

High Performance 1A Linear Regulator with Programmable Current Limiting

ISL80101A

The [ISL80101A](#) is a low dropout voltage, single output LDO with programmable current limiting. This LDO operates from input voltages of 2.2V to 6V, and is capable of providing output voltages of 0.8V to 5V. Other custom voltage options are available upon request.

A submicron BiCMOS process is utilized for this product family to deliver the best-in-class analog performance and overall value. The programmable current limiting improves system reliability of end applications. An external capacitor on the soft-start pin provides an adjustable soft-starting ramp. The ENABLE feature allows the part to be placed into a low quiescent current shutdown mode.

This CMOS LDO will consume significantly lower quiescent current as a function of load compared to bipolar LDOs, which translates into higher efficiency and packages with smaller footprints. Quiescent current is modestly compromised to achieve a very fast load transient response.

[Table 1](#) shows the differences between the ISL80101A and others in its family:

TABLE 1. KEY DIFFERENCES BETWEEN FAMILY OF PARTS

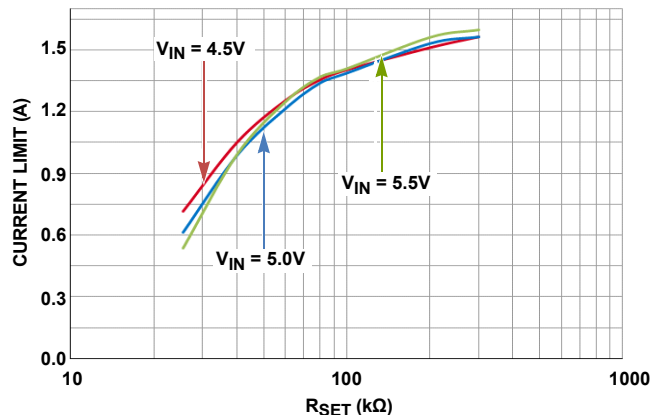
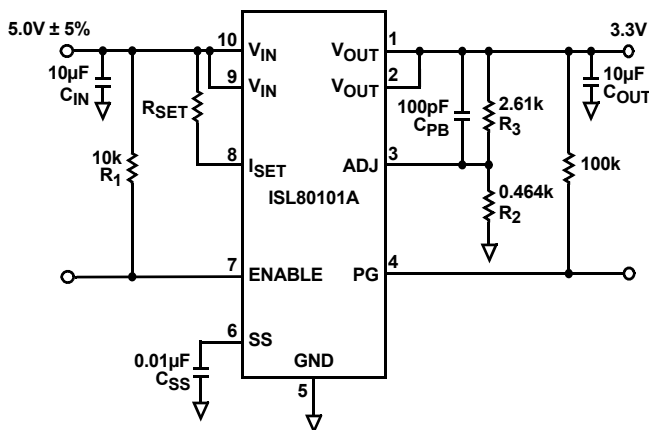
PART NUMBER	PROGRAMMABLE I _{LIMIT}	I _{LIMIT} (DEFAULT)	ADJ or FIXED V _{OUT}
ISL80101-ADJ	No	1.75A	ADJ
ISL80101	No	1.75A	1.8V, 2.5V, 3.3V, 5.0V
ISL80101A	Yes	1.62A	ADJ
ISL80121-5	Yes	0.75A	5.0V

Features

- ±2% V_{ADJ} accuracy guaranteed over line, load and T_J = -40 °C to +125 °C
- Very low 212mV dropout voltage at V_{IN} = 4.5V
- High accuracy current limit programmable up to 1.75A
- Very fast transient response
- 100µV_{RMS} output noise
- Power-good output
- Programmable soft-start
- Over-temperature protection
- Small 10 Ld DFN package

Applications

- Telecommunications and networking
- Medical equipment
- Instrumentation systems
- USB devices
- Gaming
- Routers and switchers

