



MULTIPLIER-DIVIDER

FEATURES

- LOW COST
- DIFFERENTIAL INPUT
- ACCURACY 100% TESTED AND GUARANTEED
- LOW NOISE
120 μ V, rms, 10Hz to 10kHz
- SELF-CONTAINED
No additional amplifiers
- SMALL SIZE
Hermetic TO-100 package
- WIDE TEMPERATURE OPERATION

APPLICATIONS

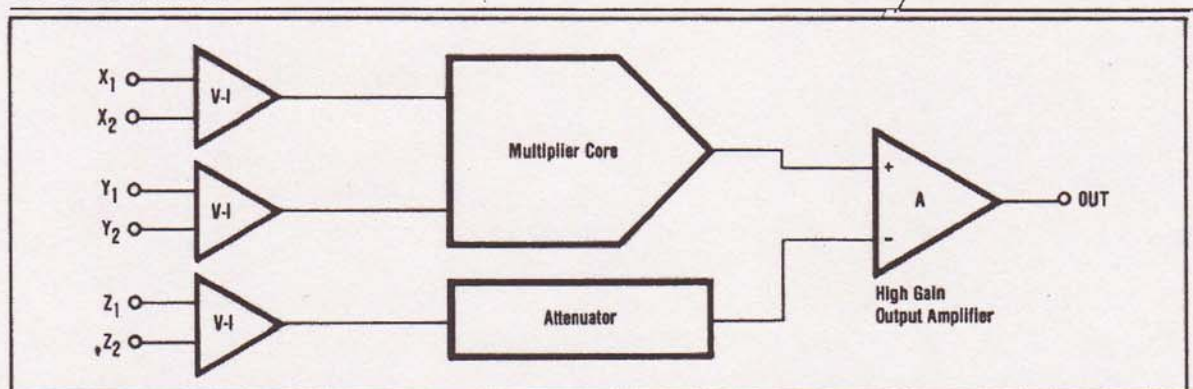
- MULTIPLICATION
- DIVISION
- SQUARING
- SQUARE ROOT
- LINEARIZATION
- POWER COMPUTATION
- ANALOG SIGNAL PROCESSING
- ALGEBRAIC COMPUTATION
- TRUE RMS-TO-DC CONVERSION

DESCRIPTION

The 4213 multiplier-divider is a low cost precision device designed for general purpose application. In addition to four-quadrant multiplication, it also performs analog square root and division without the bother of external amplifiers. The 4213 is laser-trimmed to guarantee its rated accuracy with no

external components. The internal zener regulated references make the 4213 much less sensitive to supply variation than earlier IC multipliers. Hermetic TO-100 package, wide operating temperature range, low output noise, and low cost are some of the desirable features of this versatile device.

4213 FUNCTIONAL DIAGRAM



SPECIFICATIONS

ELECTRICAL

Specifications at $T_A = +25^\circ\text{C}$ and $\pm V_{CC} = 15\text{VDC}$ unless otherwise noted.

