: Test conditions: AC2500V, 1 second.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

This document contains information on a new product. Specifications and information herein are subject to change without notice.

#### **ORDERING INFORMATION**

See detailed ordering and shipping information on page 16 of this data sheet.

#### **Electrical Characteristics** at $Tc = 25^{\circ}C$ , VD1, VD2, VD3, VD4 = 15V

Danamatan	O. made al	0.	O a a l'Ula a a			Ratings		Linit	
Parameter	Symbol	Conditions		circuit	min	typ	max	Unit	
Power output section									
Collector-emitter cut-off current	ICE	V <sub>CE</sub> = 600V		F: 4	-	-	0.1	mA	
Bootstrap diode reverse current	IR(BD)	VR(BD)		Fig.1	-	-	0.1	mA	
		Ic=10A	Upper side		-	1.4	2.3		
Collector to emitter	\(\(\sigma\)	Tj=25°C	Lower side *1	F: 0	-	1.7	2.6	l	
saturation voltage	VCE(SAT)	Ic=5A	Upper side	Fig.2	-	1.3	-	V	
		Tj=100°C	Lower side *1		-	1.6	-		
		IF=-10A	Upper side		-	1.3	2.2		
Diedo for and allere	\/_	Tj=25°C	Lower side *1	F: 0	-	1.6	2.5		
Diode forward voltage	VF	IF=-5A	Upper side	Fig.3	-	1.2	-	V	
		Tj=100°C	Tj=100°C Lower side *1		-	1.5	-		
Junction to case	θj-c(T)	IGBT			-	-	5.5		
thermal resistance	θj-c(D)	FRD			-	-	6.5	°C/W	
Control (Pre-driver) section	•				•	•	•		
		VD1, 2, 3 = 15V VD4 = 15V		Fig.4	-	0.08	0.4		
Pre-driver power dissipation	ID				-	1.6	4	mA	
High level Input voltage	Vin H				-	-	0.8	V	
Low level Input voltage	Vin L	HIN1, HIN2, I LIN1, LIN2, L			2.5	-	-	٧	
Input threshold voltage hysteresis	Vinth(hys)	LIIN I, LIINZ, L	1143 to VSS		0.5	0.8	-	V	
Logic 0 input leakage current	I <sub>IN+</sub>	VIN=+3.3V			76	118	160	μА	
Logic 1 input leakage current	I <sub>IN-</sub>	VIN=0V			97	150	203	μА	
FAULT terminal input electric current	IoSD	FAULT : ON/	VFAULT=0.1V		-	2	-	mA	
FAULT clearance delay time	FLTCLR	From time fau	ult condition clears.		6	9	12	ms	
VCC and VS undervoltage upper threshold.	V <sub>CCUV+</sub> V <sub>SUV+</sub>				10.5	11.1	11.7	V	
VCC and VS undervoltage lower threshold.	V <sub>CCUV</sub> - V <sub>SUV</sub> -				10.3	10.9	11.5	V	
VCC and VS undervoltage hysteresis	V <sub>CCUVH</sub> V <sub>SUVH</sub> -				0.14	0.2	-	Α	
Over current protection current	ISD	PW=100µs, F	RSD = 0Ω	Fig.5	10	-	17	Α	
Electric current output signal level	ISO	Io = 10A			0.30	0.33	0.36	V	

Reference voltage is " $V_{SS}$ " terminal voltage unless otherwise specified.

<sup>\*1 :</sup> The lower side's VCE(SAT) and VF include a loss by the shunt resistance

		0 177	Test		l lait		
Parameter	Symbol	Conditions	circuit	min	typ	max	Unit
Switching Character							
0 " "	tON	Io = 10A	F: 0	0.3	0.6	1.3	
Switching time	tOFF	Inductive load	Fig.6	-	1.0	1.8	μS
Turn-on switching loss	Eon	Ic=5A,V <sup>+</sup> =300V,		-	240	-	uJ
Turn-off switching loss	Eoff	V <sub>DD</sub> =15V,L=3.9mH Tc=25°C	Fig.6	-	220	-	uJ
Total switching loss	Etot			-	460	-	uJ
Turn-on switching loss	Eon	Ic=5A,V*=300V,		-	300	-	uJ
Turn-off switching loss	Eoff	V <sub>DD</sub> =15V,L=3.9mH	Fig.6	-	260	-	uJ
Total switching loss	Etot	Tc=100°C		-	560	-	uJ
Diode reverse recovery energy	Erec	I <sub>F</sub> =5A, V <sup>+</sup> =400V, V <sub>DD</sub> =15V,		-	17	-	uJ
Diode reverse recovery time	Trr	L=3.9mH, Tc=100°C		-	62	-	ns
Reverse bias safe operating area	RBSOA	Io =20A, VCE = 450V	Fig.7		Full square		
Short circuit safe operating area	SCSOA	VCE = 400V, Tc=100°C		4	-	-	μS
Allowable offset voltage slew rate	dv/dt	Between U, V, W to -		-50	-	50	V/ns