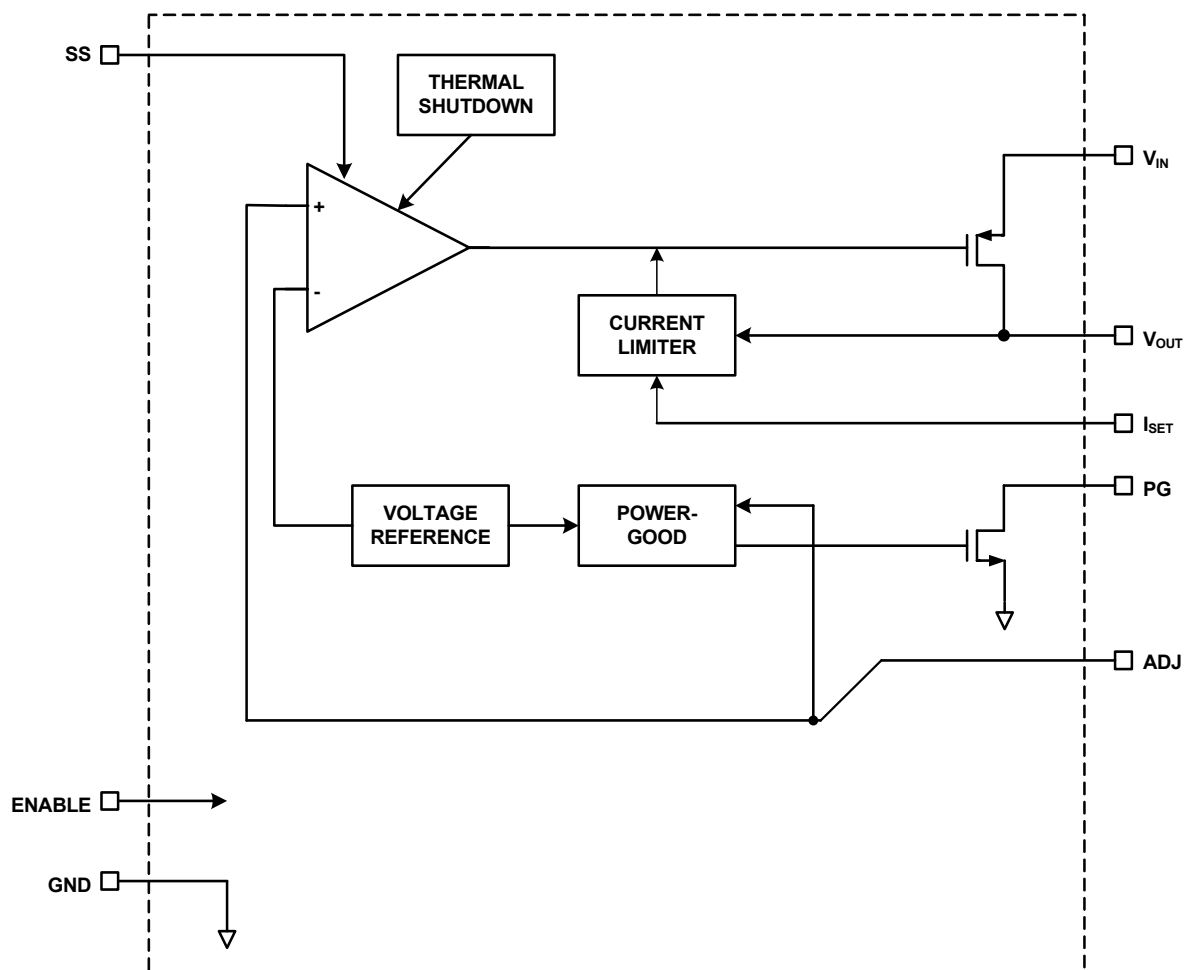


$$I_{LIMIT} \sim 1.62 - \frac{2.9 \times 2 \times V_{IN} - 1}{R_{SET} \text{ k}\Omega}$$

FIGURE 1. TYPICAL APPLICATION

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Block Diagram



Ordering Information

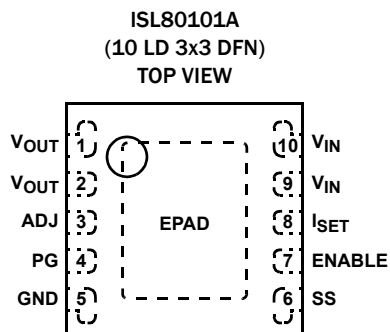
PART NUMBER (Notes 1, 2, 3)	PART MARKING	V _{OUT} VOLTAGE	TEMP. RANGE (°C)	PACKAGE (RoHS Compliant)	PKG DWG. #
ISL80101AIRAJZ	DZAC	ADJ	-40 to +125	10 Ld 3x3 DFN	L10.3x3
ISL80101AEVAL2Z	Evaluation Board				

NOTES:

1. Add "-T*" suffix for tape and reel. Please refer to [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see device information page for [ISL80101A](#). For more information on MSL please see techbrief [TB363](#).

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Pin Configuration



Pin Descriptions

PIN NUMBER	PIN NAME	DESCRIPTION
1, 2	V _{OUT}	Output voltage. A minimum 10μF X5R/X7R output capacitor (for V _{OUT} from 1.5V to 5V) is required for stability. See “External Capacitor Requirements” on page 8 for more details.
3	ADJ	LDO output feedback input. To adjust the output voltage, connect this pin to a resistive voltage divider from the V _{OUT} to GND.
4	PG	V _{OUT} in regulation signal. Logic low indicates V _{OUT} is not in regulation, and must be grounded if not used.
5	GND	Ground.
6	SS	External capacitor adjusts in-rush current.
7	ENABLE	V _{IN} -independent chip enable. TTL and CMOS compatible.
8	I _{SET}	Current limit setting. Current limit is 1.62A when this pin is left floating. This default value can be increased by tying R _{SET} to GND, or decreased by tying R _{SET} to V _{IN} . See “Programmable Current Limit” on page 8 for more details.
9, 10	V _{IN}	Input supply. A minimum of 10μF X5R/X7R input capacitor is required for stability. See “External Capacitor Requirements” on page 8 for more details.
-	EPAD	EPAD at ground potential. Soldering it directly to GND plane is required for thermal considerations. See “Power Dissipation and Thermals” on page 9 for more details.

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Absolute Maximum Ratings (Note 6)

V_{IN} Relative to GND	-0.3V to +6.5V
V_{OUT} Relative to GND	-0.3V to +6.5V
PG, ENABLE, ADJ, SS, I_{SET}	
Relative to GND	-0.3V to +6.5V
ESD Rating	
Human Body Model (Tested per JEDEC)	2.5kV
Machine Model (Tested per JEDEC)	250V
Latch Up (Tested per JEDEC)	± 100 mA at +125°C

Thermal Information

Thermal Resistance (Typical)	θ_{JA} (°C/W)	θ_{JC} (°C/W)
10 Ld 3x3 DFN Package (Notes 4, 5)	48	7
Maximum Junction Temperature (Plastic Package)	+150°C	
Storage Temperature Range	-65°C to +150°C	
Pb-Free Reflow Profile	see TB493	

Recommended Operating Conditions (Note 7)

Junction Temperature Range (T_J)	-40°C to +125°C
V_{IN} Relative to GND	2.2V to 6V
V_{OUT} Range	800mV to 5V
PG, ENABLE, ADJ, SS, I_{SET} Relative to GND	0V to 6V
PG Sink Current	10mA

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with “direct attach” features. See Tech Brief [TB379](#).
- For θ_{JC} , the “case temp” location is the center of the exposed metal pad on the package underside.
- Absolute maximum voltage rating is defined as the voltage applied for a lifetime average duty cycle above 6V of 1%.
- Electromigration specification defined as lifetime average junction temperature of +110°C where max rated DC current = lifetime average current.

Electrical Specifications

Unless otherwise noted, all parameters are established over the following specified conditions: 2.2V < V_{IN} < 6V, V_{OUT} = 0.5V, T_J = +25°C, I_{LOAD} = 0A. Applications must follow thermal guidelines of the package to determine worst case junction temperature. Please refer to [“Functional Description” on page 8](#) and Tech Brief [TB379](#). **Boldface limits apply across the operating temperature range, -40°C to +125°C.** Pulse load techniques used by ATE to ensure $T_J = T_A$ defines established limits.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNITS
DC CHARACTERISTICS						
DC ADJ Pin Voltage Accuracy	V_{ADJ}	$V_{OUT} + 0.4V < V_{IN} < 6V$, $V_{OUT} = 2.5V$; $0A < I_{LOAD} < 1A$	490	500	510	mV
DC Input Line Regulation	$(V_{OUT} \text{ low line} - V_{OUT} \text{ high line}) / V_{OUT} \text{ low line}$	$V_{OUT} + 0.4V < V_{IN} < 6V$, $V_{OUT} = 2.5V$	-1	0.2	1	%
DC Output Load Regulation	$(V_{OUT} \text{ no load} - V_{OUT} \text{ high load}) / V_{OUT} \text{ no load}$	$V_{OUT} + 0.4V < V_{IN} < 6V$, $V_{OUT} = 2.5V$; $0A < I_{LOAD} < 1A$	-1		1	%
Feedback Input Current		$V_{ADJ} = 0.5V$		0.01	1	μA
Ground Pin Current	I_Q	$I_{LOAD} = 0A$, $V_{OUT} + 0.4V < V_{IN} < 6V$, $V_{OUT} = 2.5V$		3	5	mA
		$I_{LOAD} = 1A$, $V_{OUT} + 0.4V < V_{IN} < 6V$, $V_{OUT} = 2.5V$		5	7	mA
Ground Pin Current in Shutdown	I_{SHDN}	ENABLE = 0.2V, $V_{IN} = 6V$		0.2	12	μA
Dropout Voltage (Note 9)	V_{DO}	$I_{LOAD} = 1A$, $V_{IN} = 4.5V$, $V_{ADJ} = 0V$		90	212	mV
Output Current Limit	I_{LIMIT}	$4.5V < V_{IN} < 6V$, I_{SET} is floating		1.62		A
		$V_{IN} = 5V$, $R_{SET} = 25.5k\Omega$	0.540	0.640	0.740	A
Thermal Shutdown Temperature	TSD			160		°C
Thermal Shutdown Hysteresis	TSDn			30		°C