MODEL		4213AM			4213BM			4213SM			
PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
MULTIPLIER PERFORMANCE											
Transfer Function		$\frac{(X_1 - X_2)(Y_1 - Y_2)}{10} + Z_2$.			.		
Total Error	$-10V \leq X, Y \leq 10V$.			.		
Initial	$T_A = +25^\circ C$			± 1.0			± 0.5			± 0.5	% FSR
vs Temperature	$-25^\circ C \leq T_A \leq +85^\circ C$		± 0.008	± 0.02		.	.		--	--	% FSR/°C
vs Temperature	$-55^\circ C \leq T_A \leq +125^\circ C$		--	--		--	--		± 0.025	± 0.05	% FSR/°C
vs Supply			± 0.05	--		.	--		.	--	% FSR/%
Individual Errors						.			.		
Output Offset						.			.		
Initial	$T_A = +25^\circ C$		± 10	± 50		± 7	± 25		± 7	± 25	mV
vs Temperature	$-25^\circ C \leq T_A \leq +85^\circ C$		± 0.7	± 2.0		± 0.3	± 0.7		--	--	mV/°C
vs Temperature	$-55^\circ C \leq T_A \leq +125^\circ C$		--	--		--	--		± 0.3	± 0.7	mV/°C
vs Supply			± 0.25	--		.	--		.	--	mV/%
Scale Factor Error						.			.		
Initial	$T_A = +25^\circ C$		± 0.12	--		.	--		--	--	% FSR
vs Temperature	$-25^\circ C \leq T_A \leq +85^\circ C$		± 0.008	--		.	--		--	--	% FSR/°C
vs Temperature	$-55^\circ C \leq T_A \leq +125^\circ C$		--	--		--	--		± 0.008	± 0.008	% FSR/°C
vs Supply			± 0.05	--		.	--		.	--	% FSR/%
Nonlinearity						.			.		
X Input	$X = 20V, p-p; Y = \pm 10VDC$		± 0.08	--		.	--		.	--	% FSR
Y Input	$Y = 20V, p-p; X = \pm 10VDC$		± 0.01	--		.	--		.	--	% FSR
Feedthrough	$f = 50Hz$.			.		
X Input	$X = 20V, p-p; Y = 0$		30	--		.	--		.	--	mV, p-p
Y Input	$Y = 20V, p-p; X = 0$		6	--		.	--		.	--	mV, p-p
vs Temperature	$-25^\circ C \leq T_A \leq +85^\circ C$		0.1	--		.	--		--	--	mV, p-p/°C
vs Temperature	$-55^\circ C \leq T_A \leq +125^\circ C$		--	--		--	--		0.1	0.1	mV, p-p/°C
vs Supply			0.15	--		.	--		.	--	mV, p-p/%
DIVIDER PERFORMANCE											
Transfer Function	$X_1 > X_2$	$\frac{10(Z_1 - Z_2)}{(X_1 - X_2)} + Y_2$.			.		
Total Error (with external adjustments)	$X = -10V$.			.		
	$-10V \leq Z \leq +10V$		± 0.75	--		± 0.35	--		± 0.35	--	% FSR
	$X = -1V$.			.		
	$-1V \leq Z \leq +1V$		± 2.0	--		± 1.0	--		± 1.0	--	% FSR
	$-10V \leq X \leq -0.2V$.			.		
	$-10V \leq Z \leq +10V$		± 5.0	--		± 1.0	--		± 1.0	--	% FSR
SQUARER PERFORMANCE											
Transfer Function		$\frac{(X_1 - X_2)^2}{10} + Z_2$.			.		
Total Error	$-10V \leq X \leq +10V$		± 0.6	--		± 0.3	--		± 0.3	--	% FSR
SQUARE-ROOTER PERFORMANCE											
Transfer Function	$Z_1 < Z_2$	$+\sqrt{10(Z_2 - Z_1)}$.			.		
Total Error	$1V \leq Z \leq 10V$		± 1	--		± 0.5	--		± 0.5	--	% FSR
AC PERFORMANCE											
Small-Signal Bandwidth	$\pm 3dB$		550	--		.	--		.	--	kHz
1% Amplitude Error	Small Signal		70	--		.	--		.	--	kHz
1% (0.57°) Vector Error	Small Signal		5	--		.	--		.	--	kHz
Full Power Bandwidth	$ V_d = 10V, R_L = 2k\Omega$		320	--		.	--		.	--	kHz
Slew Rate	$ V_d = 10V, R_L = 2k\Omega$		20	--		.	--		.	--	V/μsec
Settling Time	$\epsilon = \pm 1\%, \Delta V_o = 20V$		2	--		.	--		.	--	μsec
Overload Recovery	50% Output Overload		0.2	--		.	--		.	--	μsec
INPUT CHARACTERISTICS											
Input Voltage Range		± 10			.			.			V
Rated Operation				$\pm V_{CC}$.			.			V
Absolute Maximum					.			.			V
Input Resistance	$X, Y, Z^{(1)}$		10	--		.	--		.	--	MΩ
Input Bias Current	X, Y, Z		1.4	--		.	--		.	--	μA
OUTPUT CHARACTERISTICS											
Rated Output					.			.			V
Voltage	$I_o = \pm 5mA$	± 10			.			.			V
Current	$V_o = \pm 10V$	± 5			.			.			mA
Output Resistance	$f = DC$		1.5	--		.	--		.	--	Ω
OUTPUT NOISE VOLTAGE											
	$X = Y = 0$.			.		
$f_o = 1Hz$			40	--		.	--		.	--	μV/√Hz
$f_o = 10kHz$			1.0	--		.	--		.	--	μV/√Hz
1/f Corner Frequency			1060	--		.	--		.	--	Hz
$f_B = 10Hz$ to $10kHz$			125	--		.	--		.	--	μV, rms
$f_B = 10Hz$ to $10MHz$			3	--		.	--		.	--	mV, rms
POWER SUPPLY REQUIREMENTS											
Rated Voltage			± 15	--		.	--		.	--	VDC
Operating Range	Derated Performance	± 8.5		± 20	VDC
Quiescent Current			± 5.5	--		.	--		.	--	mA