Reference voltage is " $V_{SS}$ " terminal voltage unless otherwise specified.

#### Notes:

- Input ON voltage indicates a value to turn on output stage IGBT of Vin >= Vin H.
   Input OFF voltage indicates a value to turn off output stage IGBT of Vin < Vin L.</li>
- 2. Please refer to Figure 8 below.

When an under voltage condition occurs the internal protection circuit operation is indicated by FAULT asserted via an active open-drain pulldown.

This signal is non-latching and will release as soon as the fault condition has cleared.

Upon detection of a fault condition all input signals should be set low externally to ensure all outputs are off.

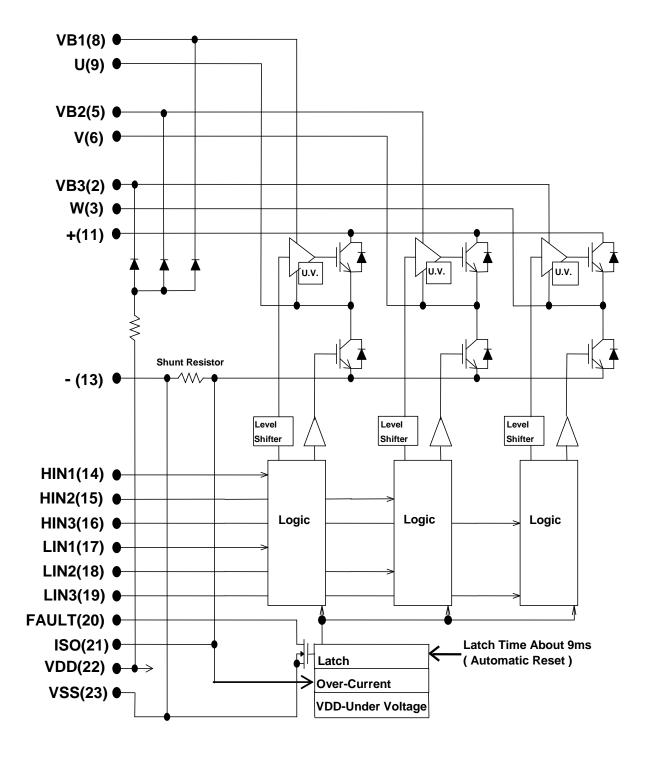
This signal has about 200mV of hysteresis.

Fault condition operation:

- 1. An under voltage fault will turn off all gates and turn off all output IGBT's.
- 2. The under voltage lockut will persist until the fault condition clears and will then release the open-drain output immediately.
- 3. The under voltage FAULT output is not asserted if the only fault is upper gate drive voltage although the upper and lower gates will be turned off.
- 3. When assembling the IPM on the heat sink, tightening torque range is 0.6Nm to 0.9Nm. Please refer to the Application note for more information.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

#### **Equivalent Block Diagram**



# **Module Pin-Out Description**

Pin	Name	Description
1	NA	None
2	VB3	High Side Floating Supply Voltage 3
3	W	Output 3 - High Side Floating Supply Offset Voltage
4	NA	None
5	VB2	High Side Floating Supply voltage 2
6	V	Output 2 - High Side Floating Supply Offset Voltage
7	NA	None
8	VB1	High Side Floating Supply voltage 1
9	U	Output 1 - High Side Floating Supply Offset Voltage
10	NA	None
11	+	Positive Bus Input Voltage
12	NA	None
13	-	Negative Bus Input Voltage
14	HIN1	Logic Input High Side Gate Driver - Phase U
15	HIN2	Logic Input High Side Gate Driver - Phase V
16	HIN3	Logic Input High Side Gate Driver - Phase W
17	LIN1	Logic Input Low Side Gate Driver - Phase U
18	LIN2	Logic Input Low Side Gate Driver - Phase V
19	LIN3	Logic Input Low Side Gate Driver - Phase W
20	FAULT	Fault output
21	ISO	Current monitor output
22	VDD	+15V Main Supply
23	VSS	Negative Main Supply

#### **Test Circuit**

The tested phase U+ shows the upper side of the U phase and U- shows the lower side of the U phase.