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Electrical Specifications Unless otherwise noted, all parameters are established over the following specified conditions: $2.2V < V_{IN} < 6V$, $V_{OUT} = 0.5V$, $T_J = +25^\circ C$, $I_{LOAD} = 0A$. Applications must follow thermal guidelines of the package to determine worst case junction temperature. Please refer to [“Functional Description” on page 8](#) and Tech Brief [TB379](#). **Boldface limits apply across the operating temperature range, $-40^\circ C$ to $+125^\circ C$.** Pulse load techniques used by ATE to ensure $T_J = T_A$ defines established limits. (Continued)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNITS
AC CHARACTERISTICS						
Input Supply Ripple Rejection	PSRR	$f = 1kHz, I_{LOAD} = 1A, V_{IN} = 5.0V, V_{OUT} = 3.3V$		48		dB
		$f = 120Hz, I_{LOAD} = 1A, V_{IN} = 5.0V, V_{OUT} = 3.3V$		48		dB
Output Noise Voltage		$I_{LOAD} = 10mA, BW = 300Hz < f < 300kHz, V_{IN} = 3.7, V_{OUT} = 3.3V$		100		μV_{RMS}
		$V_{IN} = 2.2V, V_{OUT} = 1.8V, I_{LOAD} = 1A, BW = 100Hz < f < 100kHz$		53		μV_{RMS}
ENABLE PIN CHARACTERISTICS						
Turn-on Threshold	$V_{EN(HIGH)}$		0.5	0.8	1.0	V
Hysteresis	$V_{EN(HYS)}$		10	80	200	mV
ENABLE Pin Turn-on Delay	t_{EN}	$C_{OUT} = 10\mu F, I_{LOAD} = 1A$		80		μs
ENABLE Pin Leakage Current		$V_{IN} = 6V, ENABLE = 3V$			1	μA
SOFT-START CHARACTERISTICS						
Reset Pull-down Current	I_{PD}	$V_{IN} = 3.5V, EN = 0V, SS = 1V$	0.5	1	1.3	mA
Soft-start Charge Current	I_{CHG}		-3.3	-2	-0.8	μA
PG PIN CHARACTERISTICS						
V_{OUT} PG Flag Threshold			75	84	92	% V_{OUT}
V_{OUT} PG Flag Hysteresis				4		%
PG Flag Low Voltage		$I_{SINK} = 500\mu A$		47	100	mV
PG Flag Leakage Current		$V_{IN} = 6V, PG = 6V$		0.05	1	μA

NOTES:

- Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.
- Dropout is defined by the difference in supply V_{IN} and V_{OUT} when the output is below its nominal regulation.

Typical Operating Performance

Unless otherwise noted: $V_{IN} = 5V$, $V_{OUT} = 3.3V$, $C_{IN} = C_{OUT} = 10\mu F$,

$T_J = +25^\circ C$, $I_L = 0A$.