ELECTRICAL (CONT)

MODEL		4213AM			4213BM			4213SM			UNITS
PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
TEMPERATURE RANGE (Ambient)											
Specification	Derated Performance	-25		+85	*		*	-55		+125	°C
Operating Range		-55		+125	*		*	*		*	°C
Storage		-65		+150	*		*	*		*	°C

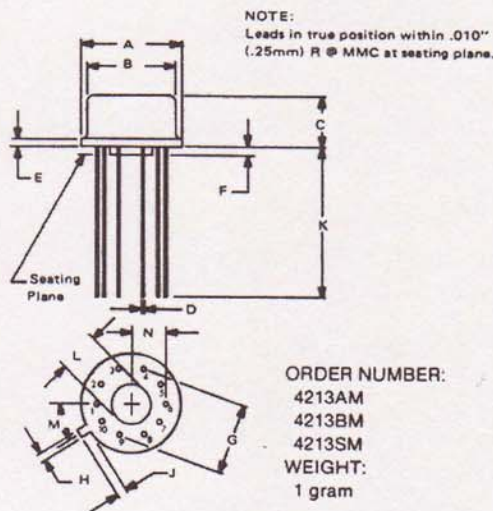
NOTES:

1. Z_2 input resistance is 10M Ω , typical, with Pin 9 open. If Pin 9 is grounded or used for optional offset adjustment, the Z_2 input resistance may be as low as 25k Ω .

*Same as 4213AM specification.

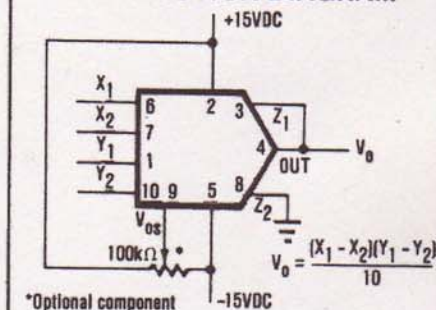
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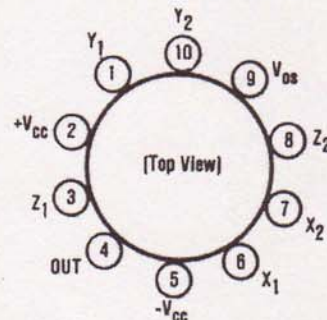


DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.335	.370	8.51	9.40
B	.305	.335	7.75	8.51
C	.165	.185	4.19	4.70
D	.016	.021	0.41	0.53
E	.010	.040	0.25	1.02
F	.010	.040	0.25	1.02
G	.230 BASIC		5.84 BASIC	
H	.028	.034	0.71	0.86
J	.029	.045	0.74	1.14
K	.500	—	12.70	—
L	.120	.160	3.05	4.06
M	36° BASIC		36° BASIC	
N	.110	.120	2.79	3.05