#### ■ ICE / IR(BD)

	U+	V+	W+	U-	V-	W-
М	11	11	11	9	6	3
N	9	6	3	13	13	13

	U(BD)	V(BD)	W(BD)
М	8	5	2
N	23	23	23

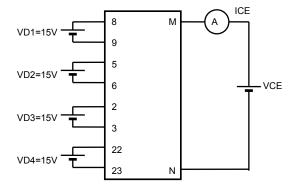


Fig.1

#### ■ VCE(SAT) (test by pulse)

	U+	V+	W+	J	V-	W-
М	11	11	11	9	6	3
N	9	6	3	13	13	13
m	14	15	16	17	18	19

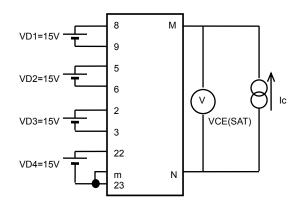
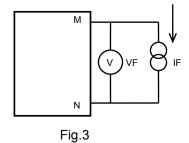


Fig.2

#### ■ VF (test by pulse)

	U+	V+	W+	U-	V-	W-
М	11	11	11	9	6	3
N	9	6	3	13	13	13



■ ID

	VD1	VD2	VD3	VD4
М	8	5	2	22
N	9	6	3	23

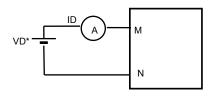
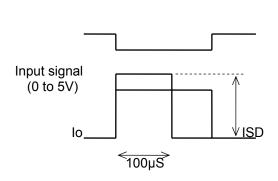


Fig.4

■ ISD



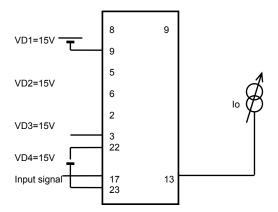
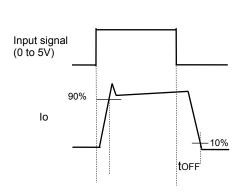


Fig.5

■ Switching time (The circuit is a representative example of the lower side U phase.)



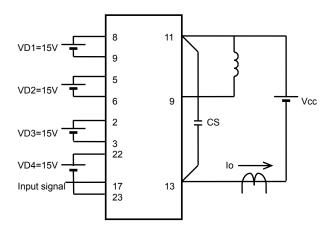
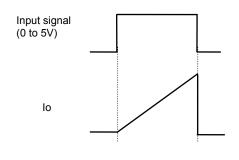


Fig.6

■ RB-SOA (The circuit is a representative example of the lower side U phase.)



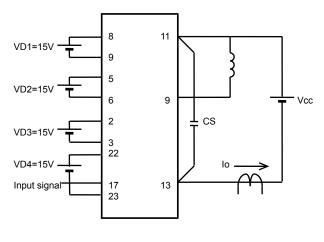


Fig.7

#### **Input / Output Timing Diagram**

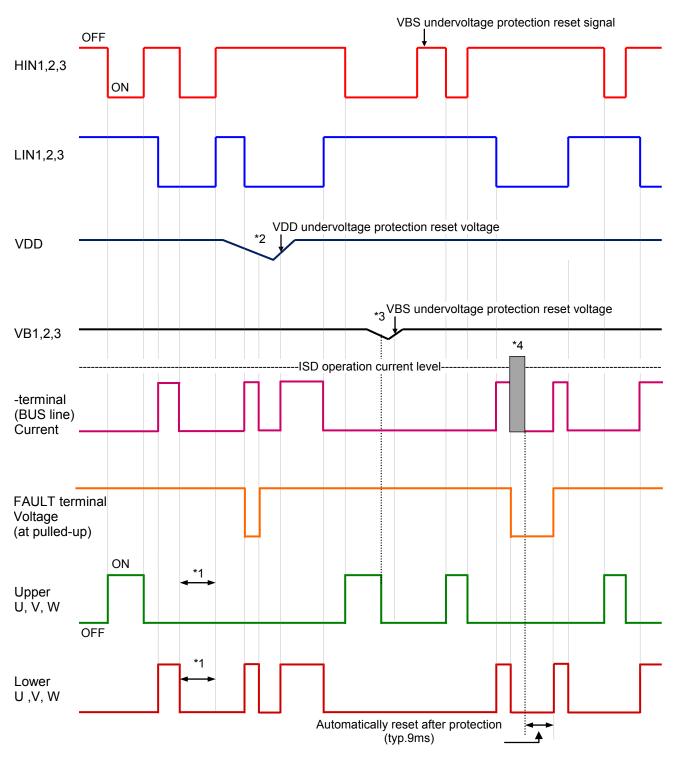
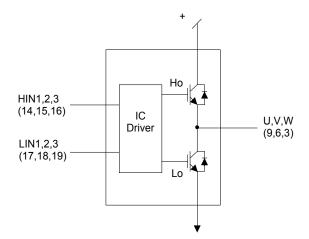