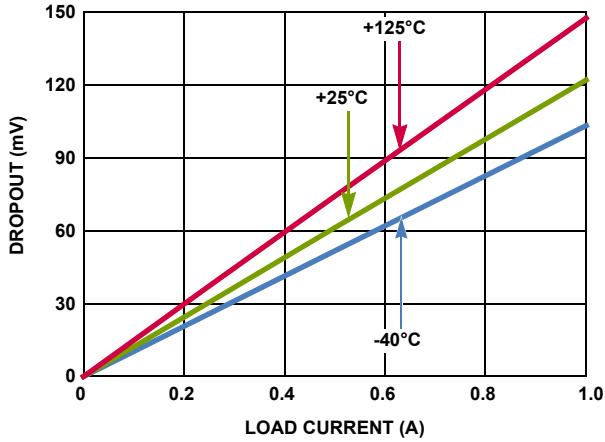


FIGURE 2. DROPOUT vs LOAD

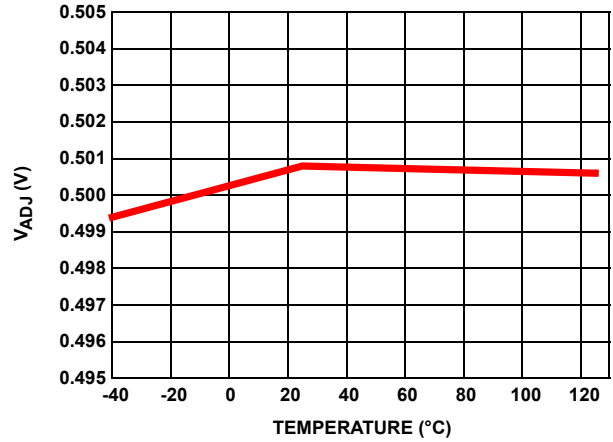


FIGURE 3. V_{ADJ} vs TEMPERATURE

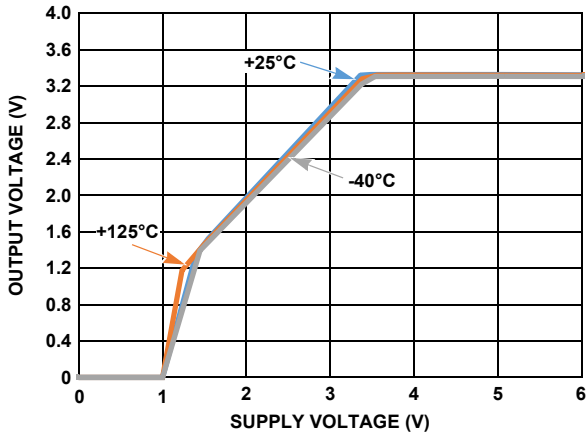


FIGURE 4. OUTPUT VOLTAGE vs SUPPLY VOLTAGE

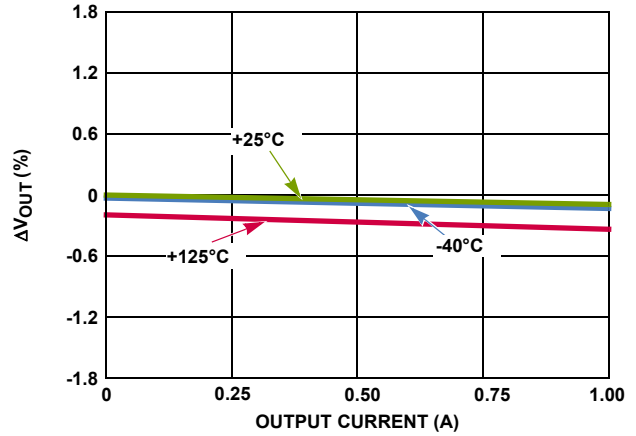


FIGURE 5. OUTPUT VOLTAGE vs OUTPUT CURRENT

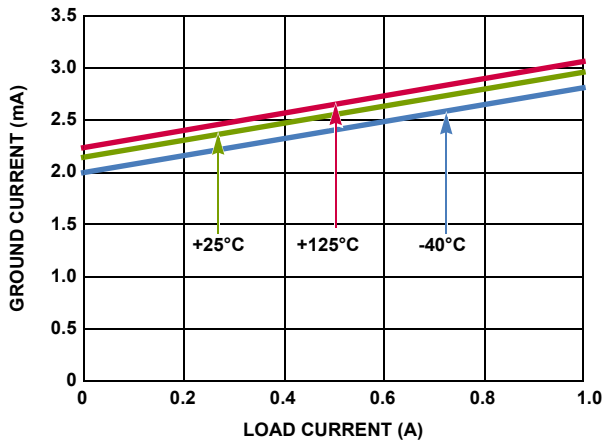


FIGURE 6. GROUND CURRENT vs LOAD CURRENT

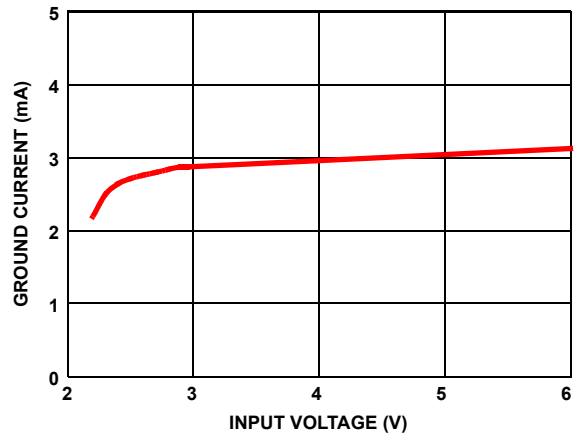


FIGURE 7. GROUND CURRENT vs SUPPLY VOLTAGE

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Typical Operating Performance

Unless otherwise noted: $V_{IN} = 5V$, $V_{OUT} = 3.3V$, $C_{IN} = C_{OUT} = 10\mu F$,
 $T_J = +25^\circ C$, $I_L = 0A$. (Continued)