CONNECTION DIAGRAM



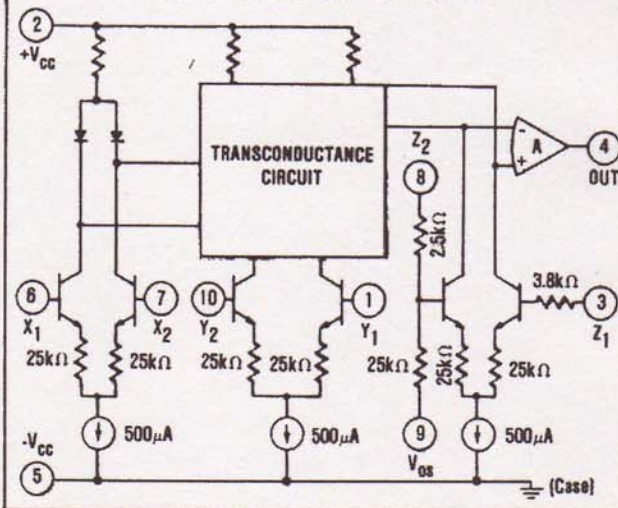
PIN CONFIGURATION



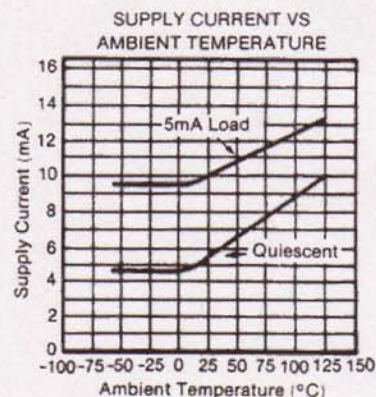
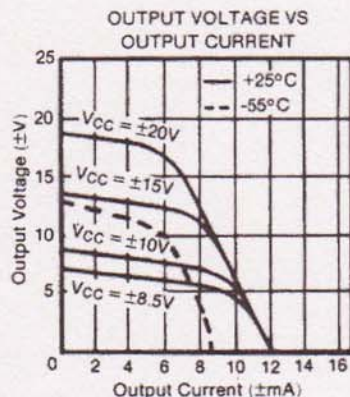
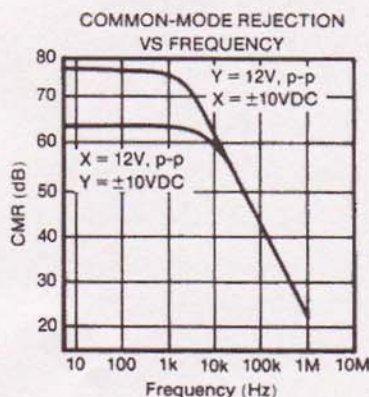
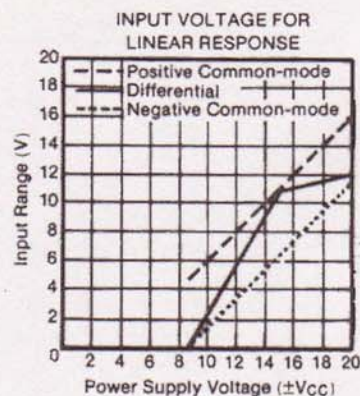
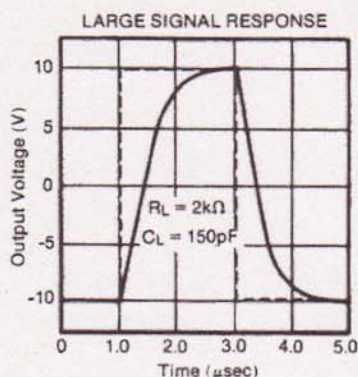
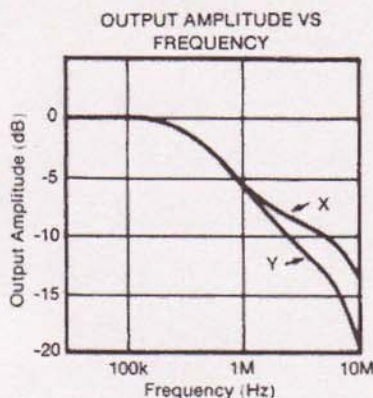
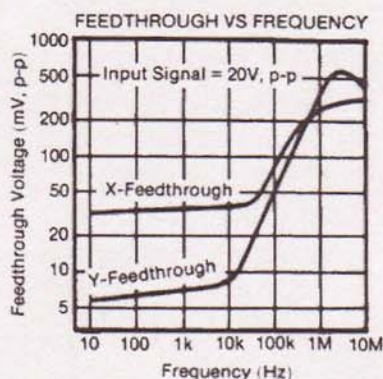
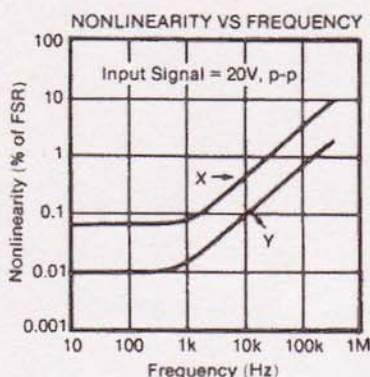
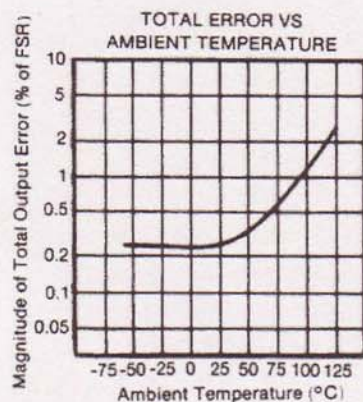
NOTES:

- V_{os} adjustment optional not normally recommended. V_{os} pin may be left open or grounded.
- All unused input pins should be grounded.
- Pin 5 is connected to the case.

SIMPLIFIED SCHEMATIC



TYPICAL PERFORMANCE CURVES



ABSOLUTE MAXIMUM RATINGS

Supply	±20VDC
Internal Power Dissipation ⁽¹⁾	500mW
Differential Input Voltage ⁽²⁾	±40VDC
Input Voltage Range ⁽²⁾	±20VDC
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	-55°C to +125°C
Lead Temperature (soldering, 10 seconds)	+300°C
Output Short-circuit Duration ⁽³⁾	Continuous
Junction Temperature	+150°C

NOTES:

- Package must be derated based on: $\theta_{JC} = 55^\circ\text{C/W}$ and $\theta_{JA} = 165^\circ\text{C/W}$.
- For supply voltages less than ±20VDC the absolute maximum input voltage is equal to the supply voltage.
- Short-circuit may be to ground only. Rating applies to +85°C ambient.

DEFINITIONS

TOTAL ERROR (Accuracy)

Total error is the actual departure of the multiplier output voltage from the ideal product of its input voltages. It includes the sum of the effects of input and output DC offsets, gain error and nonlinearity.

OUTPUT OFFSET

Output offset is the output voltage when both inputs V_X and V_Y are zero volts.

SCALE FACTOR ERROR

Scale factor error is the difference between the actual scale factor and the ideal scale factor.

NONLINEARITY

Nonlinearity is the maximum deviation from a best straightline (curve fitting on input-output graph) expressed as a percent of peak-to-peak full scale output.

FEEDTHROUGH

Feedthrough is the signal at the output for any value of V_X or V_Y within the rated range, when the other input is zero.

SMALL SIGNAL BANDWIDTH

Small signal bandwidth is the frequency at which the output is down 3dB from its low frequency value for a nominal output amplitude of 10% of full scale.

1% AMPLITUDE ERROR

The 1% amplitude error is the frequency the output amplitude is in error by 1%, measured with an output amplitude of 10% of full scale.

1% VECTOR ERROR

The 1% vector error is the frequency at which a phase error of 0.01 radians (0.57°) occurs. This is the most sensitive measure of dynamic error of a multiplier.

APPLICATIONS INFORMATION

MULTIPLICATION

Figure 1 shows the basic connection for four-quadrant multiplication.

The 4213 meets all of its specifications without trimming. Accuracy can, however, be improved by nulling the output offset voltage using the 100kΩ optional balance potentiometer shown in Figure 1.

AC feedthrough may be reduced to a minimum by applying an external voltage to the X or Y input as shown in Figure 2.

Z_2 , the optional summing input, may be used to sum a voltage into the output of the 4213. If not used, this

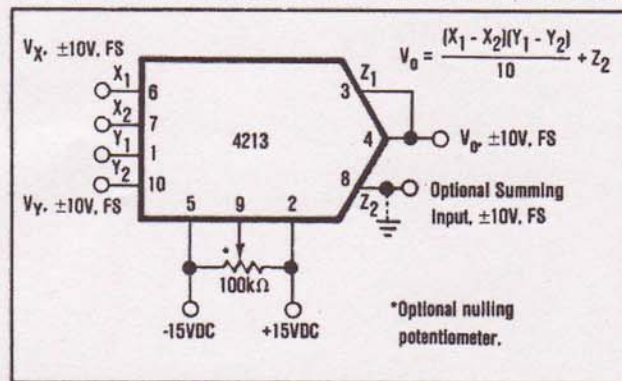


FIGURE 1 Multiplier Connection.