Fig.8

#### Notes

- 1. \*1 : Diagram shows the prevention of shoot-through via control logic. More deadtime to account for switching delay needs to be added externally.
- \*2 : If lower VDD drops all gate output signals will go low and cut off all of 6 IGBT outputs. part. When VDD rises the operation will resume immediately.
- 3. \*3: When the upper side gate voltage at VB1, VB2 and VB3 drops only the corresponding upper side output is turned off. The outputs return to normal operation immediately after the upper side gat voltage rises.
- 4. \*4: In case of over current detection all IGBT's are turned off and the FAULT output is asserted. Normal operation resumes 9ms(typ.) after the over current condition is corrected.

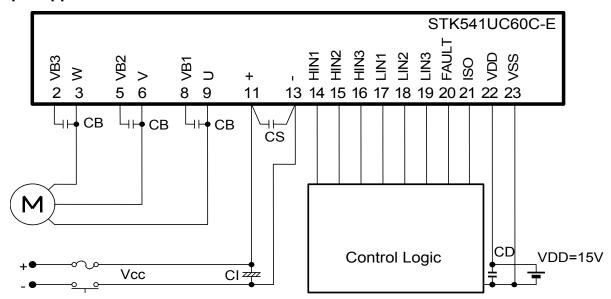
# Logic level table



1=High / 0=Low **FAULT** HIN1,2,3 LIN1,2,3 U,V,W Itrip 1 0 0 0 Off 1 0 0 1 V+ 1 0 1 0 0 1 0 1 1 Off 1 Χ Off 1 Χ 0 Χ Χ Χ Off

Fig. 9

#### **Sample Application Circuit**



#### Recommended Operating Conditions at Ta = 25°C

14	Commelle et	Complete Conditions		Ratings			
Item	Symbol	Conditions	min	typ	max	Unit	
Supply voltage	VCC	Between V+ to VRU(VRV,VRW)	0	280	450	V	
Dro driver europhy veltage	VD1,2,3	Between VB1 to U,VB2 to V,VB3 to W	12.5	15	17.5	٧	
Pre-driver supply voltage	VD4	Between VDD to VSS *1	13.5	15	16.5		
ON-state input voltage	VIN(ON)		0	-	0.3	.,,	
OFF-state input voltage	VIN(OFF)	HIN1,HIN2,HIN3,LIN1,LIN2,LIN3 terminal	3.0	-	5.0	V	
PWM frequency	fPWM		1	-	20	kHz	
Dead time	DT	Turn-off to turn-on (External)	2	-	-	μS	
Mounting torque	-	'M3' type screw	0.6	-	0.9	Nm	

<sup>\*1 :</sup> Pre-drive power supply (VD4=15±1.5V) must be have the capacity of Io=20mA(DC), 0.5A(Peak).