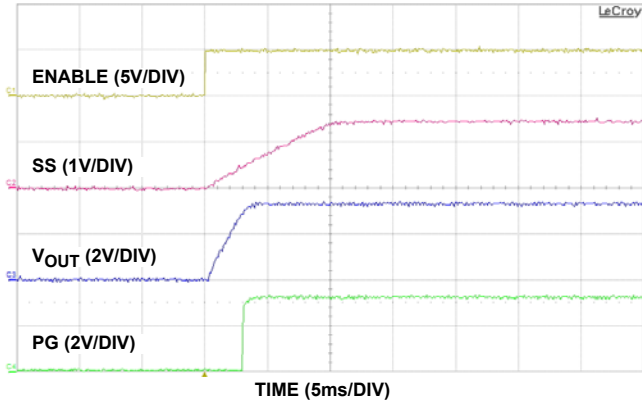


FIGURE 8. ENABLE START-UP

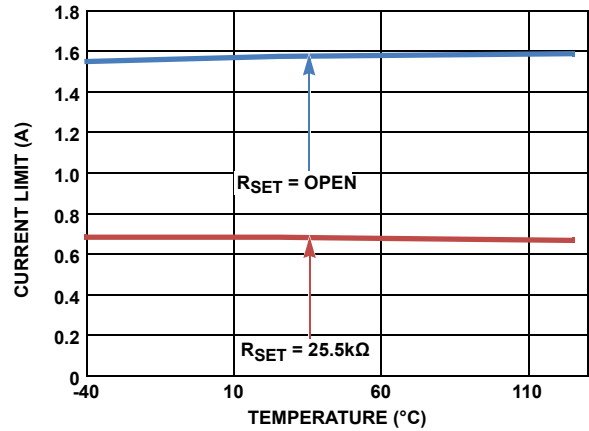


FIGURE 9. CURRENT LIMIT vs TEMPERATURE

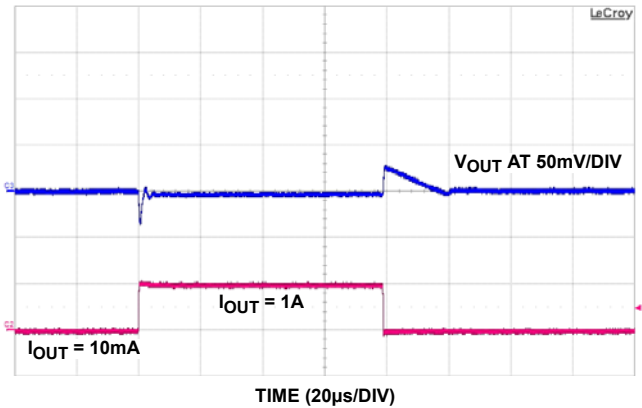


FIGURE 10. LOAD TRANSIENT RESPONSE

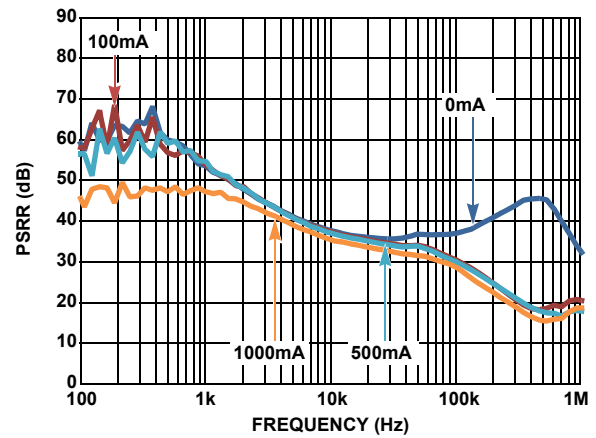


FIGURE 11. PSRR vs FREQUENCY

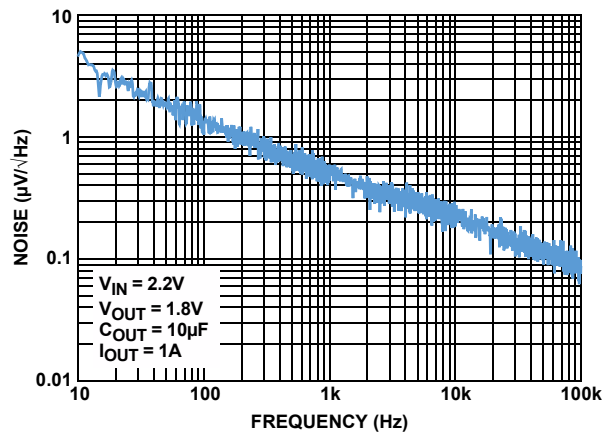


FIGURE 12. OUTPUT NOISE SPECTRAL DENSITY

Functional Description

Input Voltage Requirements

ISL80101A is capable of delivering output voltages from 0.8V to 5.0V. Due to the nature of an LDO, V_{IN} must be some margin higher than V_{OUT} plus dropout at the maximum rated current of the application if active filtering (PSRR) is expected from V_{IN} to V_{OUT} . The generous dropout specification of this family of LDOs allows applications to design for a level of efficiency that can accommodate profiles smaller than the TO220/263.

Programmable Current Limit

The ISL80101A protects against overcurrent due to short circuit and overload conditions applied to the output. When this happens, the LDO performs as a constant current source. If the short circuit or overload condition is removed, the output returns to normal voltage regulation operation.

The current limit is set at 1.62A by default when the I_{SET} pin is left floating.

This limit can be increased by tying a resistor R_{SET} from the I_{SET} pin to ground. The current limit is determined by R_{SET} as shown in Equation 1. Do not short this pin to ground. Increasing the current limit past 1.75A may cause damage to the part and is highly discouraged.

$$I_{LIMIT} \sim 1.62 + \frac{2.9}{R_{SET}(k\Omega)} \quad (EQ. 1)$$

The current limit can be decreased from the 1.62A default by tying R_{SET} from the I_{SET} pin to V_{IN} . The current limit is then determined by both R_{SET} and V_{IN} following Equation 2.

$$I_{LIMIT} \sim 1.62 - \frac{2.9 \times 2 \times V_{IN} - 1}{R_{SET}k\Omega} \quad (EQ. 2)$$

Figure 13 shows the relationship between R_{SET} and the current limit when R_{SET} is tied from the I_{SET} pin to V_{IN} for various V_{IN} values..

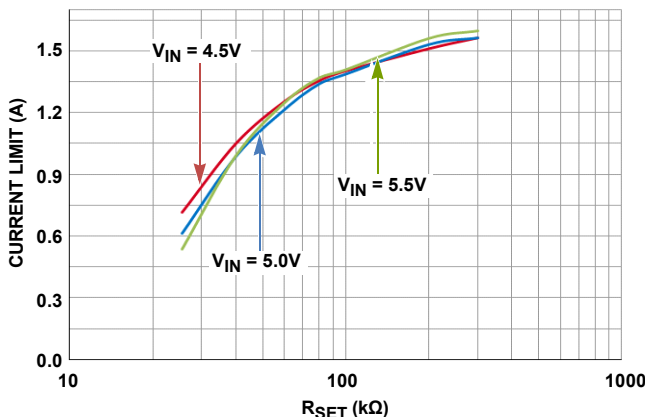


FIGURE 13. CURRENT LIMIT vs R_{SET} AT DIFFERENT V_{IN}

Enable Operation

The ENABLE turn-on threshold is typically 800mV with 80mV of hysteresis. An internal pull-up or pull-down resistor to change these values is available upon request. As a result, this pin must

not be left floating, and should be tied to V_{IN} if not used. A 1kΩ to 10kΩ pull-up resistor is required for applications that use open collector or open drain outputs to control the ENABLE pin. The ENABLE pin may be connected directly to V_{IN} for applications with outputs that are always on.

Power-good Operation

PG is a logic output that indicates the status of V_{OUT} , current limit tripping, and V_{IN} . The PG flag is an open-drain NMOS that can sink up to 10mA during a fault condition. The PG pin requires an external pull-up resistor typically connected to the V_{OUT} pin. The PG pin should not be pulled up to a voltage source greater than V_{IN} . PG goes low when the output voltage drops below 84% of the nominal output voltage, the current limit faults, or the input voltage is too low. PG functions during shutdown, but not during thermal shutdown. For applications not using this feature, connect this pin to ground.

Soft-start Operation

The soft-start circuit controls the rate at which the output voltage rises up to regulation at power-up or LDO enable. This start-up ramp time can be set by adding an external capacitor from the SS pin to ground. An internal 2μA current source charges up this C_{SS} and the feedback reference voltage is clamped to the voltage across it. The start-up time is set by Equation 3.

$$T_{start} = \frac{(C_{SS} \times 0.5)}{2\mu A} \quad (EQ. 3)$$

Equation 4 determines the C_{SS} required for a specific start-up inrush current, where V_{OUT} is the output voltage, C_{OUT} is the total capacitance on the output and I_{INRUSH} is the desired inrush current.

$$C_{SS} = \frac{(V_{OUT} \times C_{OUT} \times 2\mu A)}{I_{INRUSH} \times 0.5V} \quad (EQ. 4)$$

The external capacitor is always discharged to ground at the beginning of start-up or enabling.

Output Voltage Selection

An external resistor divider is used to scale the output voltage relative to the internal reference voltage. This voltage is then fed back to the error amplifier. The output voltage can be programmed to any level between 0.8V and 5V. An external resistor divider, R_2 and R_3 , is used to set the output voltage as shown in Equations 5 and 6. Please see Table 2 on page 9 for recommended values of R_2 and R_3 .

$$V_{OUT} = 0.5V \times \left(\frac{R_3}{R_2} + 1 \right) \quad (EQ. 5)$$

$$R_3 = R_2 \times \left(\frac{V_{OUT}}{0.5V} - 1 \right) \quad (EQ. 6)$$

External Capacitor Requirements

External capacitors are required for proper operation. Careful attention must be paid to the layout guidelines and selection of capacitor type and value to ensure optimal performance.