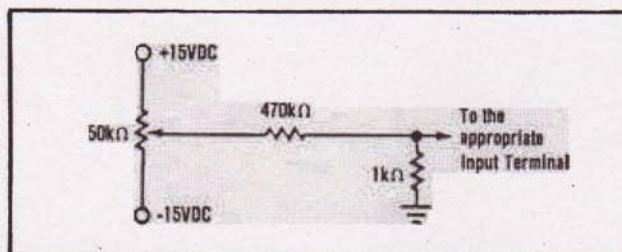


FIGURE 2. Optional Trimming Configuration.

terminal, as well as the X and Y input terminals, should be grounded. All inputs should be referenced to power supply common.

Figure 3 shows how to achieve a scale factor larger than the nominal 0.1. In this case, the scale factor is unity which makes the transfer function

$$V_o = K V_X V_Y = K (X_1 - X_2)(Y_1 - Y_2)$$

$$K = [1 + (R_1/R_2)] / 10$$

$$0.1 \leq K \leq 1$$

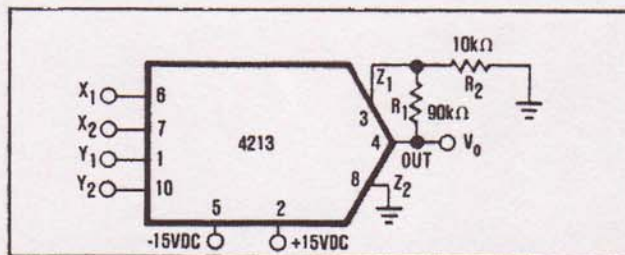


FIGURE 3. Connection For Unity Scale Factor.

This circuit has the disadvantage of increasing the output offset voltage by a factor of 10 which may require the use of the optional balance control for some applications. In addition, this connection reduces the small signal bandwidth to about 50kHz.

DIVISION

Figure 4 shows the basic connection for two-quadrant division. This configuration is a multiplier-inverted analog divider, i.e., a multiplier connected in the feedback loop of an operational amplifier. In the case of the 4213 this operational amplifier is the output amplifier of the multiplier itself.

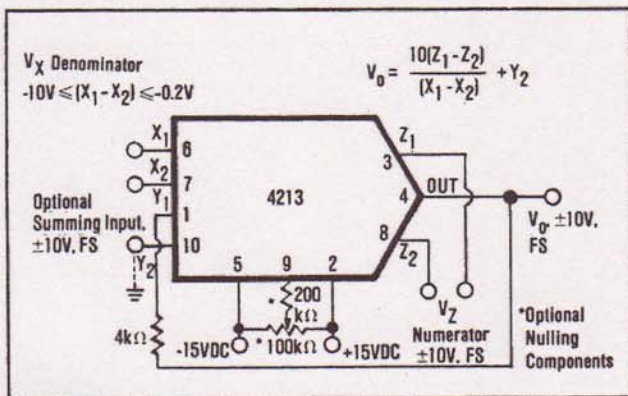


FIGURE 4. Divider Connection.

The divider error with a multiplier-inverted analog divider is approximately

$$\epsilon_{\text{divider}} = 10 \epsilon_{\text{multiplier}} / (X_1 - X_2).$$

It is obvious from this error equation that divider error becomes excessively large for small values of $X_1 - X_2$. A 10-to-1 denominator range is usually the practical limit. If more accurate division is required over a wide range of denominator voltages, an externally generated voltage may be applied to the unused X-input (see Optional Trim Configuration). To trim, apply a ramp of +100mV to +1V at 100Hz to both X_1 and Z_1 if X_2 is used for offset adjustment, otherwise reverse the signal polarity, and adjust the trim voltage to minimize the variation in the output. An alternative to this procedure would be to use the Burr-Brown DIV100, a precision log-antilog divider.

SQUARING

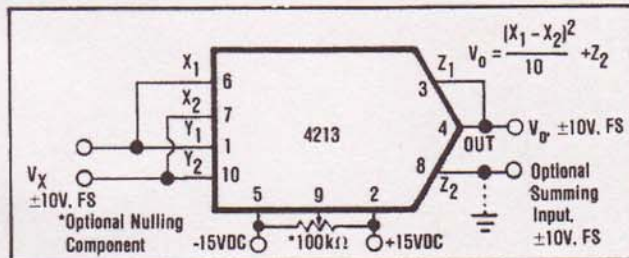


FIGURE 5. Squarer Connection.