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

#### **Usage Precaution**

- 1. This IPM includes bootstrap diode and resistor. Therefore, by adding a capacitor (CB: about 1 to 47 $\mu$ F), a single power supply drive is enabled. In this case, an electric charge is charged to "CB" by making lower side IGBT turn on. And, please select the capacitance of "CB" (externally set) equal to or less than 47 $\mu$ F( $\pm$ 20%). If selecting the capacitance more than 47 $\mu$ F( $\pm$ 20%), connect a resistor(about 20 $\Omega$ )in series between each 3-phase upper side power supply terminals(VB1,2,3) and each bootstrap capacitor. Also, the upper side power supply voltage sometimes declines by the way of controlling. Please confirm the voltage with an actual set. (When not using the bootstrap circuit, each upper side pre-drive power supply needs an external independent power supply.)
- 2. Because the jump voltage which is accompanied by the vibration in case of switching operation occurs by the influence of the floating inductance of the wiring of the outer power supply which is connected with of the "+" and "-" terminal, restrains and spares serge voltage being as the connection of the snubber circuit (Capacitor / CS /about 0.1uF to 10uF) for the voltage absorption with the neighborhood as possible between the "+" and the point of intersection of the "-" terminal, and so on, with making a wiring length (among the terminals each from "CI") short and making a wiring inductance small.
- 3. The "FAULT" terminal (20pin) is open Drain (It is operating as "FAULT" when becoming Low). This terminal serves as the shut down function of the built-in pre-driver. (When the terminal voltage is above 3V, normalcy works, and it is shut down when it is equal to or less than 0.8V. ) Please make pulling up outside so that "FAULT" terminal voltages become more than 3V. When the pull up voltage (VP) is at 5V, pull up resistor (RP) connects above  $10k\Omega$ , and in case of VP=15V, RP connects above  $39k\Omega$ .
- 4. The pull-down resistor  $(:33k\Omega(typ))$  is connected with the inside of the signal input terminal, but please connect the pull-down resistor (about  $2.2 \text{ to } 3.3k\Omega$ ) outside to decrease the influence of the noise by wiring etc.

- 5. Because the IPM can be destroyed when the motor connection terminal (pins 9, 6, and 3) is opened while the motor is running, please be especially careful of the connection (soldering condition) of this terminal.
- 6. "Iso" terminal (21pin) is for the electric current monitor. Be careful, because the overcurrent protection does not operate when short-circuiting "Iso" terminal and "Vss" terminal (23pin).
- 7. When input pulse width is less than 2µs, an output may not react to the pulse.(Both ON signal and OFF signal)
- 8. The overcurrent protection feature is not intended to protect in exceptional fault conditions. An external fuse is recommended for safety.
- 9. An external resistor "RSD" is required for proper overcurrent protection.
  - This data shows the example of the application circuit, does not guarantee a design as the mass production set.

#### The characteristic of PWM switching frequency

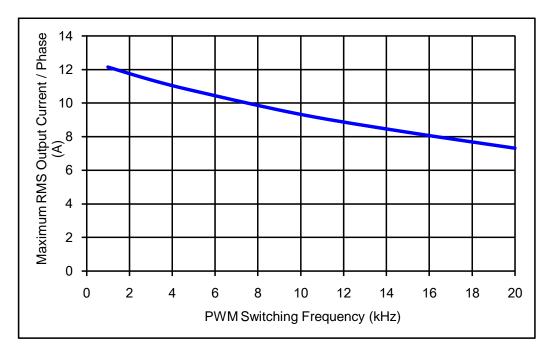


Fig. 10 Maximum sinusoidal phase current as function of switching frequency at Tc=100°C, V<sub>CC</sub>=400V