OUTPUT CAPACITOR

The ISL80101A applies state-of-the-art internal compensation to keep the selection of the output capacitor simple for the customer. Stable operation over full temperature, V_{IN} range, V_{OUT} range and load extremes are guaranteed for all capacitor types and values assuming a minimum of $10\mu\text{F}$ X5R/X7R is used for local bypass on V_{OUT} . This output capacitor must be connected to the V_{OUT} and GND pins of the LDO with PCB traces no longer than 0.5cm.

There is a growing trend to use very-low ESR multilayer ceramic capacitors (MLCC) because they can support fast load transients and also bypass very high frequency noise from other sources. However, the effective capacitance of MLCCs drops with applied voltage, age and temperature. X7R and X5R dielectric ceramic capacitors are strongly recommended as they typically maintain a capacitance range within $\pm 20\%$ of nominal voltage over full operating ratings of temperature and voltage.

Additional capacitors of any value in ceramic, POSCAP, alum/tantalum electrolytic types may be placed in parallel to improve PSRR at higher frequencies and/or load transient AC output voltage tolerances.

Phase Boost Capacitor

A small phase boost capacitor, C_{PB} , can be placed across the top resistor, R_3 , in the feedback resistor divider network in order to place a zero at:

$$F_z = \frac{1}{2\pi R_3 C_{PB}} \quad (\text{EQ. 7})$$

This zero increases the crossover frequency of the LDO and provides additional phase resulting in faster load transient response.

It is also important to note that the LDO stability and load transient are affected by the type of output capacitor used. For optimal result, empirical tuning is suggested for each specific application.

Table 2 shows the recommended C_{PB} , R_3 and R_2 for different output voltage and ceramic C_{OUT} .

TABLE 2. RECOMMENDED C_{PB} FOR DIFFERENT V_{OUT} AND C_{OUT}

V_{OUT} (V)	R_3 (k Ω)	R_2 (k Ω)	C_{OUT} (μF)	C_{PB} (pF)
5.0	2.61	0.287	10	100
3.3	2.61	0.464	10	100
2.5	2.61	0.649	10	82
1.8	2.61	1.0	10	82
1.5	2.61	1.3	10	68
1.5	2.61	1.3	22	150
1.2	2.61	1.87	22	120
1.2	2.61	1.87	47	270
1.0	2.61	2.61	47	220
0.8	2.61	4.32	47	220

INPUT CAPACITOR

For proper operation, a minimum capacitance of $10\mu\text{F}$ X5R/X7R is required at the input. This ceramic input capacitor must be connected to the V_{IN} and GND pins of the LDO with PCB traces no longer than 0.5cm.

Power Dissipation and Thermals

The junction temperature must not exceed the range specified in the "Recommended Operating Conditions (Note 7)" on page 4. The power dissipation can be calculated by using Equation 8:

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND} \quad (\text{EQ. 8})$$

The maximum allowable junction temperature, $T_{J(MAX)}$ and the maximum expected ambient temperature, $T_{A(MAX)}$ determine the maximum allowable power dissipation, as shown in Equation 9:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA} \quad (\text{EQ. 9})$$

θ_{JA} is the junction-to-ambient thermal resistance.

For safe operation, ensure that the power dissipation P_D , calculated from Equation 8, is less than the maximum allowable power dissipation $P_{D(MAX)}$.

The DFN package uses the copper area on the PCB as a heat-sink. The EPAD of this package must be soldered to the copper plane (GND plane). Figure 14 shows a curve for the θ_{JA} of the DFN package for different copper area sizes.

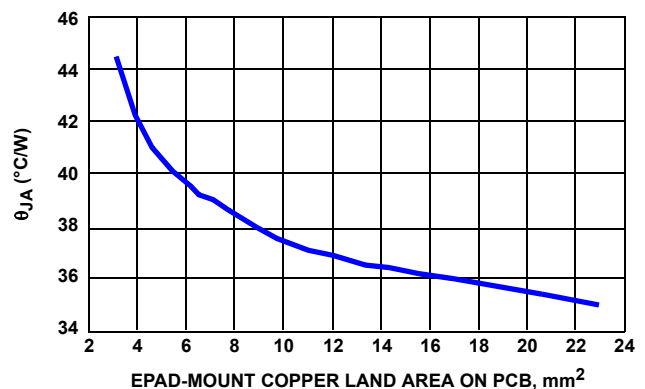


FIGURE 14. 3mmx3mm 10 LD DFN ON 4-LAYER PCB WITH THERMAL VIAS θ_{JA} vs EPAD-MOUNT COPPER LAND AREA ON PCB

Thermal Fault Protection

The power level and the thermal impedance of the package (+48°C/W for DFN) determine when the junction temperature exceeds the thermal shutdown temperature. In the event that the die temperature exceeds around +160°C, the output of the LDO will shut down until the die temperature cools down to about +130°C.

General PowerPAD Design Considerations

Figure 15 shows the recommended use of vias on the thermal pad to remove heat from the IC. This typical array populates the thermal pad footprint with vias spaced three times the radius distance from the center of each via. Small via size is advisable, but not to the extent that solder reflow becomes difficult.

All vias should be connected to the pad potential, with low thermal resistance for efficient heat transfer. Complete connection of the plated-through hole to each plane is important. It is not recommended to use “thermal relief” patterns to connect the vias.

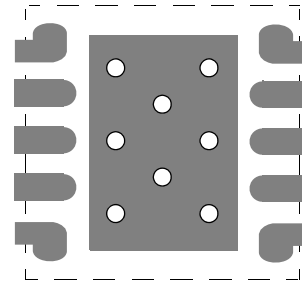


FIGURE 15. PCB VIA PATTERN

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest revision.