SQUARE ROOT

Figure 6 shows the connection for taking the square root of the voltage V_Z . The diode prevents a latching condition which could occur if the input momentarily changed polarity. This latching condition is not a design flaw in the 4213, but occurs when a multiplier is connected in the feedback loop of an operational amplifier to perform square root functions.

The load resistance R_L must be in the range of $10k\Omega \leq R_L \leq 1M\Omega$. This resistance must be in the circuit as it provides the current necessary to operate the diode.

The output offset should be nulled for optimum performance by allowing the input to be its smallest expected value and adjusting R_1 for the proper output voltage.

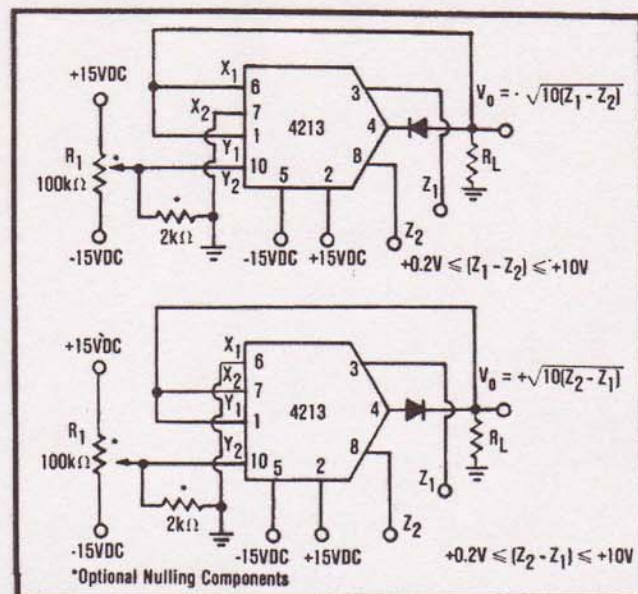


FIGURE 6. Square Root Connection.

This will improve the square root mode accuracy to about that of the multiply mode.

BRIDGE LINEARIZATION

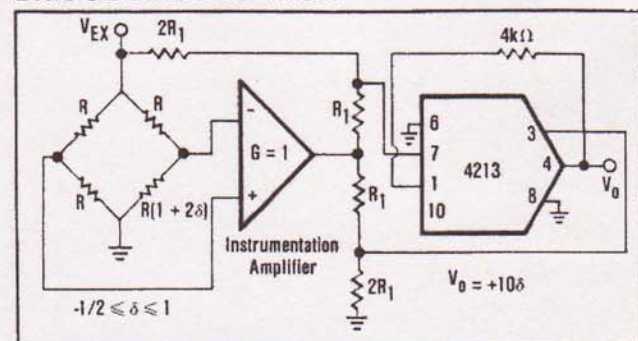


FIGURE 7. Bridge Linearization.

The use of the 4213 and the instrumentation amplifier to linearize the output from a bridge circuit makes the output V_0 independent of the bridge supply voltage.

TRUE RMS-TO-DC CONVERSION

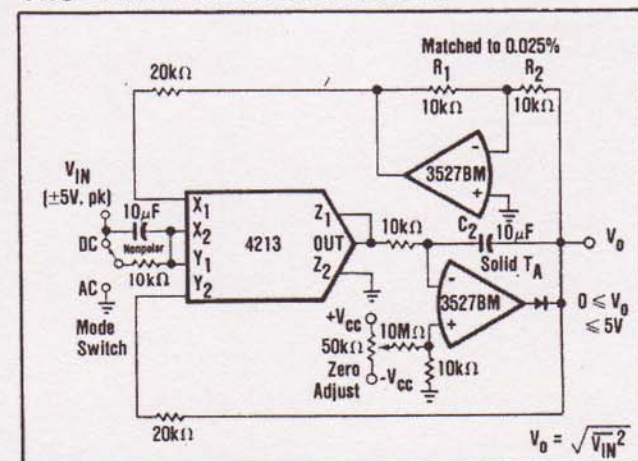
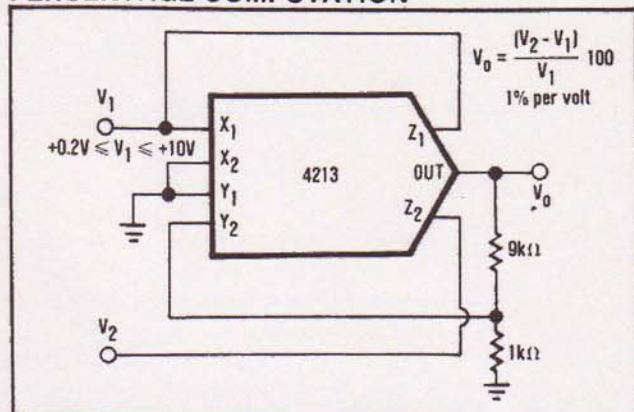