#### **Switching waveform**

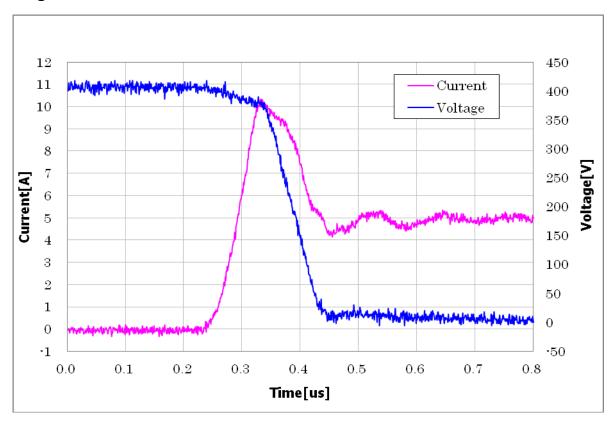


Fig. 11 IGBT Turn-on. Typical turn-on waveform at Tc=100°C, V<sub>C</sub>C=400V

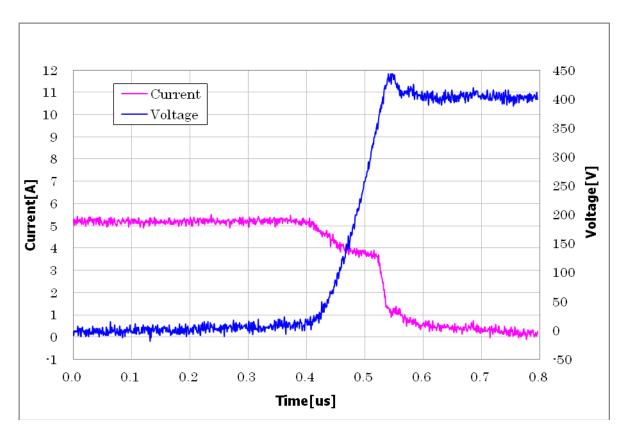


Fig. 12 IGBT Turn-off. Typical turn-off waveform at Tc=100°C, V<sub>CC</sub>=400V

#### CB capacitor value calculation for bootstrap circuit

#### **Calculate conditions**

Parameter	Symbol	Value	Unit
Upper side power supply.	VBS	15	V
Total gate charge of output power IGBT at 15V.	QG	89	nC
Upper limit power supply low voltage protection.	UVLO	12	V
Upper side power dissipation.	IDMAX	400	μΑ
ON time required for CB voltage to fall from 15V to UVLO	TONMAX	-	s

#### Capacitance calculation formula

Thus, the following formula are true VBS x CB - QG - IDMAX \* TONMAX = UVLO \* CB therefore, CB = (QG + IDMAX \* TONMAX) / (VBS - UVLO)

The relationship between TONMAX and CB becomes as follows. CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to  $47\mu\text{F}$ , however, this value needs to be verified prior to production.

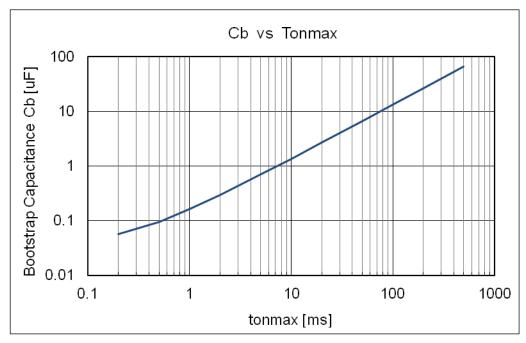


Fig. 13 TONMAX vs CB characteristic

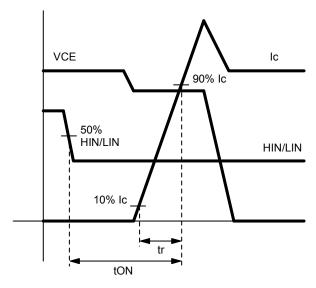


Fig. 14a Input to output propagation turn-on delay time

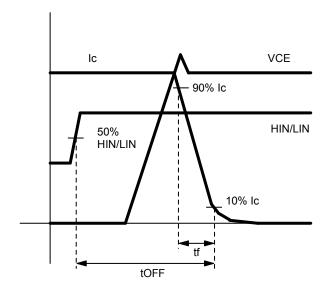


Fig. 14b Input to output propagation turn-off delay time

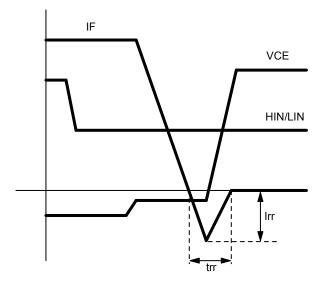


Fig. 14c Diode reverse recovery

#### **Package Dimensions**

unit: mm

note2 note3

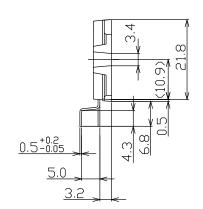
STK541UC60C

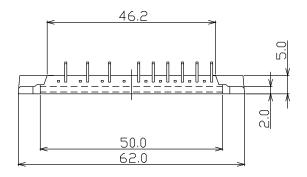
note1 1 23

20.0 2.0 0.6 +0.25

21x2.0 = 42.0

missing pin : 1, 4, 7, 10, 12





note1 : Mark for No.1 pin identification. note2 : The form of a character in this drawing differs from that of IPM.

note3 : This indicates the lot code.

The form of a character in this drawing differs from that of IPM.

#### ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
STK541UC60C-E	SIP23 56x21.8 (Pb-Free)	8 / Tube

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