DATE	REVISION	CHANGE
August 11, 2015	FN7712.4	<p>Figure 1 on page 1 - updated equation.</p> <p>Removed Sense Voltage Version from “Block Diagram” on page 2</p> <p>Changed PAD to EPAD in Pin Configuration on page 3</p> <p>“Pin Descriptions” on page 3 - V_{OUT} pin - added after the word capacitor: “(for V_{OUT} from 1.5V to 5V)”.</p> <p>Changed “SENSE” to “ADJ” under “Absolute Maximum Ratings (Note 6)” and “Recommended Operating Conditions (Note 7)” on page 4</p> <p>Changed in “Absolute Maximum Ratings (Note 6)” on page 4 - Latch-up temp from: +85 °C to +125 °C</p> <p><u>Electrical Spec Table changes beginning on page 4:</u></p> <p>Electrical Spec table conditions changed from: V_{IN} = V_{OUT} + 0.4V, V_{OUT} = 3.3V, C_{IN} = C_{OUT} = 10μF, T_J = +25 °C, I_{LOAD} = 0A, to: 2.2V < V_{IN} < 6V, V_{OUT} = 0.5V, T_J = +25 °C, I_{LOAD} = 0A</p> <p>DC ADJ Pin Voltage Accuracy - changed test conditions from: V_{OUT} + 0.4V < V_{IN} < 6V; 0A < I_{LOAD} < 1A to: V_{OUT} + 0.4V < V_{IN} < 6V, V_{OUT} = 2.5V; 0A < I_{LOAD} < 1A</p> <p>DC Input Line Regulation - changed test conditions from: V_{OUT} + 0.4V < V_{IN} < 6.0V, V_{OUT} = 5.0V to: V_{OUT} + 0.4V < V_{IN} < 6V, V_{OUT} = 2.5V. Added “-1” MIN</p> <p>DC Output Load Regulation - Test Conditions added: V_{OUT} + 0.4V < V_{IN} < 6V, V_{OUT} = 2.5V Added “1” MAX</p> <p>Ground Pin Current - changed Test Conditions from: I_{LOAD} = 0A, 2.2V < V_{IN} < 6V to: I_{LOAD} = 0A, V_{OUT} + 0.4V < V_{IN} < 6V, V_{OUT} = 2.5V. And from: I_{LOAD} = 1A, 2.2V < V_{IN} < 6V to: I_{LOAD} = 1A, V_{OUT} + 0.4V < V_{IN} < 6V, V_{OUT} = 2.5V</p> <p>Dropout voltage test condition: changed “VSENSE = 0V” to “Vadj = 0”</p> <p>Output Current Limit changed Test Conditions</p> <p>from: V_{OUT} = 2V, 4.5V < V_{IN} < 5.5V, I_{SET} is floating; to: 4.5V < V_{IN} < 6V, I_{SET} is floating</p> <p>and from: V_{OUT} = 2V, V_{IN} = 5.0V, R_{SET} = 25.5kΩ to: V_{IN} = 5V, R_{SET} = 25.5kΩ</p> <p>Thermal Shutdown Temperature - removed Test Conditions</p> <p>Thermal Shutdown Hysteresis - Removed “Rising Threshold”. Removed Test Conditions</p> <p>PSRR - added V_{out} = 3.3V to both test conditions</p> <p>Output Noise Voltage - added V_{IN} = 3.7, V_{OUT} = 3.3V to the first test conditions</p> <p>Turn-on Threshold: Removed Test Conditions. Changed MIN “0.3” to “0.5”</p> <p>Hysteresis - Removed “Rising Threshold”. Removed Test Conditions</p> <p>Reset Pull-down Current - Changed Test Condition from: V_{IN} = 5.4V, ENABLE = 0V, SS = 1V; to: V_{IN} = 3.5V, EN = 0V, SS = 1V</p> <p>Page 6-replaced/updated Figure 4 “OUTPUT VOLTAGE vs SUPPLY VOLTAGE” with new. Added Figure 12 on page 7.</p> <p>Updated Equations 1 and 2 on page 8</p> <p>Updated POD L10.3x3 on page 13 to most recent revision with changes as follows:</p> <p>Removed package outline and included center to center distance between lands on recommended land pattern.</p> <p>Removed Note 4 “Dimension b applies to the metallized terminal and is measured between 0.18mm and 0.30mm from the terminal tip.” since it is not applicable to this package. Renumbered notes accordingly</p> <p>Added missing dimension 0.415 in Typical Recommended land pattern.</p> <p>Shortened the e-pad rectangle on both the recommended land pattern and the package bottom view to line up with the centers of the corner pins.</p> <p>Removed former Note 4: Lead width applies to the metallized terminal and is measured between 0.18mm and 0.30mm from the terminal tip.</p> <p>Updated tiebar note From: Tiebar shown (if present) is a non-functional feature.</p> <p>To: Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends).</p>

Revision History

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DATE	REVISION	CHANGE
September 19, 2011	FN7712.3	Table 1 on page 1 updated to include more information on Intersil's 1A LDO portfolio. Added standard MSL Note to "Ordering Information" (Note 3)
February 2, 2011	FN7712.2	<ol style="list-style-type: none"> 1. On page 1, "Features" <ol style="list-style-type: none"> a. "±1.8% Vout Accuracy Guaranteed..." changed to "±2% Vadj Accuracy Guaranteed..." 2. Figure 1 on page 1 <ol style="list-style-type: none"> a. "Typical Applications" changed to "Typical Application" b. "82pF" for Cpb changed to "100pF" 3. On page 3, Pin Number 8 <ol style="list-style-type: none"> a. On "Description" of ISET, change 2nd sentence from "Current limit is 0.75mA when..." to "Current limit is 1.62A when..." 4. On page 4, "Electrical Specifications" <ol style="list-style-type: none"> a. "DC Input Line Regulation" given own line, added symbol, and changed test conditions b. "Feedback Input Current", added typical "0.01" and max "1" with units "µA" 5. On page 5, "Electrical Specifications" <ol style="list-style-type: none"> a. "PG PIN CHARACTERISTICS" "VOUT PG Flag Threshold", Typical "85" changed to "84" %Vout 7. On page 8, "Programmable Current Limit" <ol style="list-style-type: none"> a. Equation 1 changed to "Ilimit=1.62+..." b. Equation 2 changed to "Ilimit=1.62-..." 8. Added "The current limit can be decreased from the 0.75A default..." changed to "The current limit can be decreased from the 1.62A default..." on page 8, between Equation 1 and Equation 2 9. On page 8, beginning of last paragraph <ol style="list-style-type: none"> a. "Figure 11 shows the relationship..." changed to "Figure 13 shows the relationship..." 10. "External Capacitor Requirements" on page 8: <ol style="list-style-type: none"> a. "The ISL80121-5 applies..." changed to "The ISL80101A applies..." 11. On page 4, "Electrical Specifications", "DC CHARACTERISTICS", "Output Current Limit" <ol style="list-style-type: none"> a. "VOUT = 2V, VIN = 5.5V, RSET = 25.5k" changed to "VOUT = 2V, VIN = 5.0V, RSET = 25.5k" 12. On page 4, "Electrical Specifications", "AC CHARACTERISTICS", "Input Supply Ripple Rejection" <ol style="list-style-type: none"> a. "58db" typical changed to "48" b. "62dB" typical changed to "48" 13. On page 8, revised Figure 13. Updated same graphic on page 1 14. Throughout: All "VIN" changed to "VIN" 15. Throughout: All "VOUT" changed to "VOUT" 16. Throughout: All "RSET" changed to "RSET" 17. Throughout: All "ISET" changed to "ISET" 18. Throughout: All "EN" and "enable" changed to "ENABLE" 19. Throughout: All "PGOOD" changed to "PG" 20. "Block Diagram" on page 2, subscripted pin names for VIN, VOUT, ISET. Changed PGOOD to PG 21. On page 3, EPAD Description <ol style="list-style-type: none"> a. "directly to GND plane is optional." Changed to "directly to GND plane is required for thermal considerations. See "Power Dissipation and Thermals" on page 9 for more details." 22. On page 1, in paragraph 2, "The programmable current limiting improves system reliability of applications" changed to "The programmable current limiting improves system reliability of end applications." 23. On page 1, "Features", "Programmable Soft-starting" changed to "Programmable Soft-Start" 24. On page 4, "Electrical Specifications", "DC CHARACTERISTICS", "DC Output Voltage Accuracy" changed to "DC ADJ Pin Voltage Accuracy" 25. On page 5, Notes 10 and 11 deleted (they were not referenced in the spec table). 26. "Output Voltage Selection" on page 8, "An external resistor divider, R2 and R3, is used to set the output voltage as shown in Equation 5. The recommended value for R3 is 500Ω to 1kΩ. R2 is then chosen according to Equation 6." changed to "An external resistor divider, R2 and R3, is used to set the output voltage as shown in Equations 5 and 6. Please see Table 2 on page 9 for recommended values of R2 and R3." 29. Added "General PowerPAD Design Considerations" on page 10 30. Revised Figure 8
December 6, 2010	FN7712.1	Modified "Block Diagram" on page 2. In "Ground Pin Current" on page 4 Test Conditions: -Changed 1st line from "VOUT + 0.4V < VIN < 5V, VSENSE = 0V" to "ILOAD = 0A, 2.2V < VIN < 6V" -Changed 2nd line from "VOUT + 0.4V < VIN < 6V, VSENSE = 0V" to "ILOAD = 1A, 2.2V < VIN < 6V" Figure 2 "DROPOUT vs LOAD" on page 6: -Switched colors on 25 °C and 125 °C.
November 29, 2010	FN7712.0	Initial Release