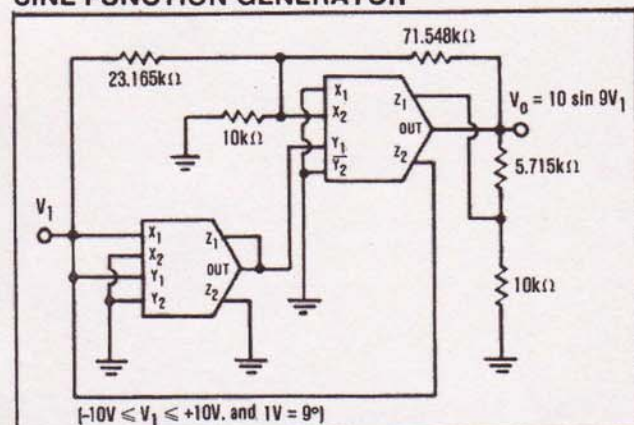
FIGURE 8. True RMS-to-DC Conversion.

PERCENTAGE COMPUTATION



The circuit of Figure 9 has a sensitivity of 1V/% and is capable of measuring 10% deviations. Wider deviation can be measured by decreasing the ratio of R_2/R_1 .

SINE FUNCTION GENERATOR



The circuit in Figure 10 uses implicit feedback to implement the following sine function approximation:

$$V_o = (1.5715V_1 - 0.004317V_1^3)/(1 + 0.001398V_1^2) \\ = 10 \sin(9V_1).$$

SINGLE-PHASE POWER MEASUREMENT

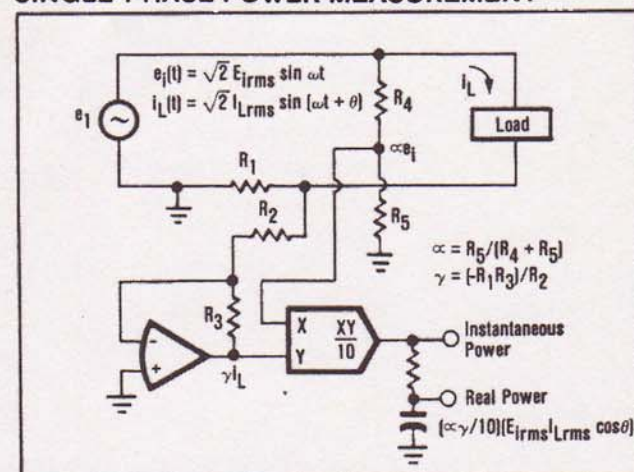


FIGURE 11. Single-Phase Instantaneous and Real Power Measurement.

In order to prevent frequency instability due to lead inductance of the power supply lines, each power supply should be bypassed. This should be done by connecting a $10\mu\text{F}$ tantalum capacitor in parallel with a 1000pF ceramic capacitor from the $+V_{\text{CC}}$ and $-V_{\text{CC}}$ pins of the 4213 to the power supply common. The connection of these capacitors should be as close to the 4213 as practical.

Stable operation is maintained with capacitive loads to 1000pF in all modes typically, except the square root mode for which 50pF is a safe upper limit. Higher capacitive loads can be driven if a 100Ω resistor is connected in series with the 4213's output.

The theory and procedures for developing virtually any function generator or linearization circuit can be found in the Burr-Brown/ McGraw Hill book "FUNCTION CIRCUITS - Design and Applications."

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