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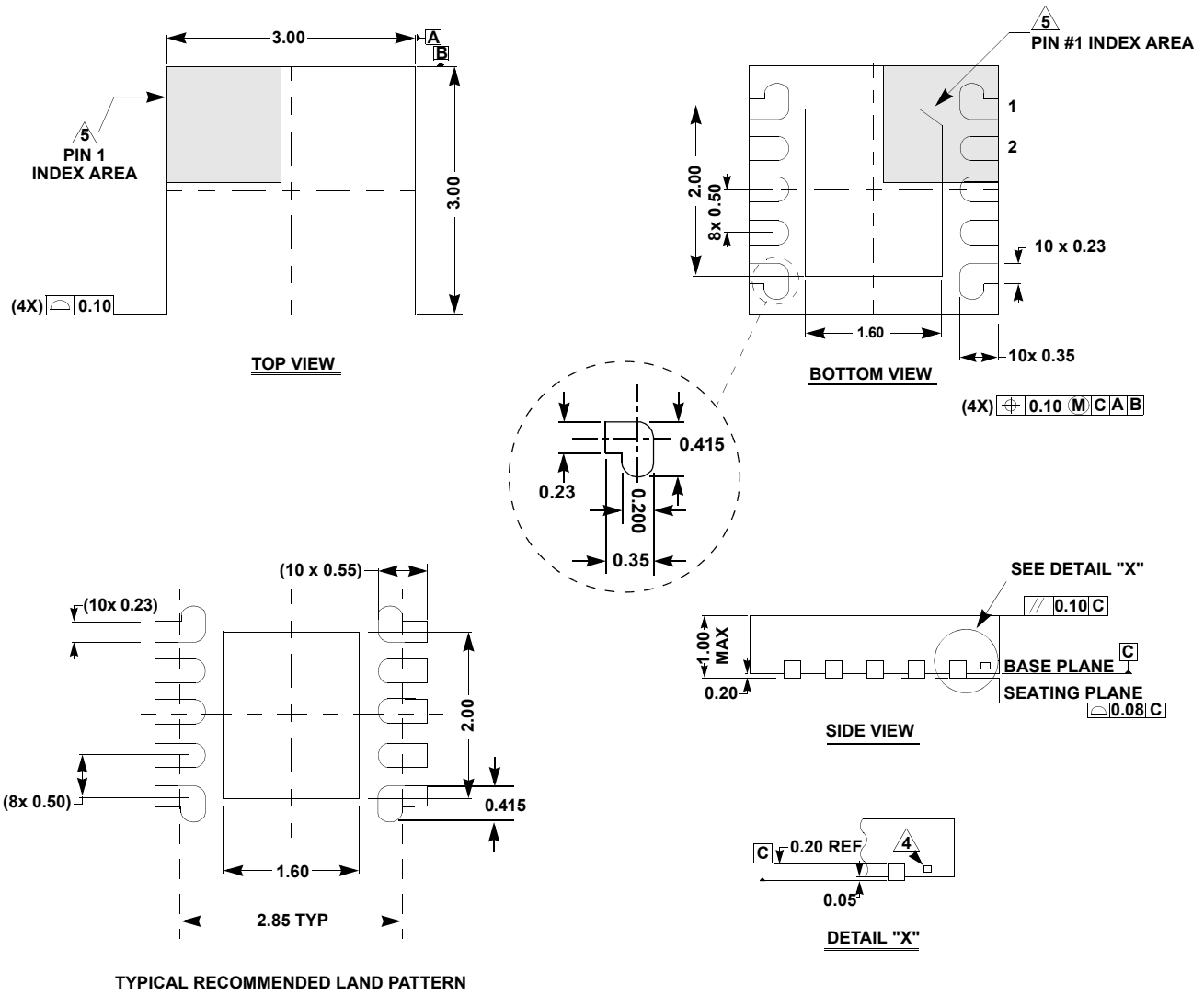
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Package Outline Drawing

L10.3x3

10 LEAD DUAL FLAT PACKAGE (DFN)

Rev 11, 3/15



NOTES:

1. Dimensions are in millimeters.
Dimensions in () for Reference Only.
2. Dimensioning and tolerancing conform to ASME Y14.5m-1994.
3. Unless otherwise specified, tolerance : Decimal ± 0.05
4. Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends).
5. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.

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Contact Us :

➤ Address :

401 Building No.5, JiuGe Business Center, Lane 2301, Yishan Rd
Minhang District, Shanghai , China

➤ Sales :

Direct +86 (21) 6401-6692

Email amall@ameya360.com

QQ 800077892

Skype [ameyasales1](#) [ameyasales2](#)

➤ Customer Service :

Email service@ameya360.com

➤ Partnership :

Tel +86 (21) 64016692-8333

Email mkt@ameya360.com